## 36V, Precision, Low-Noise, Wide-Band Amplifier

## General Description

The MAX9632 is a low-noise, precision, wide-band operational amplifier that can operate in a very wide +4.5 V to +36 V supply voltage range. The IC operates in dual ( $\pm 18 \mathrm{~V}$ ) mode.
The exceptionally fast settling time and low distortion make the IC an excellent solution for precision acquisition systems. The rail-to-rail output swing maximizes the dynamic range when driving high-resolution 24-bit $\Sigma \Delta$ ADCs even with low supply voltages.
The IC achieves 55 MHz of gain-bandwidth product and ultra-low $0.94 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ input voltage noise with only 3.9 mA of quiescent current.
The IC is offered in 8-pin SO and TDFN packages and is rated for operation over the $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ temperature range.

Applications
High-Resolution ADC Drivers
High-Resolution DAC Buffers
Medical Imaging
Low-Noise Signal Processing
Test and Measurement Systems
ATE

Features

- $0.94 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ Ultra-Low Input Voltage Noise
- Very Fast 600ns Settling Time to 16-Bit Accuracy
- THD of -128 dB at 10 kHz
- Low Input Offset Voltage 125 V (max)
- Low Input Offset Temperature Drift $0.5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (max)
- Gain-Bandwidth Product 55MHz
- +4.5V to +36V Wide Supply Range
- Rail-to-Rail Output
- Unity-Gain Stable
- 8-Pin SO and TDFN Packages
- ESD 8kV HBM and 1kV CDM

Ordering Information

| PART | TEMP RANGE | PIN- <br> PACKAGE | TOP <br> MARK |
| :---: | :---: | :--- | :---: |
| MAX9632ASA + | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 8 SO | - |
| MAX9632ATA + | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 8 TDFN-EP* | BML |

+Denotes a lead(Pb)-free/RoHS-compliant package.
*EP $=$ Exposed pad.

## 36V, Precision, Low-Noise, Wide-Band Amplifier

## ABSOLUTE MAXIMUM RATINGS

VCC to VEE $\qquad$ -0.3 V to +40 V All Other Pins..................................(VEE - 0.3V) to (VCC +0.3 V ) Short-Circuit (GND) Duration, OUT...................................... 10s Continuous Input Current (any pin)................................. $\pm 20 \mathrm{~mA}$ Continuous Power Dissipation ( $\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}$ ) (Note 1)
Multilayer SO (derate $7.4 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ )....... .588 mW
Multilayer TDFN (derate $23.8 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ )... 1905 mW
Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## PACKAGE THERMAL CHARACTERISTICS (Note 1)

6 TDFN
Junction-to-Ambient Thermal Resistance $\left(\theta_{\mathrm{JA}}\right) \ldots \ldots . . . . . .2^{\circ} \mathrm{C} / \mathrm{W}$
Junction-to-Case Thermal Resistance $\left(\theta_{\mathrm{JC}}\right) \ldots . . . . . . . . . .8^{\circ} \mathrm{C} / \mathrm{W}$

8 SO
Junction-to-Ambient Thermal Resistance ( $\theta_{\mathrm{JA}}$ ) ........ $136^{\circ} \mathrm{C} / \mathrm{W}$ Junction-to-Case Thermal Resistance ( $\theta_{\mathrm{JC}}$ ) ............... $38^{\circ} \mathrm{C} / \mathrm{W}$

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

## ELECTRICAL CHARACTERISTICS

$\left(V_{C C}=15 \mathrm{~V}, \mathrm{~V}_{\mathrm{EE}}=-15 \mathrm{~V}, R_{L}=10 \mathrm{k} \Omega\right.$ to $\mathrm{V}_{\mathrm{GND}}, \mathrm{V}_{\mathrm{IN}}+=\mathrm{V}_{I N}=\mathrm{V}_{\mathrm{GND}}=0 \mathrm{~V}, \mathrm{~V}_{S H D N}=\mathrm{V}_{\mathrm{GND}}, \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.) (Note 2)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| POWER SUPPLY |  |  |  |  |  |  |
| Supply Voltage Range | VCC | Guaranteed by PSRR | 4.5 |  | 36 | V |
| Supply Current | ICC |  |  | 3.9 | 6.5 | mA |
| Power-Supply Rejection Ratio | PSRR | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 125 | 140 |  | dB |
|  |  | $-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq+125^{\circ} \mathrm{C}$ | 120 |  |  |  |
| SHUTDOWN |  |  |  |  |  |  |
| Shutdown Input Voltage | VSHDN | Device disabled | $\begin{gathered} V_{C C} \\ -0.35 \end{gathered}$ |  | VCC | V |
|  |  | Device enabled | VEE |  | $\begin{aligned} & \hline \text { VCC } \\ & -3.0 \end{aligned}$ |  |
| Shutdown Current | ISHDN | VSHDN $=$ VCC |  | 1 | 15 | $\mu \mathrm{A}$ |
| DC SPECIFICATIONS |  |  |  |  |  |  |
| Input Offset Voltage | Vos | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 30 | 125 | $\mu \mathrm{V}$ |
|  |  | $-40^{\circ} \mathrm{C} \leq \mathrm{T} \leq+125^{\circ} \mathrm{C}$ |  |  | 165 |  |
| Input Offset Voltage Drift | $\pm \Delta \mathrm{VOS}$ | (Note 3) |  | 0.15 | 0.5 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Input Bias Current | IB |  |  | 30 | 180 | nA |
| Input Offset Current | IOS |  |  | 15 | 100 | nA |
| Input Common-Mode Range | VCM | Guaranteed by CMRR | $\begin{gathered} \hline \text { VEE }+ \\ 1.8 \end{gathered}$ |  | $\begin{gathered} \hline \text { VCC - } \\ 1.4 \end{gathered}$ | V |

## 36V, Precision, Low-Noise, Wide-Band Amplifier

## ELECTRICAL CHARACTERISTICS (continued)

$\left(V_{C C}=15 \mathrm{~V}, \mathrm{~V}_{E E}=-15 \mathrm{~V}, R_{L}=10 \mathrm{k} \Omega\right.$ to $\mathrm{V}_{\mathrm{GND}}, \mathrm{V}_{I N+}=\mathrm{V}_{I N}=\mathrm{V}_{\mathrm{GND}}=0 \mathrm{~V}, \mathrm{~V}_{S H D N}=\mathrm{V}_{G N D}, T_{A}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.) (Note 2)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Common-Mode Rejection Ratio | CMRR | $\mathrm{V}_{\text {EE }}+1.8 \mathrm{~V} \leq \mathrm{V}_{\text {CM }} \leq \mathrm{V}_{\text {CC }}-1.4 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | 120 | 135 |  | dB |
|  |  | $\begin{aligned} & V_{E E}+1.8 \mathrm{~V} \leq \mathrm{V}_{C M} \leq V_{C C}-1.4 \mathrm{~V}, \\ & -40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq+125^{\circ} \mathrm{C} \end{aligned}$ |  | 110 |  |  |  |
| Large-Signal Gain | Avol | $\mathrm{V}_{\text {EE }}+0.6 \mathrm{~V} \leq \mathrm{V}_{\text {OUT }} \leq \mathrm{VCC}-0.6 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=600 \Omega$ |  | 125 | 140 |  | dB |
|  |  |  |  | 120 | 135 |  |  |
| Output Voltage Swing | VOH | VCC - Vout | $R \mathrm{~L}=10 \mathrm{k} \Omega$ |  | 50 | 150 | mV |
|  |  |  | $R \mathrm{~L}=600 \Omega$ |  | 150 | 400 |  |
|  | VOL | Vout - Vee | $\mathrm{RL}=10 \mathrm{k} \Omega$ |  | 50 | 150 |  |
|  |  |  | $\mathrm{RL}=600 \Omega$ |  | 150 | 400 |  |
| Short-Circuit Current | ISC | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  |  | 56 |  | mA |
| AC SPECIFICATIONS |  |  |  |  |  |  |  |
| Gain-Bandwidth Product | GBWP |  |  |  | 55 |  | MHz |
| Slew Rate | SR | $0 \leq$ VOUT $\leq 5 \mathrm{~V}$ |  |  | 30 |  | V/us |
| Settling Time | ts | $\begin{aligned} & \text { To } 0.0015 \%, \text { VOUT }=10 \mathrm{VP}-\mathrm{P}, \mathrm{CL}=100 \mathrm{pF}, \\ & \text { AV }=1 \mathrm{~V} / \mathrm{V} \end{aligned}$ |  |  | 600 |  | ns |
| Total Harmonic Distortion | THD | $\begin{aligned} & \mathrm{f}=1 \mathrm{kHz}, \text { VOUT }=3 \mathrm{~V} \text { RMS, } \mathrm{RL}=600 \Omega, \mathrm{AV} \\ & =1 \mathrm{~V} / \mathrm{V} \end{aligned}$ |  |  | -136 |  | dB |
|  |  | $\begin{aligned} & f=10 \mathrm{kHz}, \text { Vout }=3 \mathrm{~V} \text { RMS, } \mathrm{RL}=600 \Omega \text {, } \mathrm{AV} \\ & =1 \mathrm{~V} / \mathrm{V} \end{aligned}$ |  |  | -128 |  |  |
| Input-Voltage Noise Density | eN | $\mathrm{f}=1 \mathrm{kHz}$ |  |  | 0.94 |  | $\mathrm{nV} / \sqrt{\mathrm{Hz}}$ |
| Input Voltage Noise |  | $0.1 \mathrm{~Hz} \leq \mathrm{f} \leq 10 \mathrm{~Hz}$ |  |  | 65 |  | nVP-P |
| Input-Current Noise Density | iN | $\mathrm{f}=1 \mathrm{kHz}$ |  |  | 3.75 |  | $\mathrm{pA} / \sqrt{\mathrm{Hz}}$ |
| Capacitive Loading | CL | No sustained oscillation, $\mathrm{AV}=1 \mathrm{~V} / \mathrm{V}$ |  |  | 350 |  | pF |

Note 2: All devices are $100 \%$ production tested at $\mathrm{TA}=+25^{\circ} \mathrm{C}$. Temperature limits are guaranteed by design.
Note 3: Guaranteed by design.

## 36V, Precision, Low-Noise, Wide-Band Amplifier

 $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)




COMMON-MODE REJECTION RATIO
vs. FREQUENCY


INPUT OFFSET VOLTAGE vs. COMMON-MODE VOLTAGE


POWER-SUPPLY REJECTION RATIO
vs. FREQUENCY



SUPPLY CURRENT
vs. SUPPLY VOLTAGE

INPUT BIAS CURRENT vs. INPUT COMMON-MODE VOLTAGE


OUTPUT VOLTAGE HIGH
vs. OUTPUT CURRENT (SOURCE)


## 36V, Precision, Low-Noise, Wide-Band Amplifier

Typical Operating Characteristics (continued)
$\left(\overline{V_{C C}}=15 \mathrm{~V}, \mathrm{~V}_{E E}=-15 \mathrm{~V}, R_{L}=10 \mathrm{k} \Omega\right.$ to $\mathrm{V}_{\mathrm{GND}}, \mathrm{V}_{I N}+=\mathrm{V}_{I N}=\mathrm{V}_{G N D}=0 \mathrm{~V}, \mathrm{~V}_{S H D N}=\mathrm{V}_{G N D}, T_{A}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)


## 36V, Precision, Low-Noise, Wide-Band Amplifier

## Typical Operating Characteristics (continued)

$\left(V_{C C}=15 \mathrm{~V}, V_{E E}=-15 \mathrm{~V}, R_{L}=10 \mathrm{k} \Omega\right.$ to $V_{G N D}, V_{I N}=V_{I N}=V_{G N D}=0 \mathrm{~V}, V_{S H D N}=V_{G N D}, T_{A}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)


RECOVERY FROM SHUTDOWN



TOTAL HARMONIC DISTORTION
vs. FREQUENCY


TOTAL HARMONIC DISTORTION
vs. OUTPUT VOLTAGE


# 36V, Precision, Low-Noise, Wide-Band Amplifier 

Pin Configuration


Pin Description

| PIN | NAME |  |
| :---: | :---: | :--- |
| 1,5 | N.C. | Not Connected |
| 2 | IN- | Negative Input |
| 3 | IN+ | Positive Input |
| 4 | VEE | Negative Supply Voltage |
| 6 | OUT | Output |
| 7 | VCC | Positive Supply Voltage |
| 8 | SHDN | Active-High Shutdown |
| - | EP | Exposed Pad (TDFN Only). Connect to a large VEE plane to maximize thermal performance. Not <br> intended as an electrical connection point. |

# 36V, Precision, Low-Noise, Wide-Band Amplifier 

## Detailed Description

The MAX9632 is designed in a new 36V, high-speed complementary BiCMOS process that is optimized for excellent AC dynamic performance combined with highvoltage operation.
The IC offers precision, high-bandwidth, ultra-low noise and exceptional distortion performance.
The IC is unity-gain stable and operates either with single-supply voltage up to 36 V or with dual supplies up to $\pm 18 \mathrm{~V}$.

## Applications Information

## Operating Supply Voltage

The IC can operate with dual supplies from $\pm 2.25 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ or with a single supply from +4.5 V to +36 V with respect to ground. Even though the IC supports highvoltage operation with excellent performance, the device can also operate in very popular applications at 5 V .

## Low Noise and Low Distortion

The IC is designed for extremely low-noise applications such as professional audio equipment, very high performance instrumentations, automated test equipment, and medical imaging. The low noise, combined with fast settling time, makes it ideal to drive high-resolution sigmadelta or SARs analog-to-digital converters.
The IC is also designed for ultra-low-distortion performance. THD specifications in the Electrical Characteristics table and Typical Operating Characteristics are calculated up to the fifth harmonic. Even when driving highvoltage swing up to 10VP-p, the IC maintains excellent low distortion operation over and above 100 kHz of bandwidth.

Rail-to-Rail Output Stage
The output stage swings to within 50 mV (typ) of either power-supply rail with a $10 \mathrm{k} \Omega$ load and provides a 55 MHz GBW with a $30 \mathrm{~V} / \mathrm{s}$ slew rate. The device is unity-gain stable and can drive a 100 pF capacitive load without compromising stability. Stability with higher capacitive loads can be improved by adding an isolation resistor in series with the op-amp output. This resistor improves the circuit's phase margin by isolating the load capacitor from the amplifier's output. The Typical Operating Characteristics show a profile of the isolation resistor and capacitive load values that maintain the device into the stable region.


Figure 1. Input Protection Circuit

## Input Differential Voltage Protection

 During normal op-amp operation, the inverting and noninverting inputs of the IC are at essentially the same voltage. However, either due to fast input voltage transients or other fault conditions, these inputs can be forced to be at two different voltages.Internal back-to-back diodes protect the inputs from an excessive differential voltage (Figure 1). Therefore, $\operatorname{IN+}$ and IN - can be any voltage within the range shown in the Absolute Maximum Ratings section. Note the protection time is still dependent on the package thermal limits.
If the input signal is fast enough to create the internal diodes' forward bias condition, the input signal current must be limited to 20 mA or less. If the input signal current is not inherently limited, an input series resistor can be used to limit the signal input current. Care should be taken in choosing the input series resistor value, since it degrades the low-noise performance of the device.

## Shutdown

The shutdown is referenced to the positive supply. See the Electrical Characteristics table for the proper levels of functionality. A high level (above Vcc -0.35 V ) disables the op amp and puts the output into a high-impedance state. A low level (below $\mathrm{V}_{\mathrm{cc}}-3 \mathrm{~V}$ ) enables the device. As an example, if the op amp is powered with dual supplies of $\pm 15 \mathrm{~V}$, the device is enabled when shutdown is at or below 12 V . The device is disabled when shutdown is at or above 14.65 V . If the op amp is powered with a single supply of 36 V , the device is enabled when shutdown is at or below 33 V . The device is disabled when shutdown is at or above 35.65 V . This input must be connected to a valid high or low voltage and should not be left disconnected.

Power Supplies and Layout The MAX9632 can operate with dual supplies from $\pm 2.25 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ or with a single supply from +4.5 V to +36 V with respect to ground. When used with dual supplies, bypass both VCC and VEE with their own $0.1 \mu \mathrm{~F}$ capacitor to ground. When used with a single supply, bypass $V_{C C}$ with a $0.1 \mu \mathrm{~F}$ capacitor to ground.

## 36V, Precision, Low-Noise, Wide-Band Amplifier

Careful layout technique helps optimize performance by decreasing the amount of stray capacitance at the op amp's inputs and outputs. To decrease stray capacitance, minimize trace lengths by placing external components close to the op amp's pins.
For high-frequency designs, ground vias are critical to provide a ground return path for high-frequency signals and should be placed near the decoupling capacitors. Signal routing should be short and direct to avoid parasitic effects. Avoid using right angle connectors since they may introduce a capacitive discontinuity and ultimately limit the frequency response.

## Electrostatic Discharge (ESD)

The IC has built-in circuits to protect it from ESD events. An ESD event produces a short, high-voltage pulse that is transformed into a short current pulse once it discharges through the device. The built-in protection circuit provides a current path around the op amp that prevents it from being damaged. The energy absorbed by the protection circuit is dissipated as heat.
ESD protection is guaranteed up to $\pm 8 \mathrm{kV}$ with the Human Body Model (HBM). The Human Body Model simulates the ESD phenomenon wherein a charged body directly transfers its accumulated electrostatic charge to the ESD-sensitive device. A common example of this phenomenon is when a person accumulates static charge by walking across a carpet and then transfers all of the charge to an ESD-sensitive device by touching it.
Not all ESD events involve the transfer of charge into the device. ESD from a charged device to another body is also a common form of ESD.

If a charged device comes into contact with another conductive body that is at a lower potential, it discharges into that body. Such an ESD event is known as Charged Device Model (CDM) ESD, which can be even more destructive than HBM ESD (despite its shorter pulse duration) because of its high current. The IC guarantees CDM ESD protection up to $\pm 1 \mathrm{kV}$.

Driving High-Resolution Sigma-Delta ADCs
The MAX9632's excellent AC specifications and 55 MHz bandwidth are a good fit for driving high-speed, precision delta-sigma ADCs. These ADCs require an ultra-low noise op amp to achieve signal-to-noise ratios (SNR) better than 100 dB . The MAX11040 is a 24-bit, 4-channel, simultane-ous-sampling ADC with 117 dB SNR at 1 ksps and 106 dB at 16ksps. The MAX11040 measures analog inputs up to $\pm 2.2 \mathrm{~V}$. Sampling up to 64 ksps , the MAX11040 achieves better than -94dB THD and 94dB SFDR.

The MAX11040 measures four differential inputs simultaneously, outputting the data through an SPITM interface to allow daisy-chaining the data outputs and inputs together. Therefore, up to eight MAX11040 devices can be placed in parallel to measure up to 32 inputs simultaneously. This is ideal for 3-phase power monitoring that requires multiple current and voltage readings and very wide dynamic range.
The Typical Application Circuit shows an example of the MAX9632 driving the MAX11040.

Chip Information
PROCESS: BiCMOS

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## Package Information

For the latest package outline information and land patterns (footprints), go to www.maxim-ic.com/packages. Note that a "+", "\#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | OUTLINE NO. | LAND PATTERN NO. |
| :---: | :---: | :---: | :---: |
| 8 SO | $\mathrm{S} 8+2$ | $\underline{21-0041}$ | $\underline{90-0096}$ |
| 8 TDFN-EP | $\mathrm{T} 833+3$ | $\underline{\mathbf{2 1 - 0 1 3 7}}$ | $\underline{90-0060}$ |



NDTES:

1. ALL dimensions are in millimeters unless atherwise specified.
2. MATERIAL MUST COMPLY WITH BANNED AND RESTRICTED SUBSTANCES SPEC \# 10-0131.
3. DIMENSİNS D AND E da NDT INCLUDE MDLD PROTRUSIUN

4LLDWABLE MDLD PRDTRUSIDN IS 0.15 MM (.006") PER SIDE
LEADS TI BE CEPLANAR WITHIN 0.10 mm (.004").
5. MEETS JEDEC MSO12
6. ALL dimensians apply ta bath leaded (-) and Pbfree (+) PkG. Cades.
-DRAWING NOT TO SCALE-


| VARIATIDN B |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| SYMBCL | INCHES |  | MM |  |
|  | MIN. | MAX. | MIN. | MAX. |
| D | .337 | .344 | 8.55 | 8.75 |
| N | 14 |  |  |  |
| MS012 | AB |  |  |  |
| PKG. <br> CDDE | S14-1, S14-4, S14-5, <br> S14-6; S14M-4, S14M-5, <br> S14M-6, S14M-7 |  |  |  |


| VARIATIUN C |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| SYMBCL | INCHES |  | MM |  |
|  | MIN. | MAX. | MIN. | MAX. |
| D | .386 | .394 | 9.80 | 10.00 |
| N | 16 |  |  |  |
| MS012 | AC |  |  |  |
| PKG. <br> CDDE | S16-1, S16-3, S16-5, S16-6, <br> S16-8, S16-7F, S16-9F, <br> S16-10F; S16M-3, S16M-6 |  |  |  |


| \|41 |  |  |
| :---: | :---: | :---: |
| TTLE:PACKAGE QUTLINE,8L, 14L, 16L SDIC 150 INCH |  |  |
|  |  |  |
|  |  |  |
| $\stackrel{\text { APPROVAL }}{ }$ | [DOCUMENT CONTROL NO. | $\stackrel{\text { REV. }}{\text { C }}$ |

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# 36V, Precision, Low-Noise, Wide-Band Amplifier 

Package Information (continued)
For the latest package outline information and land patterns (footprints), go to www.maxim-ic.com/packages. Note that a "+", "\#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| COMMON DIMENSIONS |  |  |
| :---: | :---: | :---: |
| SYMBOL | MIN. | MAX. |
| A | 0.70 | 0.80 |
| D | 2.90 | 3.10 |
| E | 2.90 | 3.10 |
| A1 | 0.00 | 0.05 |
| L | 0.20 | 0.40 |
| k | 0.25 MIN.$$ |  |
| A2 | 0.20 REF. |  |


| PACKAGE VARIATIONS |  |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PKG. CODE | N | D 2 | E 2 | e | JEDEC SPEC | b | $[(\mathrm{N} / 2)-1] \times \mathrm{e}$ |
| T633-2 | 6 | $1.50 \pm 0.10$ | $2.30 \pm 0.10$ | 0.95 BSC | MO229 / WEEA | $0.40 \pm 0.05$ | 1.90 REF |
| T833-2 | 8 | $1.50 \pm 0.10$ | $2.30 \pm 0.10$ | 0.65 BSC | MO229 / WEEC | $0.30 \pm 0.05$ | 1.95 REF |
| T833-3 | 8 | $1.50 \pm 0.10$ | $2.30 \pm 0.10$ | 0.65 BSC | MO229 / WEEC | $0.30 \pm 0.05$ | 1.95 REF |
| T1033-1 | 10 | $1.50 \pm 0.10$ | $2.30 \pm 0.10$ | 0.50 BSC | MO229 / WEED-3 | $0.25 \pm 0.05$ | 2.00 REF |
| T1033MK-1 | 10 | $1.50 \pm 0.10$ | $2.30 \pm 0.10$ | 0.50 BSC | MO229 / WEED-3 | $0.25 \pm 0.05$ | 2.00 REF |
| T1033-2 | 10 | $1.50 \pm 0.10$ | $2.30 \pm 0.10$ | 0.50 BSC | MO229 / WEED-3 | $0.25 \pm 0.05$ | 2.00 REF |
| T1433-1 | 14 | $1.70 \pm 0.10$ | $2.30 \pm 0.10$ | 0.40 BSC | ---- | $0.20 \pm 0.05$ | 2.40 REF |
| T1433-2 | 14 | $1.70 \pm 0.10$ | $2.30 \pm 0.10$ | 0.40 BSC | ---- | $0.20 \pm 0.05$ | 2.40 REF |
| T1433-3F | 14 | $1.70 \pm 0.10$ | $2.30 \pm 0.10$ | 0.40 BSC | ---- | $0.20 \pm 0.05$ | 2.40 REF |

NOTES:

1. ALL DIMENSIONS ARE IN mm. ANGLES IN DEGREES.
2. COPLANARITY SHALL NOT EXCEED 0.08 mm .
3. WARPAGE SHALL NOT EXCEED 0.10 mm .
4. PACKAGE LENGTH/PACKAGE WIDTH ARE CONSIDERED AS SPECIAL CHARACTERISTIC(S).
5. DRAWING CONFORMS TO JEDEC MO229, EXCEPT DIMENSIONS "D2" AND "E2", AND T1433-1 \& T1433-2.
6. "N" IS THE TOTAL NUMBER OF LEADS.
7. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.
8. MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
9. ALL DIMENSIONS APPLY TO BOTH LEADED ( - ) AND PbFREE (+) PKg. CODES.
-DRAWING NOT TO SCALE-


## 36V, Precision, Low-Noise, Wide-Band Amplifier

| REVISION <br> NUMBER | REVISION <br> DATE | DESCRIPTION | PAGES <br> CHANGED |
| :---: | :---: | :--- | :---: |
| 0 | $10 / 10$ | Initial release | - |
| 1 | $4 / 11$ | Updated short-circuit current spec | 3 |
| 2 | $8 / 11$ | Updated TDFN land pattern number | 11 |

