



# DCM™ in a VIA Package DC-DC Converter DCM3414x50M13C2yzz



## Isolated, Regulated DC Converter

### Features & Benefits

- Isolated, regulated DC-to-DC converter
- Up to 320 W, 26.67 A continuous
- 91.7% peak efficiency
- 183 W/in<sup>3</sup> Power density
- Wide input range 16 – 50 Vdc
- Safety Extra Low Voltage (SELV) 12.0 V Nominal Output
- ZVS high frequency switching
- Allows remote sense or local sense operation
- Tight regulation over all line and load conditions
- Fully operational current limit
- OV, OC, UV, short circuit and thermal protection
- Available in chassis mount and through hole VIA package  
3.380" x 1.400" x 0.370"  
(85.93 mm x 35.5 mm x 9.30 mm)

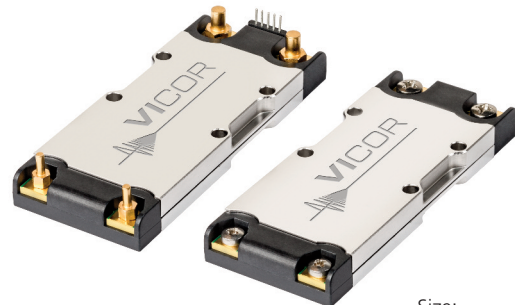
Product Ratings	
$V_{IN} = 16\text{ V to }50\text{ V}$	$P_{OUT} = 320\text{ W}$
$V_{OUT} = 12.0\text{ V}$ (7.2 V to 13.2 V Trim)	$I_{OUT} = 26.67\text{ A}$

### Product Description

The DCM in a VIA package is an Isolated, Regulated DC-to-DC Converter, operating from an unregulated, wide range input to generate an isolated 12.0 Vdc output. With its high frequency zero voltage switching (ZVS) topology, the DCM converter consistently delivers high efficiency across the input line range. The DCM provides tight output voltage regulation and offers a secondary-referenced control interface for trim, enable, and remote sense operation. DCM converters and downstream DC-DC products support efficient power distribution, providing superior power system performance and connectivity from a variety of unregulated power sources to the point-of-load. The VIA package offers flexible thermal management options with very low top and bottom side thermal impedances.

### Typical Applications

- Industrial
- Process control
- Automotive
- Heavy Equipment



Size:  
3.380 x 1.400 x 0.370 in  
85.93 x 35.5 x 9.30 mm

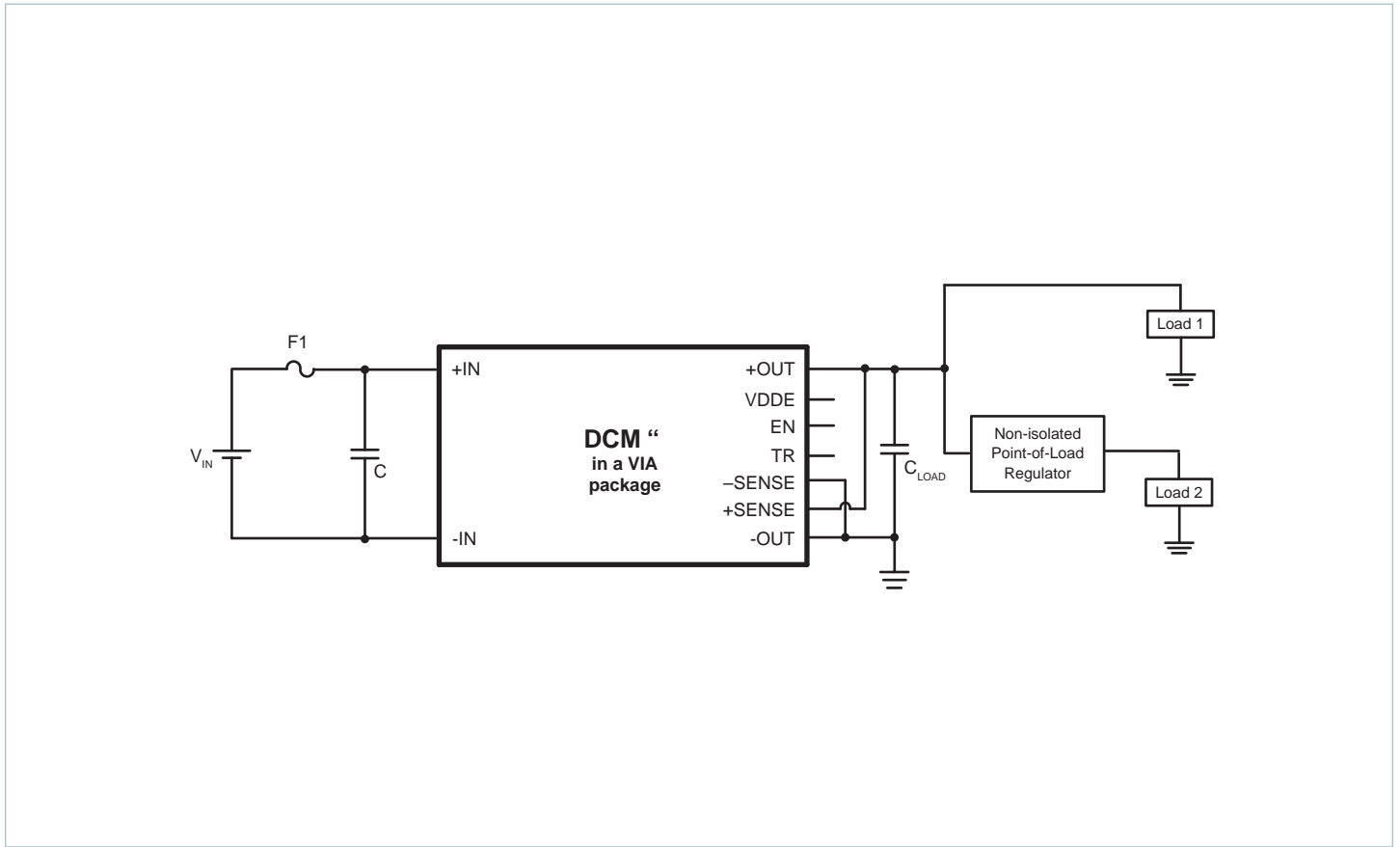
### Part Ordering Information

Product Function	Package Length	Package Width	Package Type	Max Input Voltage	Range Ratio	Max Output Voltage	Max Output Power	Product Grade (Case Temperature)	Option Field	
DCM	34	14	x	50	M	13	C2	y	z	z
DCM = DC-DC Converter	Length in Inches x 10	Width in Inches x 10	B = Board VIA V = Chassis VIA	Internal Reference				C = -20 to 100°C <sup>[1]</sup> T = -40 to 100°C <sup>[1]</sup>	01 = Chassis/Analog 05 = Short Pin/Analog 09 = Long Pin/Analog	

<sup>[1]</sup> High Temperature Power Derating may apply, see Thermal Specified Operating Area, Figure 1 on Page 4.



Typical Application



Typical Application: Single DCM3414x50M13C2yzz in Local Sense Operation, to a non-isolated regulator, and direct to load



### Absolute Maximum Ratings

The absolute maximum ratings below are stress ratings only. Operation at or beyond these maximum ratings can cause permanent damage to the device. Electrical specifications do not apply when operating beyond rated operating conditions.

Parameter	Comments	Min	Max	Unit
Input Voltage (+IN to -IN)		-0.5	65	V
Input Voltage Slew Rate			1	V/ $\mu$ s
TR to -OUT		-0.5	3.6	V
EN to -OUT		-0.5	3.6	V
VDDE to -OUT		0	12	V
-SENSE to -OUT		0	2.225	V
+OUT to +SENSE		0	2.225	V
+SENSE to -OUT		0	15.6	V
Output Voltage (+Out to -Out)		-0.5	16.3	V
Dielectric withstand (input to output)		1500		Vdc
Internal Operating Temperature	C Grade	-20	125	$^{\circ}$ C
	T Grade	-40	125	$^{\circ}$ C
	M Grade	-55	125	$^{\circ}$ C
Storage Temperature	C Grade	-20	125	$^{\circ}$ C
	T Grade	-40	125	$^{\circ}$ C
	M Grade	-65	125	$^{\circ}$ C
Average Output Current			26.7	A

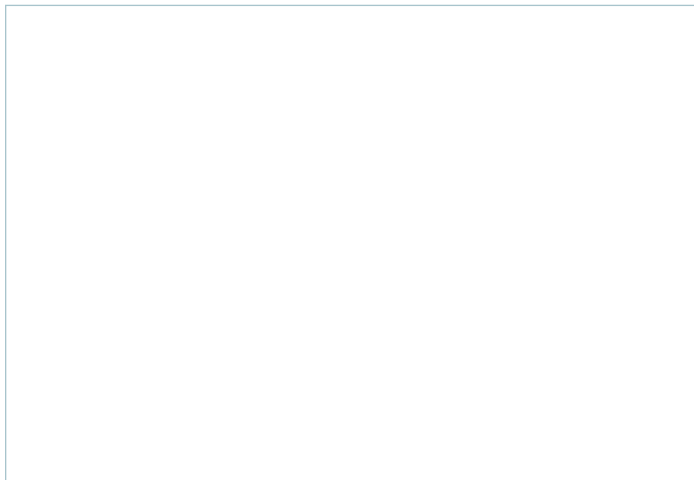


Figure 1 — Thermal Specified Operating Area: Max Output Power vs. Case Temp, Single unit at minimum full load efficiency

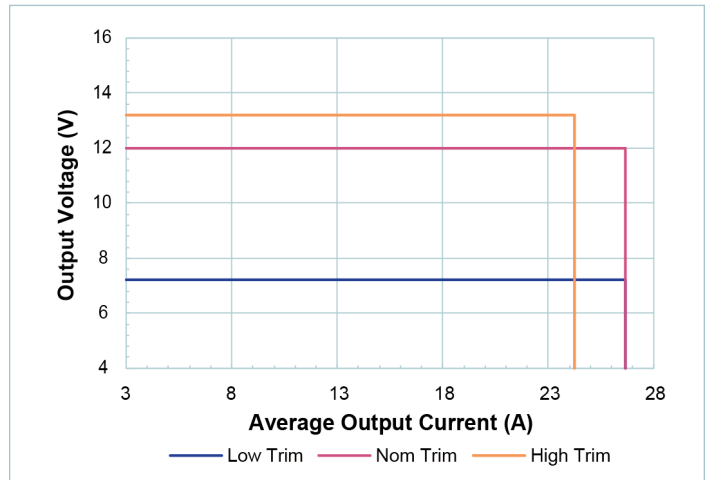


Figure 2 — Electrical Specified Operating Area

## Electrical Specifications

Specifications apply over all line, trim and load conditions, internal temperature  $T_{INT} = 25^{\circ}\text{C}$ , unless otherwise noted. **Boldface** specifications apply over the temperature range specified by the product grade.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
<b>Power Input Specification</b>						
Input voltage range	$V_{IN}$	Continuous operation	<b>16</b>	<b>28</b>	<b>50</b>	V
Inrush current (peak)	$I_{INRP}$	With maximum $C_{OUTEXT}$ , full resistive load			<b>30.0</b>	A
Input capacitance (internal)	$C_{IN-INT}$	Effective value at nominal input voltage		42.0		$\mu\text{F}$
Input capacitance (internal) ESR	$R_{CIN-INT}$	At 1 MHz		0.73		$\text{m}\Omega$
Input inductance (external)	$L_{IN}$	Differential mode, with no further line bypassing			<b>1</b>	$\mu\text{H}$
<b>No Load Specification</b>						
Input power – disabled	$P_Q$	Nominal line, see Fig. 3		0.5	0.7	W
		Worst case line, see Fig. 3			<b>0.8</b>	W
Input power – enabled with no load	$P_{NL}$	Nominal line, see Fig. 4		1.5	5.0	W
		Worst case line, see Fig. 4			<b>7.0</b>	W
<b>Power Output Specification</b>						
Output voltage set point	$V_{OUTNOM}$		11.94	12.0	12.06	V
Rated output voltage trim range	$V_{OUTTRIMMING}$	Trim range over temp. Specifies the Low, Nominal and High Trim conditions.	<b>7.2</b>	<b>12.0</b>	<b>13.2</b>	V
$V_{OUT}$ accuracy	$\%V_{OUT}$ ACCURACY	Total output voltage setpoint accuracy for all line conditions and for all load conditions above 10% of rated load, with trim inactive and SENSE pins connected (either at the load for remote sense, or at the DCM output for local sense)			<b>0.5</b>	%
$V_{OUT}$ accuracy with Sense Pins floating	$\%V_{OUT}$ ACCURACY-SF	Total output voltage setpoint accuracy for all line conditions and for all load conditions above 10% of rated load, with trim inactive and Sense Pins floating (Local Sense only)			<b>1.0</b>	%
$V_{OUT}$ accuracy with trim active	$\%V_{OUT}$ ACCURACY-TRIM	Total output voltage setpoint accuracy for all line conditions and for all load conditions above 10% of rated load, with trim active			<b>2.0</b>	%
$V_{OUT}$ accuracy light load	$\%V_{OUT}$ ACCURACY-LL	Total output voltage setpoint accuracy for all line and trim conditions, for load conditions below or equal to 10% of rated load			<b>5.0</b>	%
Rated output power	$P_{OUT}$	Continuous, $V_{OUT} \geq 12.0\text{ V}$	<b>320</b>			W
Rated output current	$I_{OUT}$	Continuous, $V_{OUT} \leq 12.0\text{ V}$	<b>26.67</b>			A
Output current limit	$I_{OUT-LIM}$	Of rated $I_{OUT}$ max. Fully operational current limit, for nominal trim and below	103	115	135	%
Current limit delay	$t_{IOUT-LIM}$	The module will power limit in a fast transient event		1		ms
Efficiency	$\eta$	Full load, nominal line, nominal trim	90.3	91.5		%
		Full load, over line and temperature, nominal trim	<b>87.0</b>			%
		50% load, over rated line, temperature and trim	<b>89.0</b>			%
Output voltage ripple	$V_{OUT-PP}$	Over all operating steady-state line, load and trim conditions, 20 MHz BW, with minimum $C_{OUTEXT}$		180		mV
Output capacitance (internal)	$C_{OUT-INT}$	Effective value at nominal output voltage		201		$\mu\text{F}$
Output capacitance (internal) ESR	$R_{COUT-INT}$	At 1 MHz		0.07		$\text{m}\Omega$

## Electrical Specifications (cont.)

Specifications apply over all line, trim and load conditions, internal temperature  $T_{INT} = 25^{\circ}\text{C}$ , unless otherwise noted. **Boldface** specifications apply over the temperature range specified by the product grade.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
<b>Power Output Specifications (Cont.)</b>						
Output capacitance (external)	$C_{OUT-EXT}$	Excludes component temperature coefficient. For load transients that remain > 2% rated load	<b>1000</b>		<b>10000</b>	$\mu\text{F}$
Output capacitance (external)	$C_{OUT-EXT-TRANSIENT}$	Excludes component temperature coefficient. For load transients down to 0% rated load	<b>10000</b>		<b>10000</b>	$\mu\text{F}$
Output capacitance, ESR (ext.)	$R_{COUT-EXT}$	At 10 kHz, excludes component tolerances	<b>10</b>			$\text{m}\Omega$
Initialization delay	$t_{INIT}$	See state diagram		25	<b>40</b>	ms
Output turn-on delay	$t_{ON}$	From rising edge EN, with VDDE pre-applied. See timing diagram		200		$\mu\text{s}$
Output turn-off delay	$t_{OFF}$	From falling edge EN. See timing diagram			<b>600</b>	$\mu\text{s}$
Soft start ramp time	$t_{SS}$	Full load (soft-start ramp time) with minimum $C_{OUT-EXT}$		300		ms
$V_{OUT}$ threshold for max rated load current	$V_{OUTFL-THRESH}$	During startup, $V_{OUT}$ must achieve this threshold before output can support full rated current			<b>6.0</b>	V
$I_{OUT}$ at startup	$I_{OUT-START}$	Max load current at startup while $V_{OUT}$ is below $V_{OUTFL-THRESH}$	<b>2.66</b>			A
Monotonic soft-start threshold voltage	$V_{OUT-MONOTONIC}$	Output voltage rise becomes monotonic with 1% of preload once it crosses $V_{OUT-MONOTONIC}$			<b>6.0</b>	V
Minimum required disabled duration	$t_{OFF-MIN}$	This refers to the minimum time a module needs to be in the disabled state before it will attempt to start via EN			<b>2</b>	ms
Minimum required disabled duration for predictable restart	$t_{OFF-MONOTONIC}$	This refers to the minimum time a module needs to be in the disabled state before it is guaranteed to exhibit monotonic soft-start and have predictable startup timing			<b>100</b>	ms
Voltage deviation (transient)	$\%V_{OUT-TRANS}$	Minimum $C_{OUT-EXT}$ (10 $\leftrightarrow$ 90% load step)		<10		%
Settling time	$t_{SETTLE}$			2.0		ms
<b>Powertrain Protections</b>						
Input Voltage Initialization threshold	$V_{IN-INIT}$	Threshold to start $t_{INIT}$ delay			<b>6</b>	V
Input Voltage Reset threshold	$V_{IN-RESET}$	Latching faults will clear once $V_{IN}$ falls below $V_{IN-RESET}$	<b>3</b>			V
Input undervoltage lockout threshold	$V_{IN-UVLO-}$		<b>12</b>		<b>14</b>	V
Input undervoltage recovery threshold	$V_{IN-UVLO+}$	See Timing diagram			<b>16</b>	V
Input overvoltage lockout threshold	$V_{IN-OVLO+}$				<b>60</b>	V
Input overvoltage recovery threshold	$V_{IN-OVLO-}$	See Timing diagram	<b>50</b>			V
Output overvoltage threshold	$V_{OUT-OVP}$	Latched shutdown	<b>15.6</b>			V
Minimum current limited $V_{OUT}$	$V_{OUT-LVP}$	Over all operating steady-state line and trim conditions			<b>4</b>	V
Overtemperature threshold (internal)	$T_{INT-OTP}$		<b>130</b>			$^{\circ}\text{C}$
Power limit	$P_{LIM}$				<b>630</b>	W
$V_{IN}$ overvoltage to cessation of powertrain switching	$t_{OVLO-SW}$	Independent of fault logic		1.5		$\mu\text{s}$
$V_{IN}$ overvoltage response time	$t_{OVLO}$	For fault logic only			<b>200</b>	$\mu\text{s}$
$V_{IN}$ undervoltage response time	$t_{UVLO}$				<b>100</b>	ms
Short circuit response time	$t_{SC}$	Powertrain on, operational state			<b>200</b>	$\mu\text{s}$
Short circuit, or temperature fault recovery time	$t_{FAULT}$	See Timing diagram		1		s

## Signal Specifications

Specifications apply over all line, trim and load conditions, internal temperature  $T_{INT} = 25^{\circ}\text{C}$ , unless otherwise noted. **Boldface** specifications apply over the temperature range specified by the product grade. **Please note:** For chassis mount models, Vicor part number 42550 will be needed for applications requiring the use of signal pins (Enable, Trim and Sense functions).<sup>[2]</sup>

Enable: EN								
<ul style="list-style-type: none"> <li>The EN pin enables and disables the DCM; when held low the unit will be disabled.</li> <li>The EN pin is activated only if VDDE is preapplied before <math>V_{IN}</math> is applied. Otherwise, EN is inactive and will be ignored until <math>V_{IN}</math> is removed and reapplied. Additionally, if VDDE is removed at any time, EN will return to inactive mode.</li> <li>The EN pin is referred to the –OUT of the converter and isolated from the primary side</li> </ul>								
SIGNAL TYPE	STATE	ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	NOM	MAX	UNIT
DIGITAL INPUT	Any	EN enable threshold	$V_{ENABLE-EN}$				<b>2.31</b>	V
		EN disable threshold	$V_{ENABLE-DIS}$	Needs VDDE preapplied	<b>0.99</b>			V
		Internally generated $V_{CC}$	$V_{CC}$		<b>3.23</b>	3.30	<b>3.37</b>	V
		EN internal pull up resistance to $V_{CC}$	$R_{ENABLE-INT}$		<b>0.990</b>	1.000	<b>1.010</b>	k $\Omega$

Trim: TR								
<ul style="list-style-type: none"> <li>The TR pin enables and disables trim functionality when <math>V_{IN}</math> is initially applied to the DCM converter. When <math>V_{in}</math> first crosses <math>V_{IN-UVLO+}</math>, the voltage on TR determines whether or not trim is active.</li> <li>If TR is not floating at power up and has a voltage less than TR trim enable threshold, trim is active.</li> <li>If trim is active, the TR pin provides dynamic trim control with at least 250 Hz of -3dB control bandwidth over the output voltage of the DCM converter.</li> <li>The TR pin has an internal pull-up to <math>V_{CC}</math> and is referenced to the –OUT pin of the converter.</li> <li><math>V_{TRIM-RANGE}</math> represents the trim pin functional limits only. Module performance is guaranteed within rated output voltage trim range <math>V_{OUT-TRIMMING}</math>, see electrical specification on page 5.</li> </ul>								
SIGNAL TYPE	STATE	ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	NOM	MAX	UNIT
DIGITAL INPUT	Startup	TR trim disable threshold	$V_{TRIM-DIS}$	Trim disabled when TR above this threshold at power up			<b>3.20</b>	V
		TR trim enable threshold	$V_{TRIM-EN}$	Trim enabled when TR below this threshold at power up	<b>3.10</b>			V
ANALOG INPUT	Operational with Trim enabled	Internally generated $V_{CC}$	$V_{CC}$		<b>3.23</b>	3.30	<b>3.37</b>	V
		TR pin functional range	$V_{TRIM-RANGE}$	Functional limits only	0.00	2.37	3.10	V
		$V_{OUT}$ referred TR pin resolution	$V_{OUT-RES}$	With $V_{CC} = 3.3\text{ V}$		8.79		mV
		TR internal pull up resistance to $V_{CC}$	$R_{TRIM-INT}$		<b>0.999</b>	1.000	<b>1.001</b>	k $\Omega$

VDDE								
<ul style="list-style-type: none"> <li>VDDE powers the internal controller.</li> <li>VDDE needs to be preapplied before <math>V_{IN}</math> in order to activate EN functionalities.</li> <li>If not preapplied, VDDE is derived from <math>V_{OUT}</math>; however, in this case, the enable function is not activated (the unit is always enabled and can be disabled only by removing <math>V_{IN}</math>.)</li> <li>If VDDE is removed during operation, the unit will return in "always enabled" mode, ignoring the EN pin until VDDE is reapplied and <math>V_{IN}</math> is cycled.</li> </ul>								
SIGNAL TYPE	STATE	ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	NOM	MAX	UNIT
POWER INPUT	Any	Power input for internal controller	VDDE		4	5	10	V
		VDDE current consumption	$I_{VDDE}$			35	50	mA

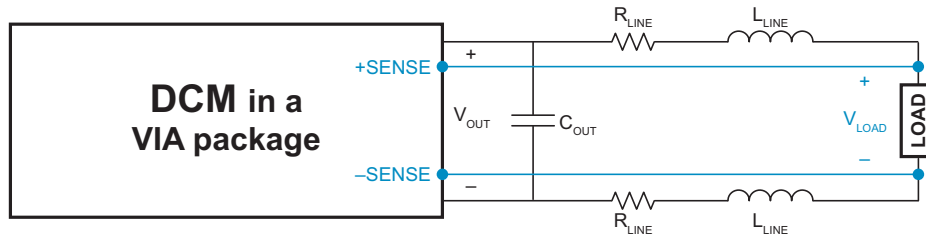
<sup>[2]</sup> Signal cable 42550 is rated for up to 5 insertions and extractions. To avoid unnecessary stress on the connector, the cable should be tied to the chassis.

Signal Specifications (Cont.)

Specifications apply over all line, trim and load conditions, internal temperature  $T_{INT} = 25^{\circ}C$ , unless otherwise noted. **Boldface** specifications apply over the temperature range specified by the product grade. **Please note:** For chassis mount models, Vicor part number 42550 will be needed for applications requiring the use of signal pins (Enable, Trim and Sense functions).

**+SENSE, -SENSE**

- Provide Remote Sense capability.
- If floating, the DCM automatically implements Local Sense Operation. To achieve maximum regulation accuracy in local sense, the SENSE pins should be connected directly to their respective OUT pins. If SENSE pins are floating, the regulation accuracy degrades (see "V<sub>OUT</sub> accuracy with Sense Pins floating" on Page 5).



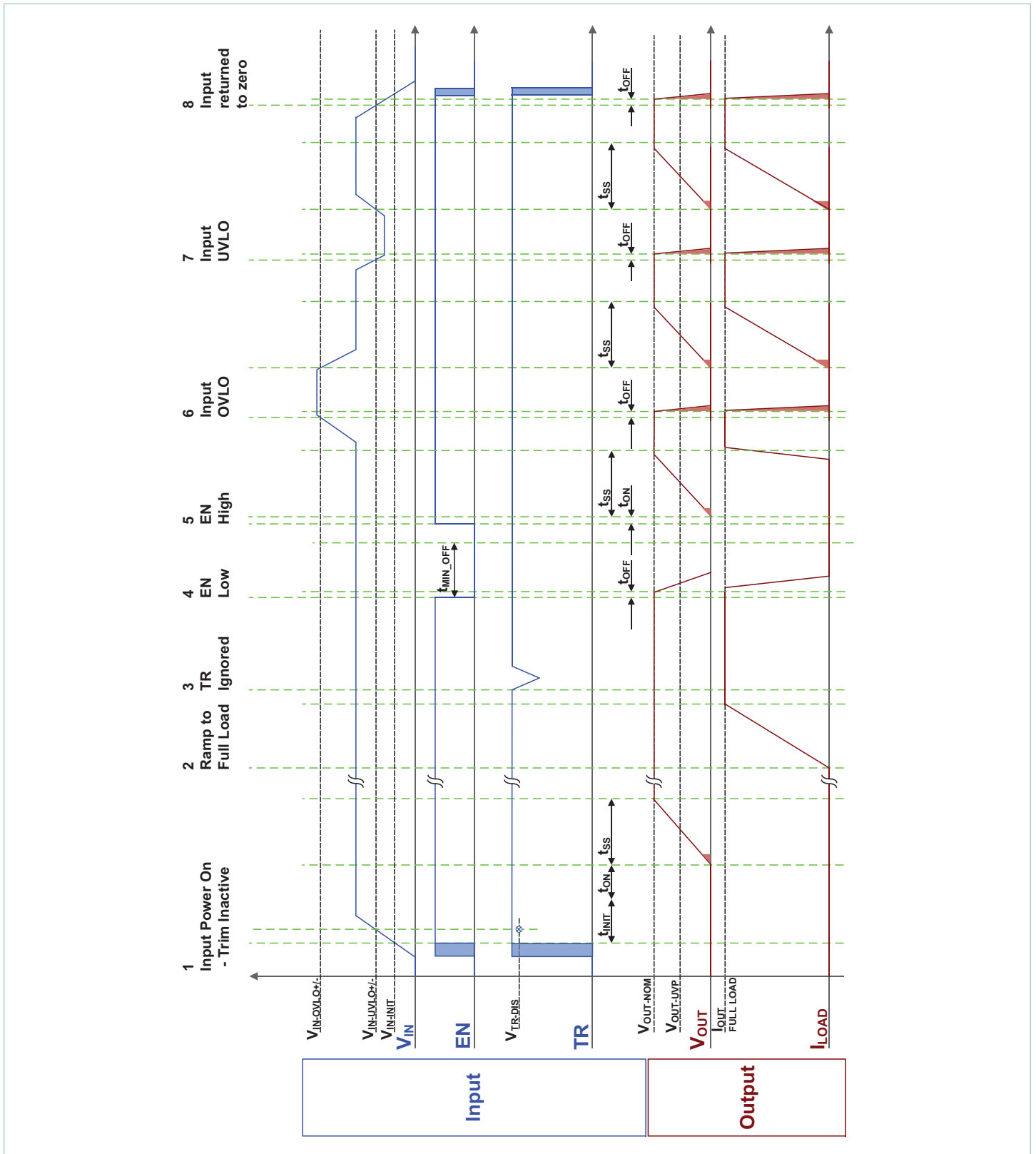
- Module performance is guaranteed for  $\Delta V_{OUT\_TO\_LOAD}$  within rated limits specified below. For  $\Delta V_{OUT\_TO\_LOAD}$  higher than the specified limits, regulation accuracy is not guaranteed. Also, high  $\Delta V_{OUT\_TO\_LOAD}$  might trigger OVP (for above nominal trim conditions), and might cause load voltage runaway (which might trigger UVP).

ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	NOM	MAX	UNIT
Remote Sense rated Voltage Drop	$\Delta V_{OUT\_TO\_LOAD}$	Rated voltage drop between DCM package output and sense point at the load, in Remote Sense Operation. Module performance is guaranteed for $\Delta V_{OUT\_TO\_LOAD}$ below this threshold.	<b>120</b>			mV



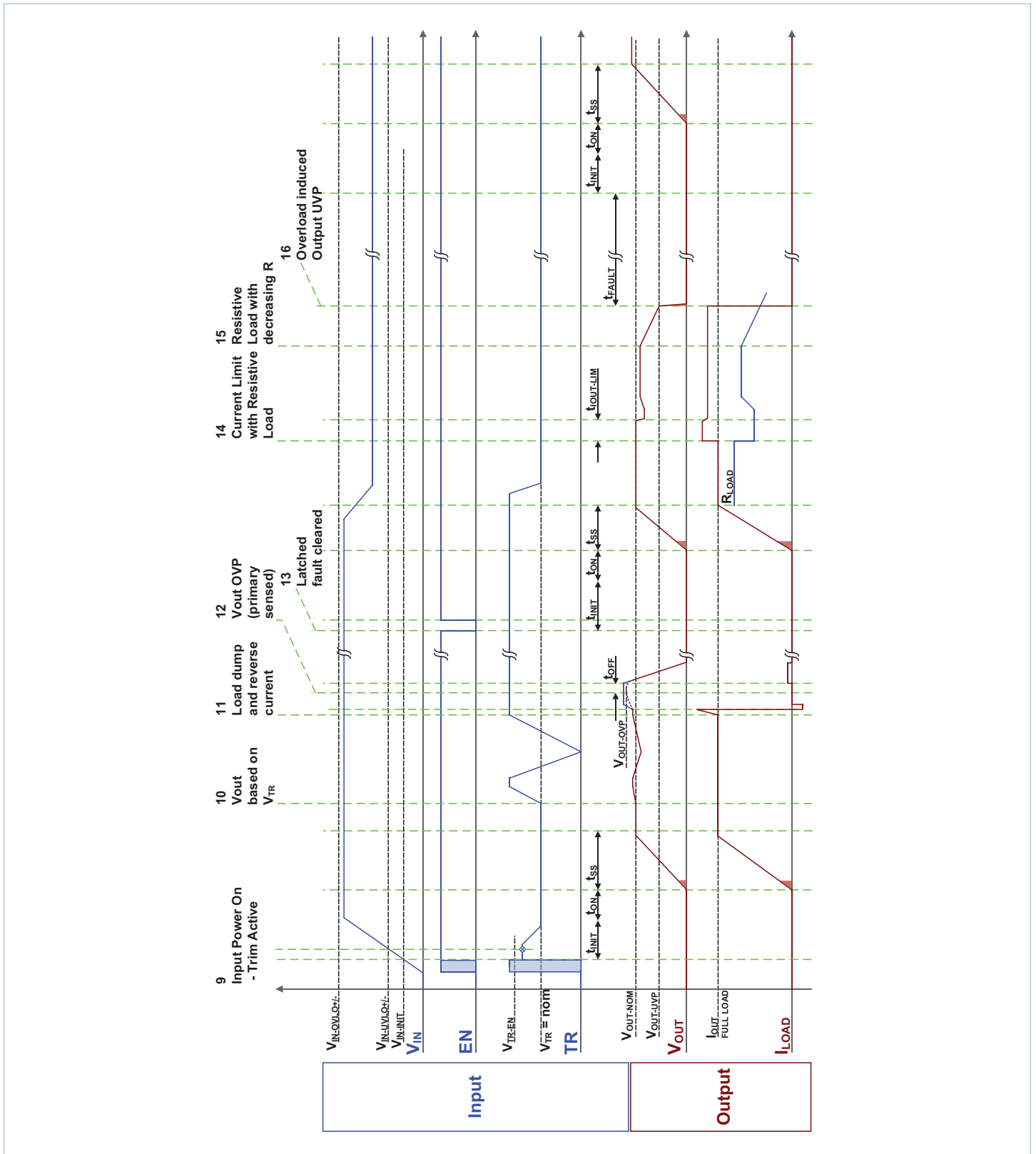
Timing Diagrams

Module Inputs are shown in blue; Module Outputs are shown in brown.



Timing Diagrams (Cont.)

Module Inputs are shown in blue; Module Outputs are shown in brown.



Typical Performance Characteristics

The following figures present typical performance at  $T_C = 25^\circ\text{C}$ , unless otherwise noted. See associated figures for general trend data.

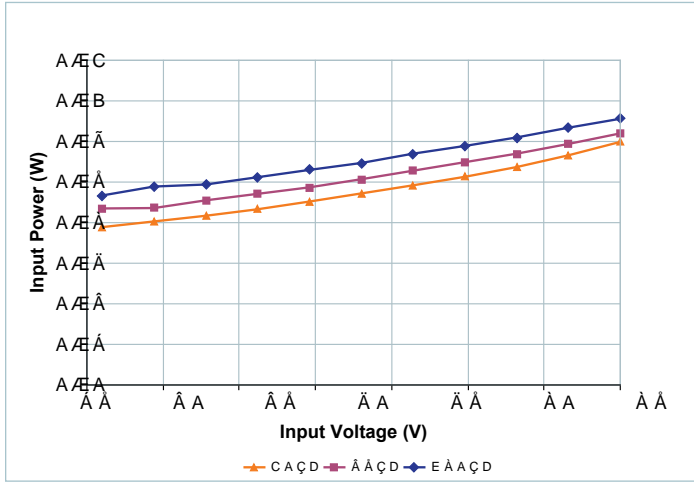


Figure 3 — Disabled power dissipation vs.  $V_{IN}$

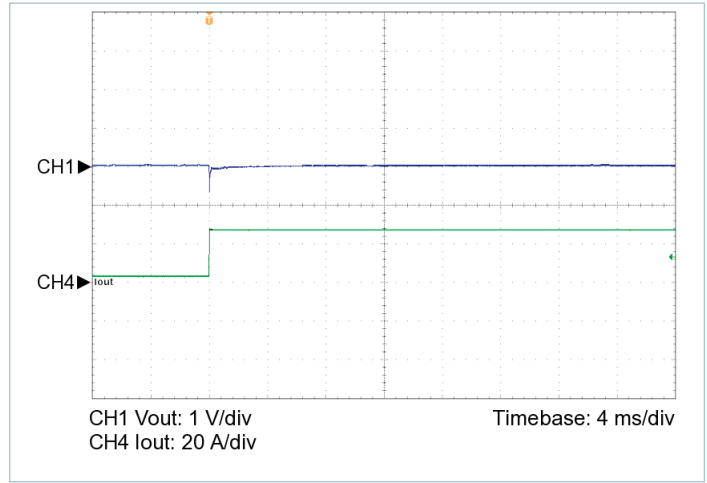


Figure 6 — 10% to 100% load transient response,  $V_{IN} = 28\text{ V}$ , nominal trim,  $C_{OUT\_EXT} = 1000\ \mu\text{F}$

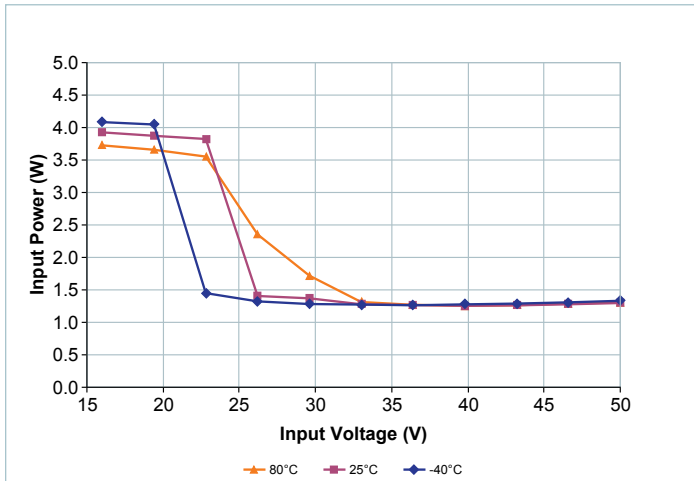


Figure 4 — No load power dissipation vs.  $V_{IN}$  at nominal trim

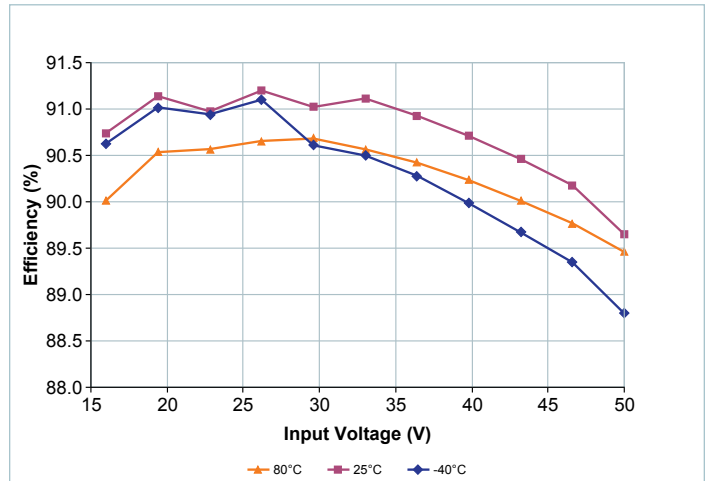


Figure 7 — Full Load Efficiency vs.  $V_{IN}$  at low trim

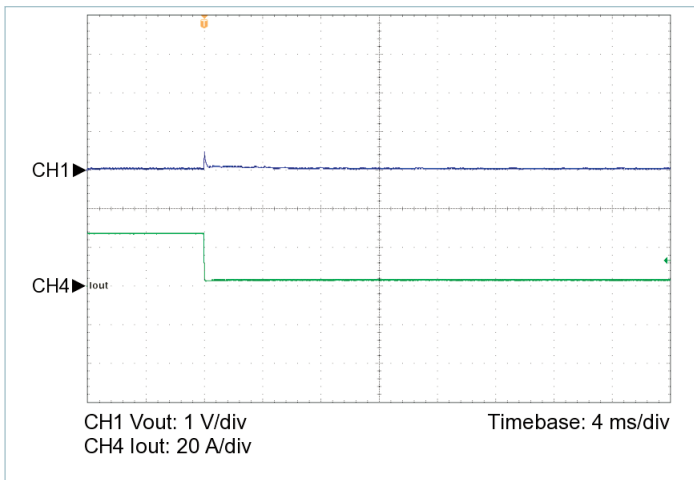


Figure 5 — 100% to 10% load transient response,  $V_{IN} = 28\text{ V}$ , nominal trim,  $C_{OUT\_EXT} = 1000\ \mu\text{F}$

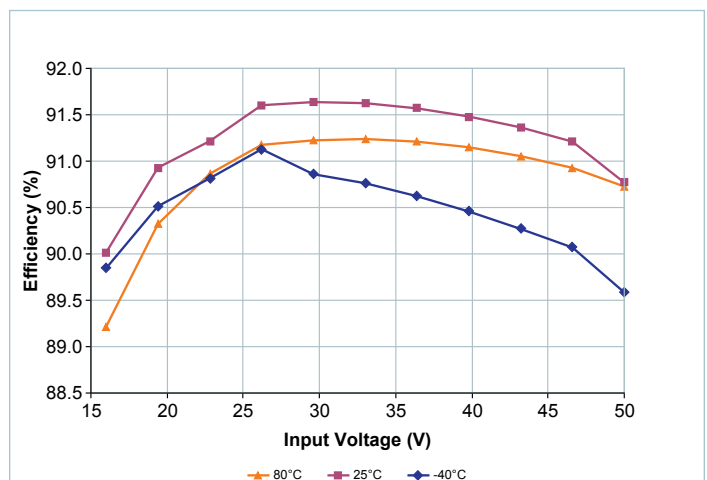


Figure 8 — Full Load Efficiency vs.  $V_{IN}$  at nominal trim

Typical Performance Characteristics (cont.)

The following figures present typical performance at  $T_C = 25^\circ\text{C}$ , unless otherwise noted. See associated figures for general trend data.

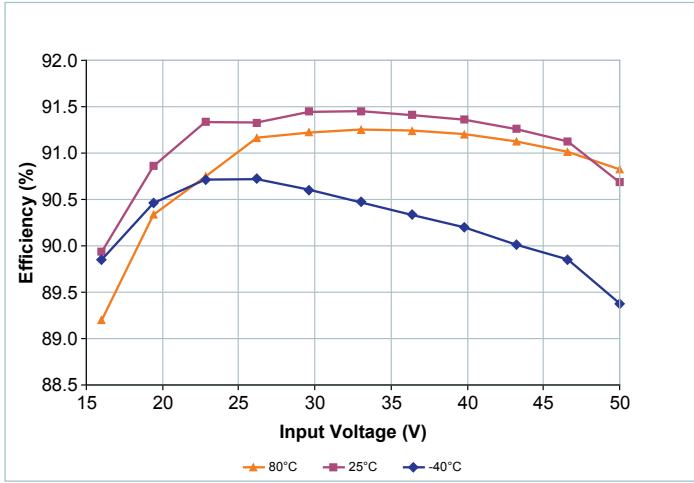


Figure 9 — Full Load Efficiency vs.  $V_{IN}$  at high trim

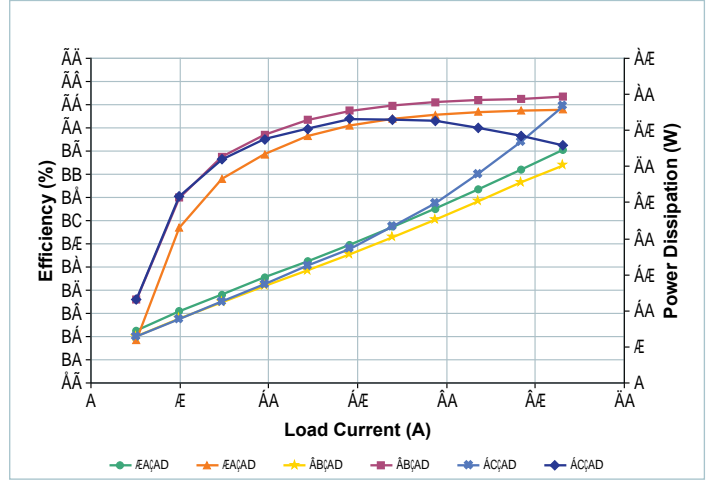


Figure 12 — Efficiency (top three curves) and power dissipation (bottom three curves) vs. load at  $T_{CASE} = 80^\circ\text{C}$ , nominal trim

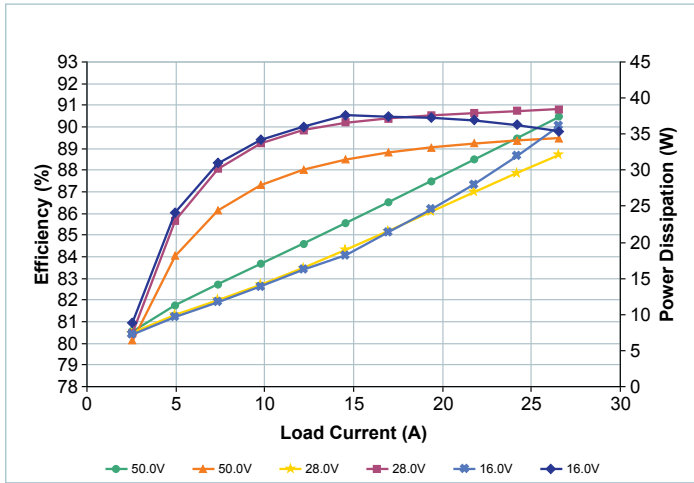


Figure 10 — Efficiency (top three curves) and power dissipation (bottom three curves) vs. load at  $T_{CASE} = -40^\circ\text{C}$ , nominal trim

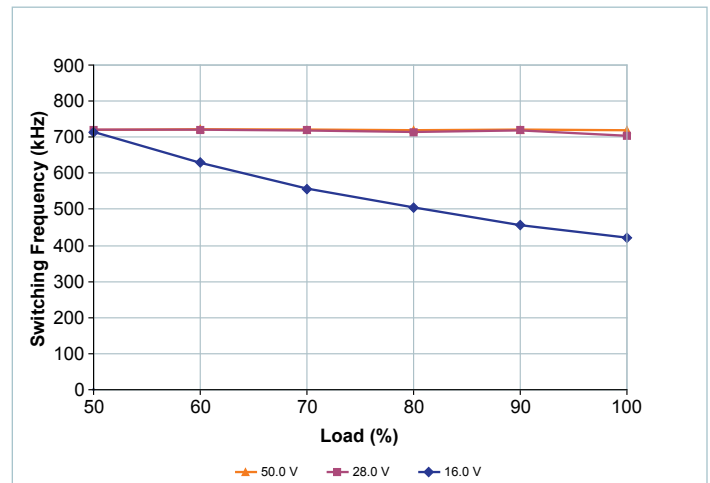


Figure 13 — Nominal powertrain switching frequency vs. load, at nominal trim

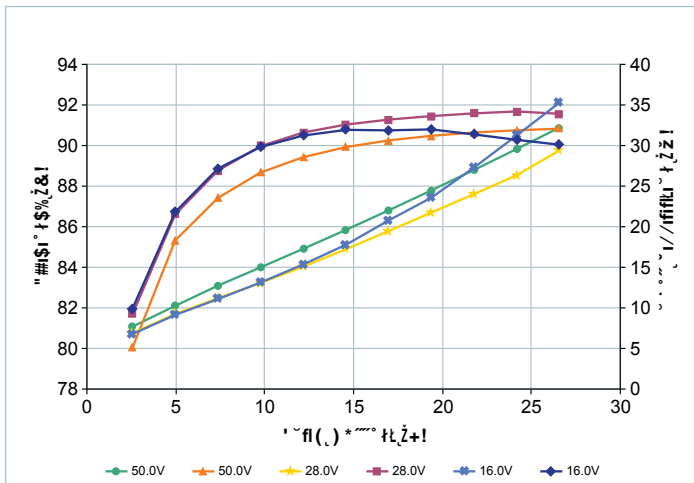


Figure 11 — Efficiency (top three curves) and power dissipation (bottom three curves) vs. load at  $T_{CASE} = 25^\circ\text{C}$ , nominal trim



Figure 14 — Effective internal input capacitance vs. applied voltage

Typical Performance Characteristics (cont.)

The following figures present typical performance at  $T_C = 25^\circ\text{C}$ , unless otherwise noted. See associated figures for general trend data.

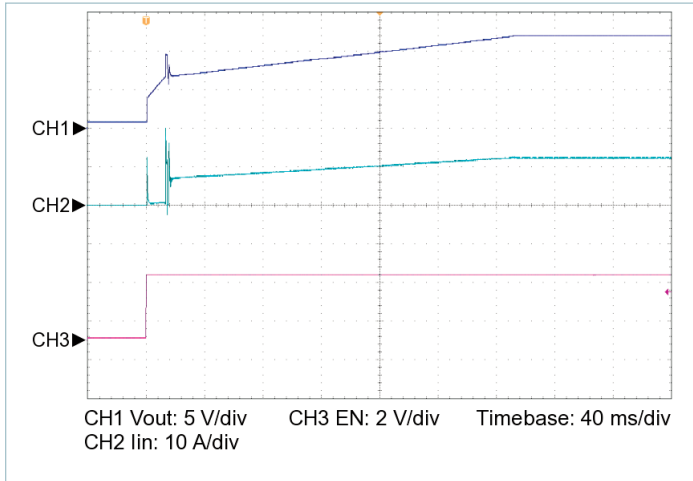


Figure 15 — Startup from EN,  $V_{IN} = 28\text{ V}$ ,  $C_{OUT\_EXT} = 10000\ \mu\text{F}$ ,  $R_{LOAD} = 0.450\ \Omega$

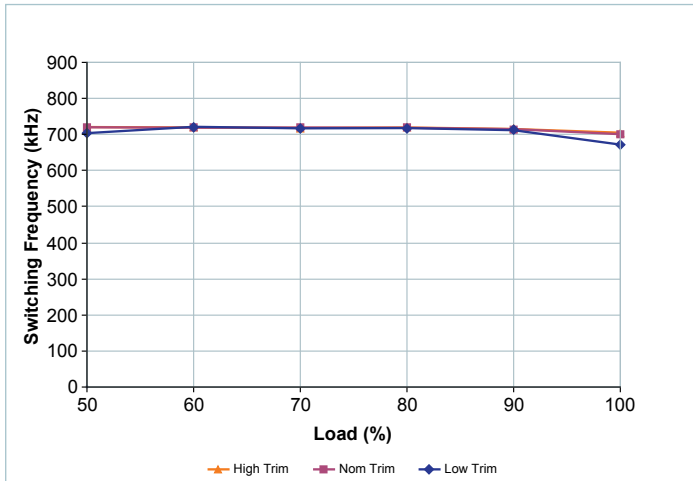


Figure 16 — Nominal powertrain switching frequency vs. load, at nominal  $V_{IN}$

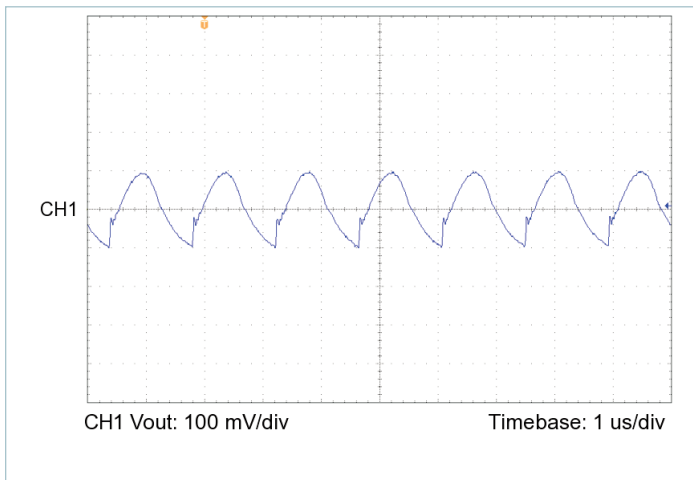


Figure 17 — Output voltage ripple,  $V_{IN} = 28\text{ V}$ ,  $V_{OUT} = 12.0\text{ V}$ ,  $C_{OUT\_EXT} = 1000\ \mu\text{F}$ ,  $R_{LOAD} = 0.450\ \Omega$

## General Characteristics

Specifications apply over all line, trim and load conditions, internal temperature  $T_{INT} = 25^{\circ}\text{C}$ , unless otherwise noted. **Boldface** specifications apply over the temperature range specified by the product grade.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
<b>Mechanical</b> <sup>[3]</sup>						
Length	L		85.68/[3.37]	85.93/[3.38]	86.18/[3.3900]	mm/[in]
Width	W		35.29/[1.39]	35.54/[1.4]	35.79/[1.41]	mm/[in]
Height	H		8.919/[0.355]	9.3/[0.37]	9.6810/[0.385]	mm/[in]
Volume	Vol	No heat sink		28.40/[1.75]		cm <sup>3</sup> /[in <sup>3</sup> ]
Weight	W			101.0/[3.56]		g/[oz]
<b>Thermal</b>						
Operating internal temperature	$T_{INT}$	C-Grade	-20		125	°C
		T-Grade	-40		125	°C
		M-Grade	-55		125	°C
Thermal resistance top side	$R_{JC\_TOP}$	Estimated thermal resistance to maximum temperature internal component from isothermal top housing		1.31		°C/W
Thermal resistance housing	$R_{HOU}$	Estimated thermal resistance from top housing to bottom housing		0.55		°C/W
Thermal resistance bottom side	$R_{JC\_BOT}$	Estimated thermal resistance to maximum temperature internal component from isothermal bottom housing		2.33		°C/W
Thermal capacity				35.0		Ws/°C
<b>Assembly</b>						
Storage temperature	$T_{ST}$	C-Grade	-20		125	°C
		T-Grade	-40		125	°C
		M-Grade	-65		125	°C
ESD rating	HBM	Method per Human Body Model Test ESDA/JEDEC JDS-001-2012	CLASS 1C			V
	CDM	Charged Device Model JESD22-C101E	CLASS 2			
<b>Soldering</b> <sup>[4]</sup>						
Peak temperature top case		For further information, please contact factory applications			130	°C

<sup>[3]</sup> Product appearance may change over time depending upon environmental exposure. This change has no impact on product performance.

<sup>[4]</sup> Product is not intended for reflow solder attach.

## General Characteristics (Cont.)

Specifications apply over all line, trim and load conditions, internal temperature  $T_{INT} = 25^{\circ}\text{C}$ , unless otherwise noted. **Boldface** specifications apply over the temperature range specified by the product grade.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
<b>Safety</b>						
Dielectric Withstand Test	$V_{HIPOUT}$	IN to OUT	<b>1500</b>			Vdc
		IN to CASE	<b>1500</b>			Vdc
		OUT to CASE	<b>707</b>			Vdc
<b>Reliability</b>						
MTBF		MIL-HDBK-217Plus Parts Count - 25°C Ground Benign, Stationary, Indoors / Compute		1.70		MHrs
		Telcordia Issue 2 - Method I Case III; 25°C Ground Benign, Controlled		2.20		MHrs
<b>Agency Approvals</b>						
Agency approvals/standards						
		CE Marked for Low Voltage Directive and RoHS Recast Directive, as applicable				

## Pin Functions

### +IN, -IN

Input power pins.

### +OUT, -OUT

Output power pins. –Out also serves as reference for the secondary-referenced control pins.

### EN (Enable)

This pin enables and disables the DCM converter; when held low the unit will be disabled. It is referenced to the -OUT pin of the converter. EN is active only if VDDE is preapplied before  $V_{IN}$  is applied. Otherwise, EN is inactive and will be ignored until  $V_{IN}$  is removed and reapplied.

**Output enable:** When EN is allowed to pull up above the enable threshold, the module will be enabled. If leaving EN floating, it is pulled up to  $V_{CC}$  and the module will be enabled.

**Output disable:** EN may be pulled down externally in order to disable the module.

EN is an input only, it does not pull low in the event of a fault.

### TR (Trim)

The TR pin is used to select the trim mode and to trim the output voltage of the DCM converter. The TR pin has an internal pull-up to  $V_{CC}$ .

The DCM package will latch trim behavior at application of  $V_{IN}$  (once  $V_{IN}$  exceeds  $V_{IN-UVLO+}$ ), and persist in that same behavior until loss of input voltage.

At application of  $V_{IN}$ , if TR is sampled at above  $V_{TRIM-DIS}$ , the module will latch in a non-trim mode, and will ignore the TR input for as long as  $V_{IN}$  is present.

At application of  $V_{IN}$ , if TR is sampled at below  $V_{TRIM-EN}$ , the TR will serve as an input to control the real time output voltage. It will persist in this behavior until  $V_{IN}$  is no longer present.

If trim is active when the DCM is operating, the TR pin provides dynamic trim control at a typical 250 Hz of -3dB bandwidth over the output voltage. TR also decreases the current limit threshold when trimming above  $V_{OUT-NOM}$ .

### +SENSE, -SENSE

These pins are Remote Sense pins, which allow the users to sense the voltage at the point of load so that the DCM can use the load voltage to regulate its output voltage accordingly. If “+SENSE” and “-SENSE” are floating, Local Sense is implemented automatically.

However, when SENSE pins are floating, the regulation accuracy deteriorates (see “ $V_{OUT}$  accuracy with Sense pin floating” on Page 5). To achieve maximum regulation accuracy, the SENSE pins should be used also for local sense operation, by connecting them directly to their respective OUT pins (+SENSE to +OUT and -SENSE to -OUT).

**Please note:** For chassis mount models, Vicor part number 42550 will be needed for applications requiring the use of signal pins (Enable, Trim and Sense functions).

## Design Guidelines

### Building Blocks and System Design

The DCM converter input accepts the full 16.0 to 50.0 V range, and it generates an isolated trimmable 12.0 Vdc output.

The DCM converter provides a tightly regulated output voltage. With trim inactive (TR pin floating), regulation accuracy is within 0.5% of the setpoint for all line conditions and for any load above 10% of the rated load. With trim active, regulation accuracy is within 2.0% for all line conditions and for any load condition above 10% of the rated load.

The DCM3414x50M13C2yzz is designed to be used in applications where the output power requirements are up to 320 W.

### Soft Start

When the DCM starts, it will go through a soft start. The soft start routine ramps the output voltage by modulating the internal error amplifier reference. This causes the output voltage to approximate a piecewise linear ramp. The output ramp finishes when the voltage reaches either the nominal output voltage, or the trimmed output voltage in cases where trim mode is active.

During soft-start, the maximum load current capability is reduced. Until  $V_{out}$  achieves at least  $V_{OUT-FL-THRESH}$ , the output current must be less than  $I_{OUT-START}$  in order to guarantee startup. Note that this is current available to the load, above that which is required to charge the output capacitor.

### Trim Mode and Output Trim Control

When the input voltage is initially applied to a DCM, and after  $t_{INIT}$  elapses, the trim pin voltage  $V_{TR}$  is sampled. The TR pin has an internal pull up resistor to  $V_{CC}$ , so unless external circuitry pulls the pin voltage lower, it will pull up to  $V_{CC}$ . If the initially sampled trim pin voltage is higher than  $V_{TRIM-DIS}$ , then the DCM will disable trimming as long as the  $V_{IN}$  remains applied. In this case, for all subsequent operation the output voltage will be programmed to the nominal. This minimizes the support components required for applications that only require the nominal rated  $V_{OUT}$ , and also provides the best output setpoint accuracy, as there are no additional errors from external trim components.

If at initial application of  $V_{IN}$ , the TR pin voltage is prevented from exceeding  $V_{TRIM-EN}$ , then the DCM will activate trim mode, and it will remain active for as long as  $V_{IN}$  is applied.

$V_{OUT}$  set point can be calculated using the equation below:

$$V_{OUT-TRIMMING} = 6.600 + (7.512 \cdot V_{TR}/V_{CC}) \quad (1)$$

Note that the trim mode is not changed when a DCM recovers from any fault condition or being disabled.

Module performance is guaranteed through output voltage trim range  $V_{OUT-TRIMMING}$ . If  $V_{OUT}$  is trimmed above this range, then certain combinations of line and load transient conditions may trigger the output OVP.

### Output Current Limit

The DCM features a fully operational current limit which effectively keeps the module operating inside the Safe Operating Area (SOA) for all valid trim and load profiles. The current limit approximates a “brick wall” limit, where the output current is



prevented from exceeding the current limit threshold by reducing the output voltage via the internal error amplifier reference. The current limit threshold at nominal trim and below is typically 115% of rated output current, but it can vary between 103% to 135%. In order to preserve the SOA, when the converter is trimmed above the nominal output voltage, the current limit threshold is automatically reduced to limit the available output power.

When the output current exceeds the current limit threshold, current limit action is held off by 1ms, which permits the DCM to momentarily deliver higher peak output currents to the load. Peak output power during this time is still constrained by the internal Power Limit of the module. The fast Power Limit and relatively slow Current Limit work together to keep the module inside the SOA. Delaying entry into current limit also permits the DCM to minimize droop voltage for load steps.

Sustained operation in current limit is permitted, and no derating of output power is required.

Current limit can reduce the output voltage to as little as the UVP threshold ( $V_{OUT-UVLP}$ ). Below this minimum output voltage compliance level, further loading will cause the module to shut down due to the output undervoltage fault protection.

#### Line Impedance, Input Slew rate and Input Stability Requirements

Connect a high-quality, low-noise power supply to the +IN and -IN terminals. Additional capacitance may have to be added between +IN and -IN to make up for impedances in the interconnect cables as well as deficiencies in the source.

Excessive source impedance can bring about system stability issues for a regulated DC-DC converter, and must either be avoided or compensated. A 1000  $\mu$ F input capacitor is the minimum recommended in case the source impedance is insufficient to satisfy stability requirements.

Additional information can be found in the filter design application note:

[www.vicorpower.com/documents/application\\_notes/vichip\\_appnote23.pdf](http://www.vicorpower.com/documents/application_notes/vichip_appnote23.pdf)

Please refer to this input filter design tool to ensure input stability: <http://app2.vicorpower.com/filterDesign/intiFilter.do>.

Ensure that the input voltage slew rate is less than 1V/us, otherwise a pre-charge circuit is required for the DCM input to control the input voltage slew rate and prevent overstress to input stage components.

#### Input Fuse Selection

The DCM is not internally fused in order to provide flexibility in configuring power systems. Input line fusing is recommended at the system level, in order to provide thermal protection in case of catastrophic failure. The fuse shall be selected by closely matching system requirements with the following characteristics:

- Current rating (usually greater than the DCM converter's maximum current)

- Maximum voltage rating (usually greater than the maximum possible input voltage)

- Ambient temperature

- Breaking capacity per application requirements

- Nominal melting  $I^2t$

Recommended fuse: See Agency Approvals for Recommended Fuse <http://www.vicorpower.com>

## Fault Handling

### Input Undervoltage Fault Protection (UVLO)

The converter's input voltage is monitored to detect an input under voltage condition. If the converter is not already running, then it will ignore enable commands until the input voltage is greater than  $V_{IN-UVLO+}$ . If the converter is running and the input voltage falls below  $V_{IN-UVLO-}$ , the converter recognizes a fault condition, the powertrain stops switching, and the output voltage of the unit falls.

Input voltage transients which fall below UVLO for less than  $t_{UVLO}$  may not be detected by the fault protection logic, in which case the converter will continue regular operation. No protection is required in this case.

Once the UVLO fault is detected by the fault protection logic, the converter shuts down and waits for the input voltage to rise above  $V_{IN-UVLO+}$ . Provided the converter is still enabled, it will then restart.

### Input Overvoltage Fault Protection (OVLO)

The converter's input voltage is monitored to detect an input over voltage condition. When the input voltage is more than the  $V_{IN-OVLO+}$ , a fault is detected, the powertrain stops switching, and the output voltage of the converter falls.

After an OVLO fault occurs, the converter will wait for the input voltage to fall below  $V_{IN-OVLO-}$ . Provided the converter is still enabled, the powertrain will restart.

The powertrain controller itself also monitors the input voltage. Transient OVLO events which have not yet been detected by the fault sequence logic may first be detected by the controller if the input slew rate is sufficiently large. In this case, powertrain switching will immediately stop. If the input voltage falls back in range before the fault sequence logic detects the out of range condition, the powertrain will resume switching and the fault logic will not interrupt operation. Regardless of whether the powertrain is running at the time or not, if the input voltage does not recover from OVLO before  $t_{OVLO}$ , the converter fault logic will detect the fault.

### Output Undervoltage Fault Protection (UVP)

The converter determines that an output overload or short circuit condition exists by measuring its output voltage and the output of the internal error amplifier. In general, whenever the powertrain is switching and the output voltage falls below  $V_{OUT-UVP}$  threshold, a short circuit fault will be registered. Once an output undervoltage condition is detected, the powertrain immediately stops switching, and the output voltage of the converter falls. The converter remains disabled for a time  $t_{FAULT}$ . Once recovered and provided the converter is still enabled, the powertrain will again enter the soft start sequence after  $t_{INIT}$  and  $t_{ON}$ .

### Temperature Fault Protections (OTP)

The fault logic monitors the internal temperature of the converter. If the measured temperature exceeds  $T_{INT-OTP}$ , a temperature fault is registered. As with the under voltage fault protection, once a temperature fault is registered, the powertrain immediately stops switching, the output voltage of the converter falls, and the converter remains disabled for at least time  $t_{FAULT}$ . Then, the converter waits for the internal temperature to return to below  $T_{INT-OTP}$  before recovering. Provided the converter is still enabled, the DCM will restart after  $t_{INIT}$  and  $t_{ON}$ .

**Output Overvoltage Fault Protection (OVP)**

The converter monitors the output voltage during each switching cycle. If the output voltage exceeds  $V_{OUT-OVP}$ , the OVP fault protection is triggered. The control logic disables the powertrain, and the output voltage of the converter falls.

This type of fault is latched, and the converter will not start again until the latch is cleared. Clearing the fault latch is achieved by either disabling the converter via the EN pin, or else by removing the input power such that the input voltage falls below  $V_{IN-INIT}$ .

**External Output Capacitance**

The DCM converter internal compensation requires a minimum external output capacitor. An external capacitor in the range of 1000 to 10000  $\mu F$  with ESR of 10 m $\Omega$  is required for control loop compensation purposes.

However some DCM models require an increase to the minimum external output capacitor value in certain loading condition. In applications where the load can go below 2% of rated load, the range of output capacitor required is given by

$C_{OUT-EXT-TRANSIENT}$  in the Electrical Specifications table.

**Thermal Considerations**

The VIA package provides effective conduction cooling from either of the two module surfaces. Heat may be removed from the top surface, the bottom surface or both. The extent to which these two surfaces are cooled is a key component for determining the maximum power that can be processed by a DCM in a VIA package, as can be seen from specified thermal operating area on Page 4. Since the VIA package has a maximum internal temperature rating, it is necessary to estimate this internal temperature based on a system-level thermal solution. To this purpose, it is helpful to simplify the thermal solution into a roughly equivalent circuit where power dissipation is modeled as a current source, isothermal surface temperatures are represented as voltage sources and the thermal resistances are represented as resistors. Figure 18 shows the “thermal circuit” for the VIA package.

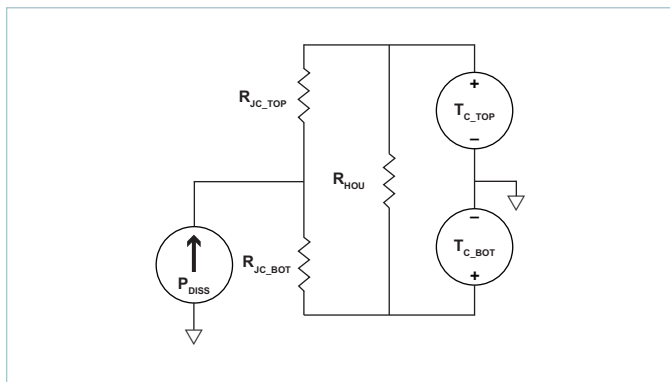


Figure 18 – Double sided cooling VIA thermal model

In this case, the internal power dissipation is  $P_{DISS}$ ,  $R_{JC\_TOP}$  and  $R_{JC\_BOT}$  are thermal resistance characteristics of the VIA package and the top and bottom surface temperatures are represented as  $T_{C\_TOP}$  and  $T_{C\_BOT}$ . It is interesting to notice that the package itself provides a high degree of thermal coupling between the top and bottom case surfaces (represented in the model by the resistor  $R_{HOU}$ ). This feature enables two main options regarding thermal designs:

Single side cooling: the model of Figure 18 can be simplified by calculating the parallel resistor network and using one simple thermal resistance number and the internal power dissipation curves; an example for bottom side cooling only is shown in Figure 19.

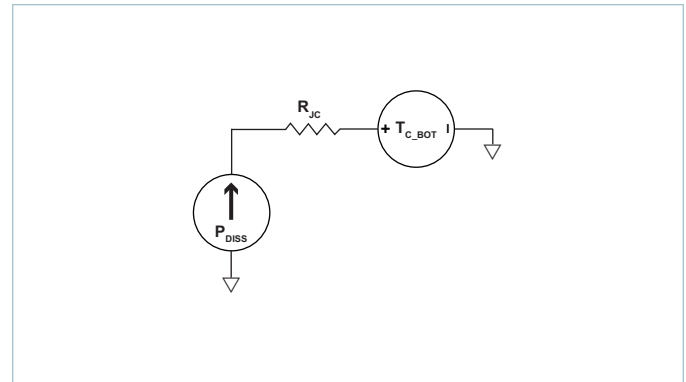


Figure 19 – Single-sided cooling VIA thermal model

In this case,  $R_{JC}$  can be derived as following:

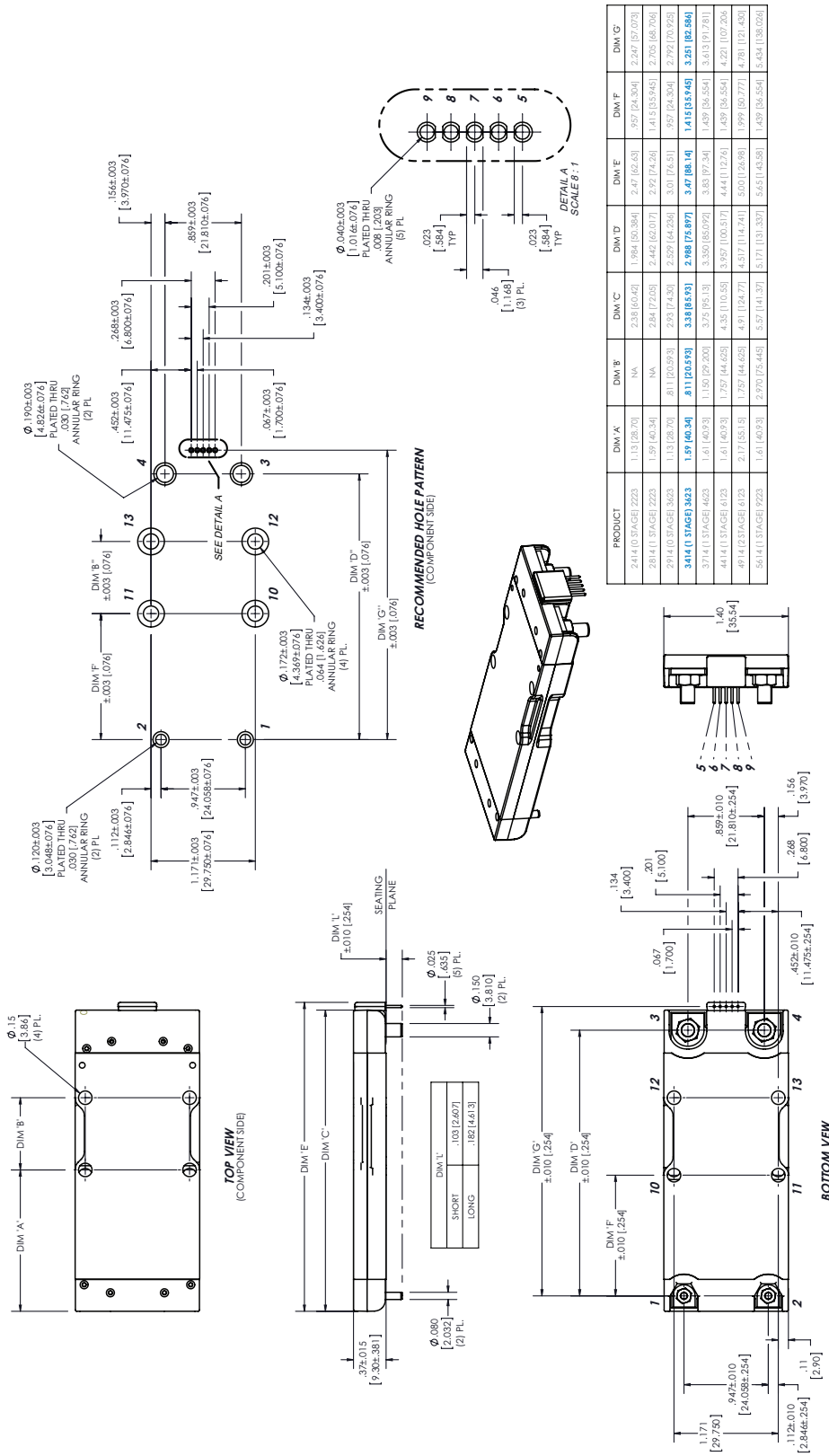
$$R_{JC} = \frac{(R_{JC\_TOP} + R_{HOU}) \cdot R_{JC\_BOT}}{R_{JC\_TOP} + R_{HOU} + R_{JC\_BOT}}$$

Double side cooling: while this option might bring limited advantage to the module internal components (given the surface-to-surface coupling provided), it might be appealing in cases where the external thermal system requires allocating power to two different elements, like for example heatsinks with independent airflows or a combination of chassis/air cooling.

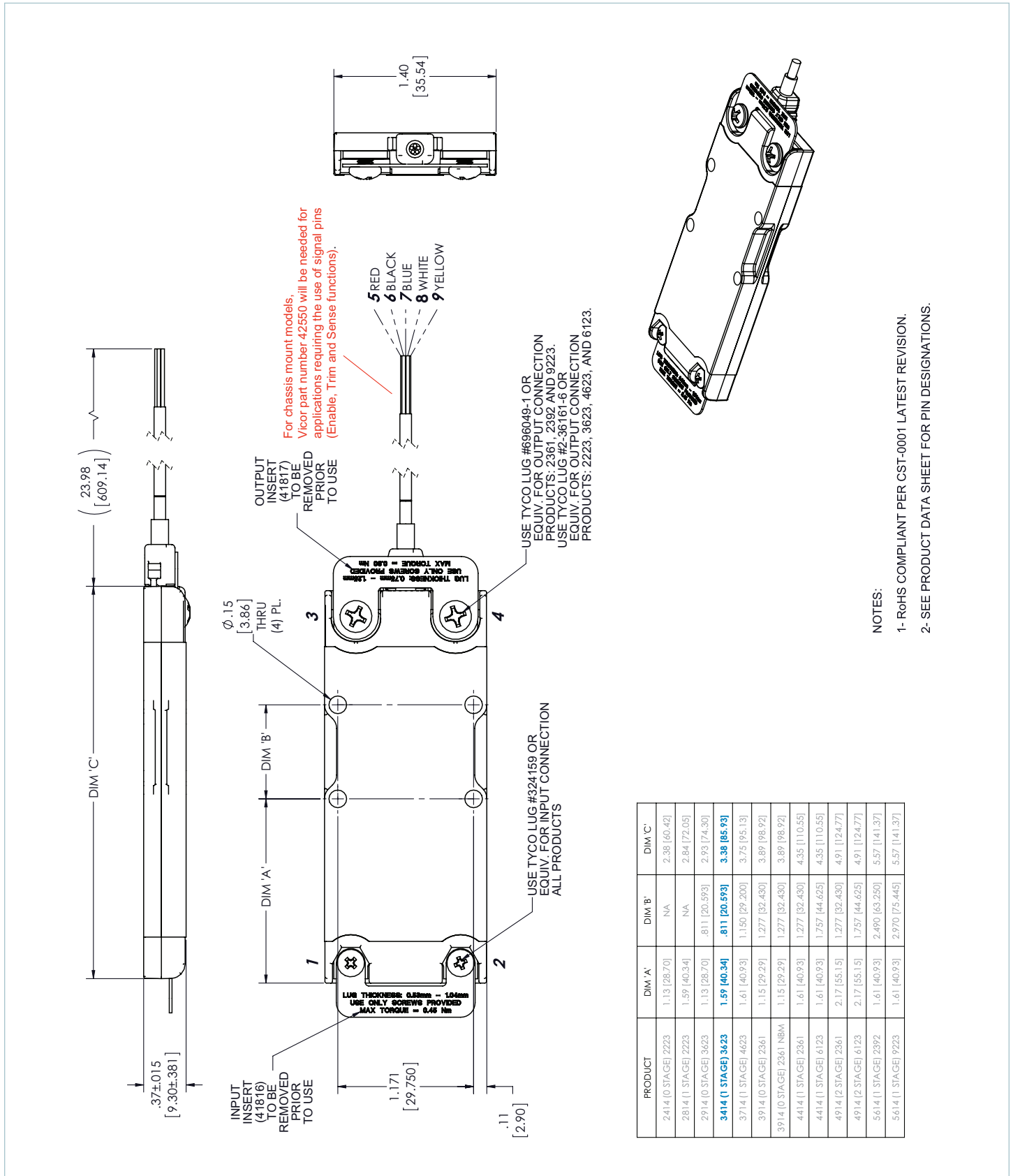
**Grounding Considerations**

The chassis of the VIA package is required to be connected to Protective Earth when installed in the end application and must satisfy the requirements of IEC 60950-1 for Class I products. Both sides of the housing are required to be connected to Protective Earth to satisfy safety and EMI requirements. Protective earthing can be accomplished through dedicated wiring harness (example: ring terminal clamped by mounting screw) or surface contact (example: pressure contact on bare conductive chassis or PCB copper layer with no solder mask).

DCM in a VIA package Module Mechanical Drawing - PCB Mount Type



DCM in a VIA package Module Mechanical Drawing - Chassis Mount Type



PRODUCT	DIM 'A'	DIM 'B'	DIM 'C'
2414 (0 STAGE) 2223	1.13 [28.70]	NA	2.38 [60.42]
2814 (1 STAGE) 2223	1.59 [40.34]	NA	2.84 [72.05]
2914 (0 STAGE) 3623	1.13 [28.70]	.811 [20.593]	2.93 [74.30]
<b>3414 (1 STAGE) 3423</b>	<b>1.59 [40.34]</b>	<b>.811 [20.593]</b>	<b>3.38 [85.93]</b>
3714 (1 STAGE) 4623	1.61 [40.93]	1.150 [29.200]	3.75 [95.13]
3914 (0 STAGE) 2361	1.15 [29.29]	1.277 [32.430]	3.89 [98.92]
3914 (0 STAGE) 2361 NBM	1.15 [29.29]	1.277 [32.430]	3.89 [98.92]
4414 (1 STAGE) 2361	1.61 [40.93]	1.277 [32.430]	4.35 [110.55]
4414 (1 STAGE) 6123	1.61 [40.93]	1.757 [44.625]	4.35 [110.55]
4914 (2 STAGE) 2361	2.17 [55.15]	1.277 [32.430]	4.91 [124.77]
4914 (2 STAGE) 6123	2.17 [55.15]	1.757 [44.625]	4.91 [124.77]
5614 (1 STAGE) 2392	1.61 [40.93]	2.490 [63.250]	5.57 [141.37]
5614 (1 STAGE) 9223	1.61 [40.93]	2.970 [75.445]	5.57 [141.37]

## Revision History

Revision	Date	Description	Page Number(s)
1.0	09/08/15	Intital release	n/a
1.1	09/30/15	Added graphic reference on Pin Configuration	3
1.2	10/01/15	Updated trim resolution value	7
1.3	10/13/15	Updated peak efficiency & case temperature	1
1.4	12/17/15	Added VDDE current consumption spec	7
1.5	02/16/16	Updated pin configuration and pin descriptions	3
1.6	03/31/16	Updated thermal capacitance value	14

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