

**10-bit high-speed analog-to-digital converter****TDA8760****FEATURES**

- 10-bit resolution
- Sampling rate up to 50 MHz
- Total harmonic distortion (THD): -65 dB at 4.43 MHz full scale and a 40 MHz clock frequency
- High signal-to-noise ratio over a large analog input frequency range (8.8 effective bits at 10 MHz full-scale input at a 40 MHz clock frequency)
- +5 V power supplies
- Binary or two's complement 3-state TTL outputs
- In-range 3-state TTL output
- TTL compatible digital inputs
- LOW-level AC clock input signal allowed
- Power dissipation 850 mW (typical)
- Low analog input capacitance (typ. 4.5 pF), no buffer amplifier required
- No external sample-and-hold circuit required
- Analog Input; single or differential
- External amplitude range control
- Voltage controlled regulator included.

**APPLICATIONS**

- High-speed analog-to-digital conversion for
  - Video signal digitizing
  - High Definition TV (HDTV)
  - Digital video broadcasting (satellite and cable)
  - Transient signal analysis
  - High energy physics research
  - Sigma-delta (SD) modulators
  - Medical imaging
  - Radar pulse digitizing.

**GENERAL DESCRIPTION**

The TDA8760 is a monolithic bipolar 10-bit analog-to-digital converter (ADC) for video or other applications. It converts the analog input signal into 10-bit binary coded digital words at a maximum sampling rate of 50 MHz. All digital inputs and outputs are TTL compatible. However, a sine-wave clock input signal is allowed.

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## QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{CCA}$	analog supply voltage		4.75	5.0	5.25	V
$V_{CCD}$	digital supply voltage		4.75	5.0	5.25	V
$V_{CCO}$	output supply voltage		4.75	5.0	5.25	V
$I_{CCA}$	analog supply current		–	95	100	mA
$I_{CCD}$	digital supply current		–	40	45	mA
$I_{CCO}$	output supply current		–	35	40	mA
ILE	DC integral linearity error	$f_{clk} = 4 \text{ MHz}$	–	$\pm 1.0$	$\pm 2.0$	LSB
DLE	DC differential linearity error	$f_{clk} = 4 \text{ MHz}$	–	$\pm 0.6$	$\pm 1.0$	LSB
AILE	AC integral linearity error	$f_{clk} = 40 \text{ MHz};$ $f_i = 4.43 \text{ MHz}$	–	$\pm 1.2$	$\pm 2.0$	LSB
$f_{clk(max)}$	maximum clock frequency					
	TDA8760K/2		20	–	–	MHz
	TDA8760K/4		40	–	–	MHz
	TDA8760K/5		50	–	–	MHz
$P_{tot}$	total power dissipation		–	850	970	mW
$T_{amb}$	operating ambient temperature		0	–	+70	°C

## ORDERING INFORMATION

EXTENDED TYPE NUMBER	PACKAGE				
	PINS	PIN POSITION	MATERIAL	CODE	SAMPLING FREQUENCY (MHz)
TDA8760K/2	44	PLCC	plastic	SOT187	20
TDA8760K/4	44	PLCC	plastic	SOT187	40
TDA8760K/5	44	PLCC	plastic	SOT187	50

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## BLOCK DIAGRAM

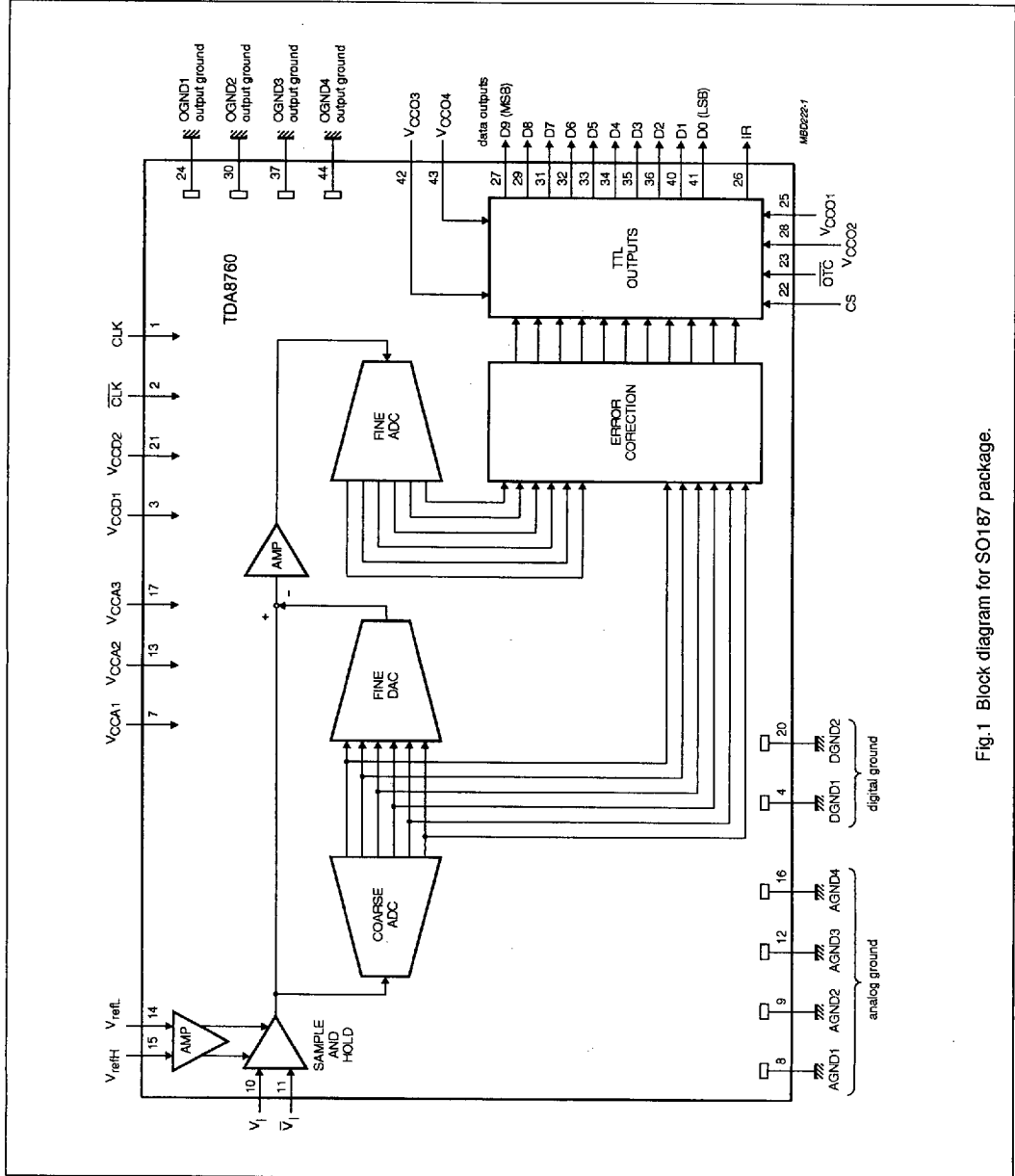


Fig.1 Block diagram for SO187 package.

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## PINNING

SYMBOL	PIN	DESCRIPTION
CLK	1	clock input
$\overline{\text{CLK}}$	2	complementary clock input
V <sub>CCD1</sub>	3	digital supply voltage (+5 V)
DGND1	4	digital ground
n.c.	5	not connected
n.c.	6	not connected
V <sub>CCA1</sub>	7	analog supply voltage (+5 V)
AGND1	8	analog ground
AGND2	9	analog ground
V <sub>I</sub>	10	analog input voltage
$\overline{V}_I$	11	complementary analog input voltage
AGND3	12	analog ground
V <sub>CCA2</sub>	13	analog supply voltage (+5 V)
V <sub>refL</sub>	14	reference voltage LOW
V <sub>refH</sub>	15	reference voltage HIGH
AGND4	16	analog ground
V <sub>CCA3</sub>	17	analog supply voltage (+5 V)
n.c.	18	not connected
n.c.	19	not connected
DGND2	20	digital ground
V <sub>CCD2</sub>	21	digital supply voltage (+5 V)
CS	22	chip select input (TTL level input; active HIGH)
$\overline{\text{OTC}}$	23	output two's complement
OGND1	24	output ground
V <sub>CCO1</sub>	25	output supply voltage (+5 V)
IR	26	in-range output
D9	27	data output, bit 9 (MSB)
V <sub>CCO2</sub>	28	output supply voltage (+5 V)
D8	29	data output, bit 8
OGND2	30	output ground
D7	31	data output, bit 7
D6	32	data output, bit 6
D5	33	data output, bit 5
D4	34	data output, bit 4
D3	35	data output, bit 3
D2	36	data output, bit 2
OGND3	37	output ground
n.c.	38	not connected
n.c.	39	not connected
D1	40	data output, bit 1

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SYMBOL	PIN	DESCRIPTION
D0	41	data output, bit 0 (LSB)
V <sub>CC03</sub>	42	output supply voltage (+5 V)
V <sub>CC04</sub>	43	output supply voltage (+5 V)
OGND4	44	output ground

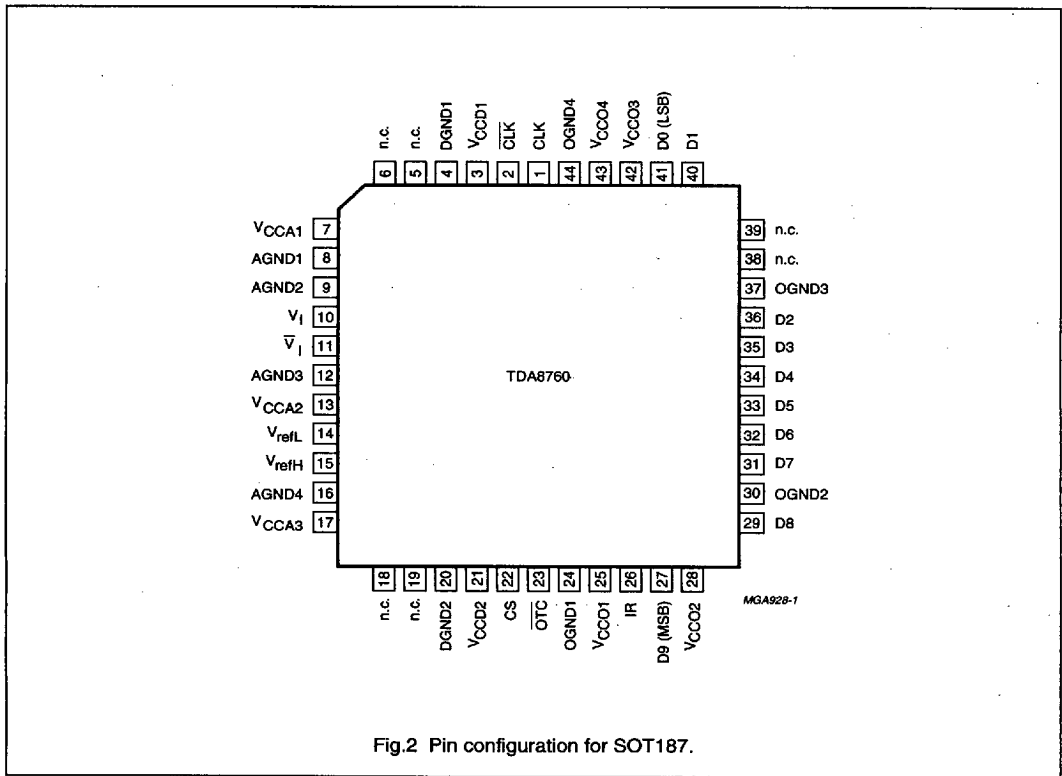


Fig.2 Pin configuration for SOT187.

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## LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_{CCA}$	analog supply voltage		-0.3	+7.0	V
$V_{CCD}$	digital supply voltage		-0.3	+7.0	V
$V_{CCO}$	output supply voltage		-0.3	+7.0	V
$\Delta V_{CC1}$	supply voltage difference between $V_{CCA}$ and $V_{CCD}$		-0.5	+0.5	V
$\Delta V_{CC2}$	supply voltage difference between $V_{CCO}$ and $V_{CCB}$		-0.5	+0.5	V
$\Delta V_{CC3}$	supply voltage difference between $V_{CCA}$ and $V_{CCO}$		-0.5	0.5	V
$V_I$	input voltage	referenced to AGND	0.3	$V_{CCA}$	V
$V_{I(p-p)}$	input voltage for differential clock drive (peak-to-peak value)		-	$V_{CCD}$	V
$I_O$	output current		-	10	mA
$T_{stg}$	storage temperature		-55	+150	°C
$T_{amb}$	operating ambient temperature		0	+70	°C
$T_j$	junction temperature		-	+150	°C

## HANDLING

Inputs and outputs are protected against electrostatic discharges in normal handling. However, to be totally safe, it is desirable to take normal precautions appropriate to handling integrated circuits.

## THERMAL RESISTANCE

SYMBOL	PARAMETER	THERMAL RESISTANCE
$R_{th\ j-a}$	from junction to ambient in free air TDA8760K/5; TDA8760K/4 TDA8760K/2	35 K/W 46 K/W

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**CHARACTERISTICS**

$V_{CCA} = V_{CCD} = V_{CCO} = 4.75$  to  $5.25$  V; AGND and DGND shorted together;

$V_{CCA} - V_{CCD} = V_{CCO} - V_{CCD} = V_{CCA} - V_{CCO} = -0.25$  to  $+0.25$  V;  $T_{amb} = 0$  to  $+70$  °C; unless otherwise specified.

Typical values measured at  $V_{CCA} = V_{CCD} = V_{CCO} = 5$  V;  $T_{amb} = 25$  °C.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Supply</b>						
$V_{CCA}$	analog supply voltage		4.75	5.0	5.25	V
$V_{CCD}$	digital supply voltage		4.75	5.0	5.25	V
$V_{CCO}$	output supply voltage		4.75	5.0	5.25	V
$I_{CCA}$	analog supply current		–	95	100	mA
$I_{CCD}$	digital supply current		–	40	45	mA
$I_{CCO}$	output supply current	all outputs LOW	–	35	40	mA
<b>Inputs</b>						
CLK and $\overline{\text{CLK}}$ (REFERENCED TO DGND); NOTE 1						
$V_{IL}$	LOW level input voltage		0	–	0.8	V
$V_{IH}$	HIGH level input voltage		2.0	–	$V_{CCD}$	V
$I_{IL}$	LOW level input current	$V_{\text{clk}}$ or $V_{\overline{\text{clk}}} = 0.4$ V	–400	–	–	mA
$I_{IH}$	HIGH level input current	$V_{\text{clk}}$ or $V_{\overline{\text{clk}}} = 2.0$ V	–	–	100	mA
		$V_{\text{clk}}$ or $V_{\overline{\text{clk}}} = V_{CCD}$	–	–	300	mA
$Z_I$	input impedance	$f_{\text{clk}} = 40$ MHz	–	2	–	k $\Omega$
$C_I$	input capacitance	$f_{\text{clk}} = 40$ MHz	–	4.5	–	pF
$\Delta V_{\text{clk}}$	AC input voltage for switching ( $V_{\text{clk}} - V_{\overline{\text{clk}}}$ )	DC level = 1.5 V	0.5	–	2.0	V
		DC level = 2.5 V	1.5	–	5.0	V
<b>OTC and CS (REFERENCED TO DGND); SEE TABLE 3</b>						
$V_{IL}$	LOW level input voltage		0	–	0.8	V
$V_{IH}$	HIGH level input voltage		2.0	–	$V_{CCD}$	V
$I_{IL}$	LOW level input current	$V_{IL} = 0.8$ V	–400	–	–	$\mu$ A
$I_{IH}$	HIGH level input current	$V_{IH} = 2.0$ V	–	–	20	$\mu$ A

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b><math>V_I</math> AND <math>\bar{V}_I</math> (REFERENCED TO AGND; SEE ALSO TABLES 1 AND 2)</b>						
$I_{IL}$	LOW level input current	$V_{refH} - V_{refL} = 1.5 \text{ V}$	–	7	–	$\mu\text{A}$
$I_{IH}$	HIGH level input current	$V_{refH} - V_{refL} = 1.5 \text{ V}$	–	22	–	$\mu\text{A}$
$Z_I$	input impedance	$f_i = 4.43 \text{ MHz}$	–	2	–	$\text{k}\Omega$
$C_I$	input capacitance	$f_i = 4.43 \text{ MHz}$	–	4.5	–	$\text{pF}$
$V_{\text{offset}(d)}$	input offset voltage	differential mode; $V_I = \bar{V}_I$ ; output code 511; Table 1 $V_{CCA} = 5 \text{ V}$ $V_{CCA} = 4.75 \text{ V}$ $V_{CCA} = 5.25 \text{ V}$	3.3 3.2 3.3	3.4 – –	3.6 3.45 3.8	V V V
$V_{\text{offset}(s)}$	input offset voltage	single mode; $V_I = V_{\text{offset}(s)}$ ; output code 511; Table 2 $V_{CCA} = 5 \text{ V}$ $V_{CCA} = 4.75 \text{ V}$ $V_{CCA} = 5.25 \text{ V}$	3.6 tbf tbf	3.7 – –	3.8 tbf tbf	V V V
<b>Voltage controlled regulator inputs <math>V_{refH}</math> and <math>V_{refL}</math> (referenced to AGND); differential input</b>						
$V_{refH}$	reference voltage HIGH		4.0	4.5	$V_{CCA}$	V
$V_{refL}$	reference voltage LOW		2.5	3.0	3.5	V
$V_{I(p-p)}$	input voltage amplitude (peak-to-peak value)		1.4	1.5	1.6	V
$I_{refH}$	input current at $V_{refH}$		–	10	–	$\mu\text{A}$
$I_{refL}$	input current at $V_{refL}$		–	10	–	$\mu\text{A}$
<b>Voltage controlled regulator inputs <math>V_{refH}</math> and <math>V_{refL}</math> (referenced to AGND); single input</b>						
$V_{refH}$	reference voltage HIGH		4.0	4.4	$V_{CCA}$	V
$V_{refL}$	reference voltage LOW		2.5	3.0	3.5	V
$V_{I(p-p)}$	input voltage amplitude (peak-to-peak value)		1.3	1.4	1.5	V
$I_{refH}$	input current at $V_{refH}$		–	10	–	$\mu\text{A}$
$I_{refL}$	input current at $V_{refL}$		–	10	–	$\mu\text{A}$
<b>Outputs (referenced to DGND)</b>						
<b>DIGITAL OUTPUTS D9 TO D0 AND IR (REFERENCED TO DGND)</b>						
$V_{OL}$	LOW level output voltage	$I_O = 2 \text{ mA}$	0	–	0.4	V
$V_{OH}$	HIGH level output voltage	$I_O = -0.4 \text{ mA}$	2.4	–	$V_{CCD}$	V
$I_O$	output current in 3-state mode	$0.4 \text{ V} < V_O < V_{CCD}$	–20	–	+20	$\mu\text{A}$



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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Switching characteristics</b>						
CLOCK FREQUENCY $f_{clk}$ (NOTE 1; SEE FIG.3)						
$f_{clk(min)}$	minimum clock frequency		–	–	1	MHz
$f_{clk(max)}$	maximum clock frequency					
	TDA8760K/5		50	–	–	MHz
	TDA8760K/4		40	–	–	MHz
	TDA8760K/2		20	–	–	MHz
$t_{CPH}$	clock pulse width HIGH	note 7	10	–	–	ns
$t_{CPL}$	clock pulse width LOW		8	–	–	ns
<b>Analog signal processing in differential input mode; see Table 1; 50% clock duty factor;</b> $V_{I(p-p)} = V_{refH} - V_{refL} = 1.5 V$						
LINEARITY						
ILE	DC integral linearity error	$f_{clk} = 4 MHz$	–	$\pm 1.0$	$\pm 2.0$	LSB
DLE	DC differential linearity error	$f_{clk} = 4 MHz$	–	$\pm 0.6$	$\pm 1.0$	LSB
AILE	AC integral linearity error	note 3	–	$\pm 1.2$	$\pm 2.0$	LSB
BANDWIDTH ( $f_{clk} = 50 MHz$ ); NOTE 9						
B	Analog bandwidth	-1 dB	–	140	–	MHz
		-3 dB	–	220	–	MHz
HARMONICS ( $f_{clk} = 40 MHz$ ); SEE FIGS 6, 8, 9 AND 10						
$f_1$	fundamental harmonics (full scale)	$f_1 = 4.43 MHz$	–	–	0	dB
$f_{all}$	harmonics (full scale); all components	$f_1 = 4.43 MHz$	–	–70	–63	dB
			–	–70	–63	dB
			–	–70	–63	dB
THD <sub>d</sub>	total harmonic distortion	$f_1 = 4.43 MHz$ ; note 2	–	–65	–60	dB
SIGNAL-TO-NOISE RATIO; NOTES 4 AND 5; SEE FIGS 6, 8, 9 AND 10						
SNR	signal-to-noise ratio	without harmonics; $f_{clk} = 40 MHz$ ; $f_1 = 4.43 MHz$	54	56	–	dB
EFFECTIVE BITS; NOTES 4 AND 5; SEE FIGS 6, 8, 9 AND 10						
EB	effective bits TDA8760K/2 ( $f_{clk} = 20 MHz$ )	$f_1 = 4.43 MHz$	–	8.90	–	bits
		$f_1 = 7.5 MHz$	–	8.70	–	bits
	effective bits TDA8760K/4 ( $f_{clk} = 40 MHz$ )	$f_1 = 4.43 MHz$	–	8.80	–	bits
		$f_1 = 10 MHz$	–	8.80	–	bits
		$f_1 = 15 MHz$	–	8.70	–	bits
	effective bits TDA8760K/5 ( $f_{clk} = 50 MHz$ )	$f_1 = 4.43 MHz$	–	8.70	–	bits
		$f_1 = 10 MHz$	–	8.65	–	bits
		$f_1 = 15 MHz$	–	8.60	–	bits
		$f_1 = 20 MHz$	–	8.20	–	bits

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT	
<b>TWO-TONE</b>							
Two-tone	two-tone intermodulation rejection	$f_{\text{clk}} = 40 \text{ MHz}$ ; note 8	-	-65	-	dB	
<b>BIT ERROR RATE</b>							
BER	bit error rate	$f_{\text{clk}} = 40 \text{ MHz}$ ; $f_1 = 4.43 \text{ MHz}$ ; $V_1 = \pm 16 \text{ LSB}$ at code 512	-	$2 \times 10^{-12}$	-	times/ samples	
<b>DIFFERENTIAL GAIN; SEE FIG.5</b>							
$G_{\text{diff}}$	differential gain	$f_{\text{clk}} = 20 \text{ MHz}$ ; $f_1 = 4.43 \text{ MHz}$	-	0.5	tbf	%	
		$f_{\text{clk}} = 40 \text{ MHz}$ ; $f_1 = 4.43 \text{ MHz}$	-	1.0	tbf	%	
<b>DIFFERENTIAL PHASE</b>							
$\Phi_{\text{diff}}$	differential phase	$f_{\text{clk}} = 40 \text{ MHz}$ ; $f_1 = 4.43 \text{ MHz}$	-	0.1	0.2	deg	
<b>Analog signal processing in single input mode; see Table 2; 50% clock duty factor; <math>V_{\text{I(P-P)}} = V_{\text{refH}} - V_{\text{refL}} = 1.4 \text{ V}</math></b>							
<b>LINEARITY</b>							
ILE	DC integral linearity error	$f_{\text{clk}} = 4 \text{ MHz}$	-	$\pm 1.0$	$\pm 2.0$	LSB	
DLE	DC differential linearity error	$f_{\text{clk}} = 4 \text{ MHz}$	-	$\pm 0.6$	$\pm 1.0$	LSB	
AILE	AC integral linearity error	note 3	-	$\pm 1.2$	$\pm 2.0$	LSB	
<b>BANDWIDTH (<math>f_{\text{clk}} = 50 \text{ MHz}</math>); NOTE 9</b>							
B	Analog bandwidth	-1 dB	-	140	-	MHz	
		-3 dB	-	220	-	MHz	
<b>HARMONICS (<math>f_{\text{clk}} = 40 \text{ MHz}</math>); SEE FIG.7</b>							
$f_1$	fundamental harmonics (full scale)	$f_1 = 4.43 \text{ MHz}$	-	-	0	dB	
$f_{\text{all}}$	harmonics (full scale); all components	$f_1 = 4.43 \text{ MHz}$	-	-	-	-	
			second harmonics	-	-61	-	dB
			third harmonics	-	-62	-	dB
$\text{THD}_s$	total harmonic distortion	$f_1 = 4.43 \text{ MHz}$ ; note 2	-	-59	-	dB	
<b>SIGNAL-TO-NOISE RATIO; NOTES 4 AND 5; SEE FIG.7</b>							
SNR	signal-to-noise ratio	without harmonics; $f_{\text{clk}} = 40 \text{ MHz}$ ; $f_1 = 4.43 \text{ MHz}$	54	56	-	dB	

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
EFFECTIVE BITS; NOTES 4 AND 5; SEE FIG.7						
EB	effective bits TDA8760K/2 ( $f_{\text{clk}} = 20 \text{ MHz}$ )	$f_i = 4.43 \text{ MHz}$	–	8.70	–	bits
		$f_i = 7.5 \text{ MHz}$	–	8.50	–	bits
	effective bits TDA8760K/4 ( $f_{\text{clk}} = 40 \text{ MHz}$ )	$f_i = 4.43 \text{ MHz}$	–	8.50	–	bits
		$f_i = 10 \text{ MHz}$	–	8.20	–	bits
effective bits TDA8760K/5 ( $f_{\text{clk}} = 50 \text{ MHz}$ )	$f_i = 4.43 \text{ MHz}$	–	8.25	–	bits	
	$f_i = 10 \text{ MHz}$	–	8.00	–	bits	
TWO-TONE						
Two-tone	two-tone intermodulation rejection	$f_{\text{clk}} = 40 \text{ MHz}$ ; note 8	–	–60	–	dB
BIT ERROR RATE						
BER	bit error rate	$f_{\text{clk}} = 40 \text{ MHz}$ ; $f_i = 4.43 \text{ MHz}$ ; $V_i = \pm 16 \text{ LSB}$ at code 512	–	$2 \times 10^{-12}$	–	times/ samples
DIFFERENTIAL GAIN; SEE FIG.5						
$G_{\text{diff}}$	differential gain	$f_{\text{clk}} = 20 \text{ MHz}$ ; $f_i = 4.43 \text{ MHz}$	–	0.5	tbf	%
		$f_{\text{clk}} = 40 \text{ MHz}$ ; $f_i = 4.43 \text{ MHz}$	–	1.0	tbf	%
DIFFERENTIAL PHASE						
$\Phi_{\text{diff}}$	differential phase	$f_{\text{clk}} = 40 \text{ MHz}$ ; $f_i = 4.43 \text{ MHz}$	–	0.1	0.2	deg
Timing (note 6; see Fig.3; $C_L = 15 \text{ pF}$ )						
$t_{\text{ds}}$	sampling delay time		–	–	2	ns
$t_{\text{h}}$	output hold time		8	–	–	ns
$t_{\text{d}}$	output delay time		–	12	16	ns
3-state output delay times (see Fig.4)						
$t_{\text{dZH}}$	enable HIGH		–	12	16	ns
$t_{\text{dZL}}$	enable LOW		–	12	16	ns
$t_{\text{dHZ}}$	disable HIGH		–	8	12	ns
$t_{\text{dLZ}}$	disable LOW		–	16	20	ns

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## Notes

- The circuit has two clock inputs: CLK and  $\overline{\text{CLK}}$ . There are three modes of operation:

### TTL mode 1:

CLK input is at TTL level with a threshold voltage of 1.5 V and sampling is taken on the falling edge of the clock input signal.  $\overline{\text{CLK}}$  decoupled to DGND via a 100 nF capacitor.

### TTL mode 2:

$\overline{\text{CLK}}$  input is at TTL level with threshold voltage of 1.5 V and sampling is taken on the rising edge of the clock input signal. CLK decoupled to DGND via a 100 nF capacitor.

### TTL mode3:

CLK and  $\overline{\text{CLK}}$  inputs are at differential TTL levels.

### AC driving modes:

When driving the CLK input directly and with any AC signal of minimum 0.5 V (p-p) and with a DC level of 1.5 V, the sampling takes place at the falling edge of the clock signal.

When driving the  $\overline{\text{CLK}}$  input with the same signal, sampling takes place at the rising edge of the clock signal. It is recommended to decouple the  $\overline{\text{CLK}}$  or CLK input to DGND via a 100 nF capacitor.

- THD (total harmonic distortion) is obtained with the addition of the first five harmonics:

$$\text{THD} = 20 \log \frac{F}{\sqrt{(2\text{nd})^2 + (3\text{rd})^2 + (4\text{th})^2 + (5\text{th})^2 + (6\text{th})^2}}$$

F being the fundamental harmonic referenced at 0 dB for a full-scale sinewave input.

- AC linearity: full-scale differential sinewave ( $f_i = 4.43 \text{ MHz}$ ;  $f_{\text{clk}} = 40 \text{ MHz}$ ).
- Effective bits with differential input and single input are respectively executed with full scale differential input and full scale single sinewave.
- Effective bits are obtained via a Fast Fourier Transformer (FFT) treatment taking 8K acquisition points per period. The calculation takes into account all harmonics and noise up to half of the clock frequency (NYQUIST frequency). Conversion to SNR:  $\text{SNR} = \text{EB} \times 6.02 + 1.76 \text{ dB}$ .
- Output data acquisition: the output data is available after the maximum delay of  $t_d$ .
- $t_{\text{CPH}}$  of 9 ns (minimum) can be applied at the penalty of 0.5 effective bit drop compared to typical values.
- Intermodulation measured relative to either tone with analog input frequencies of 4.43 MHz and 4.53 MHz. The two input signals have the same amplitude and the total amplitude of both signals provides full scale to the converter.
- The -3 dB (or -1 dB) analog bandwidth is determined by the 3 dB (or 1 dB) reduction in the reconstructed output, the input being a full-scale sine wave.

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**Table 1** Output coding with differential inputs (typical values to AGND);  $V_{I(p-p)} = V_{refH} - V_{refL} = 1.5$  V.

CODE	$V_{I(p-p)}$	$\bar{V}_{I(p-p)}$	IR	BINARY OUTPUTS	TWO'S COMPLEMENT OUTPUTS
				D9 TO D0	D9 TO D0
underflow	<3.025	>3.775	0	0000000000	1000000000
0	3.025	3.775	1	0000000000	1000000000
1	—	—	1	0000000001	1000000001
•	—	—	•	••••••••••	••••••••••
511	3.40	3.40	1	0111111111	1111111111
•	—	—	•	••••••••••	••••••~••••••
1022	—	—	1	1111111110	0111111110
1023	3.775	3.025	1	1111111111	0111111111
overflow	>3.775	<3.025	0	1111111111	0111111111

**Table 2** Output coding with single inputs (typical values to AGND);  $V_{I(p-p)} = V_{refH} - V_{refL} = 1.4$  V;  $\bar{V}_{I(p-p)} = 3.7$  V.

CODE	$V_{I(p-p)}$	IR	BINARY OUTPUTS	TWO'S COMPLEMENT OUTPUTS
			D9 TO D0	D9 TO D0
underflow	<3.0	0	0000000000	1000000000
0	3.0	1	0000000000	1000000000
1	—	1	0000000001	1000000001
•	—	•	••••••••••	••••••~••••••
511	3.7	1	0111111111	1111111111
•	—	•	••••••~••••••	••••••~••••••
1022	—	1	1111111110	0111111110
1023	4.4	1	1111111111	0111111111
overflow	>4.4	0	1111111111	0111111111

**Table 3** Mode selection.

OTC	CS	D0 TO D9 AND IR
1	1	binary; active
0	1	two's complement; active
X <sup>(1)</sup>	0	high impedance

**Note**

- Where: X = don't care.

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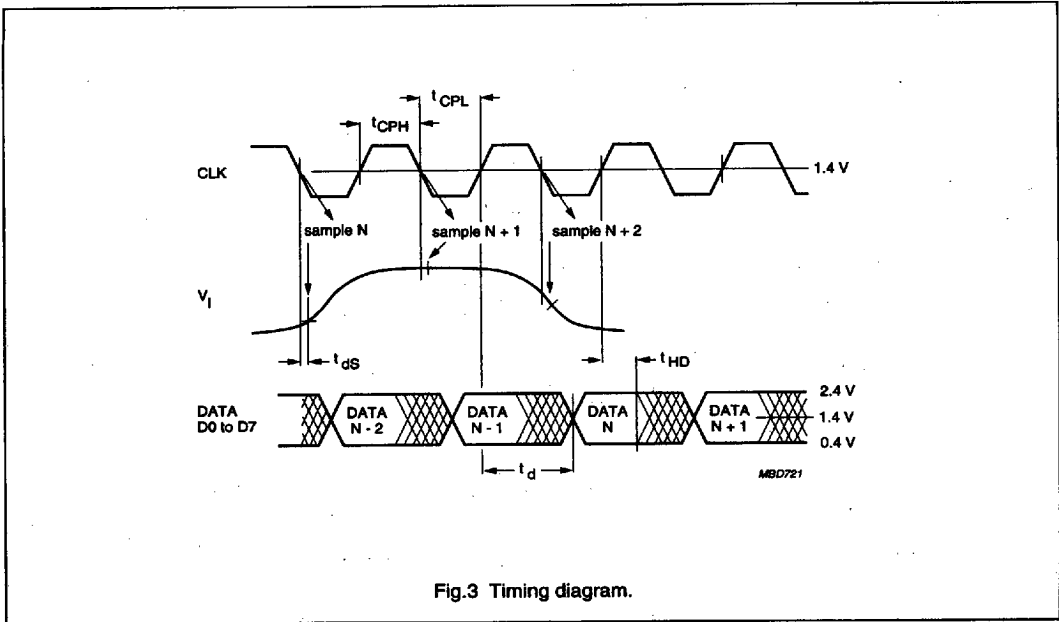


Fig.3 Timing diagram.

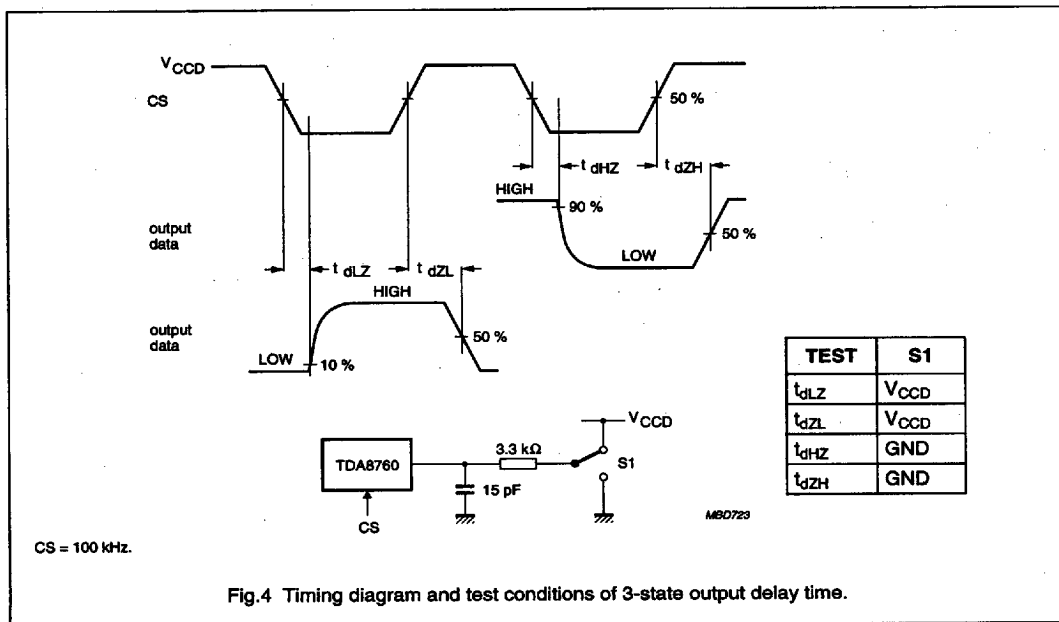
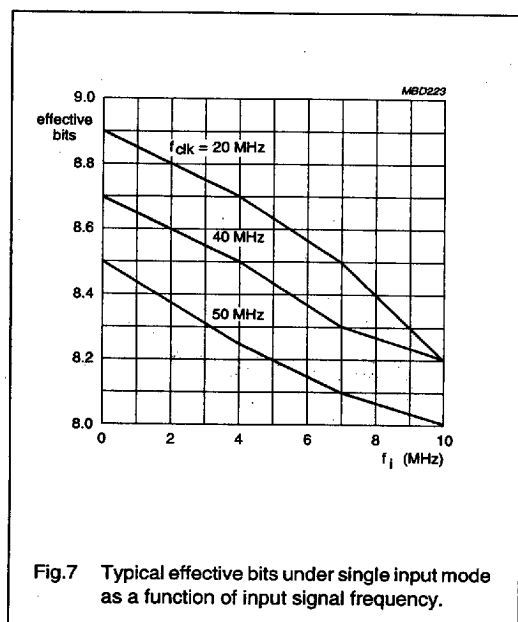
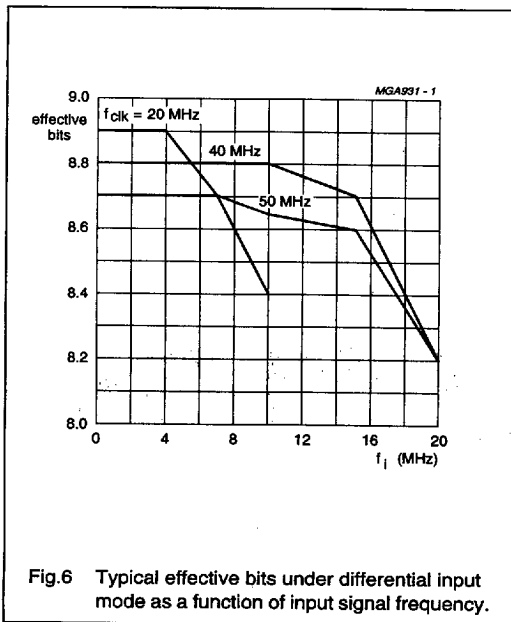
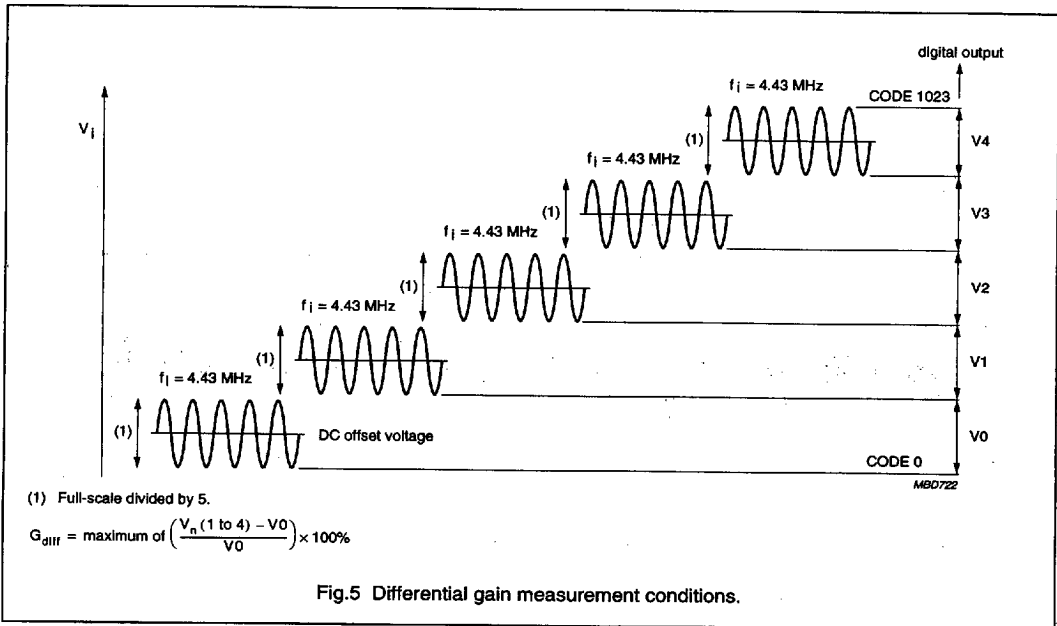


Fig.4 Timing diagram and test conditions of 3-state output delay time.

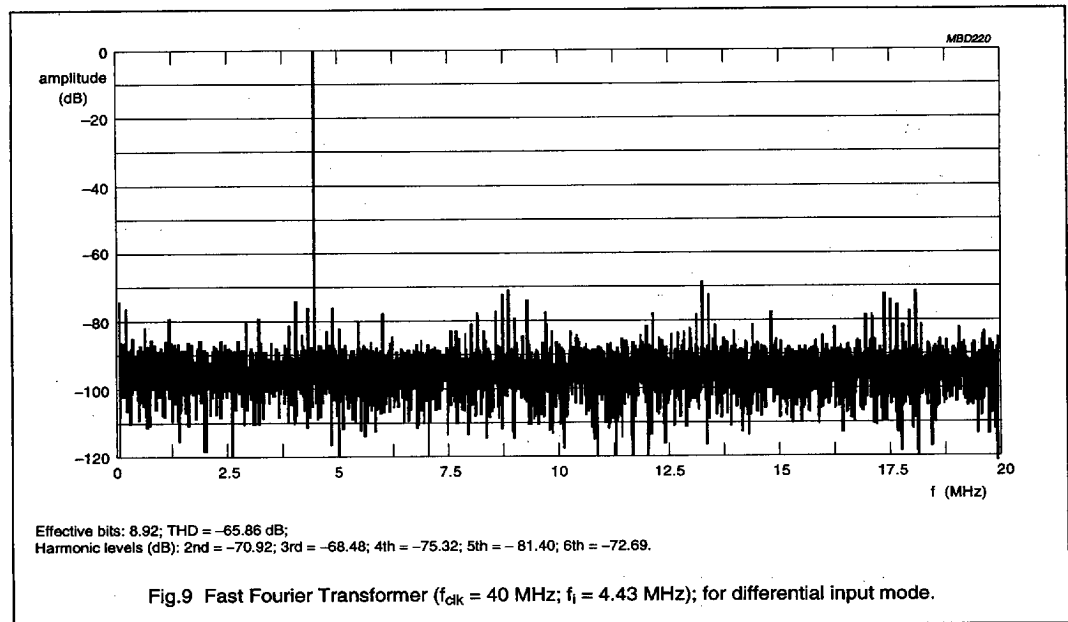
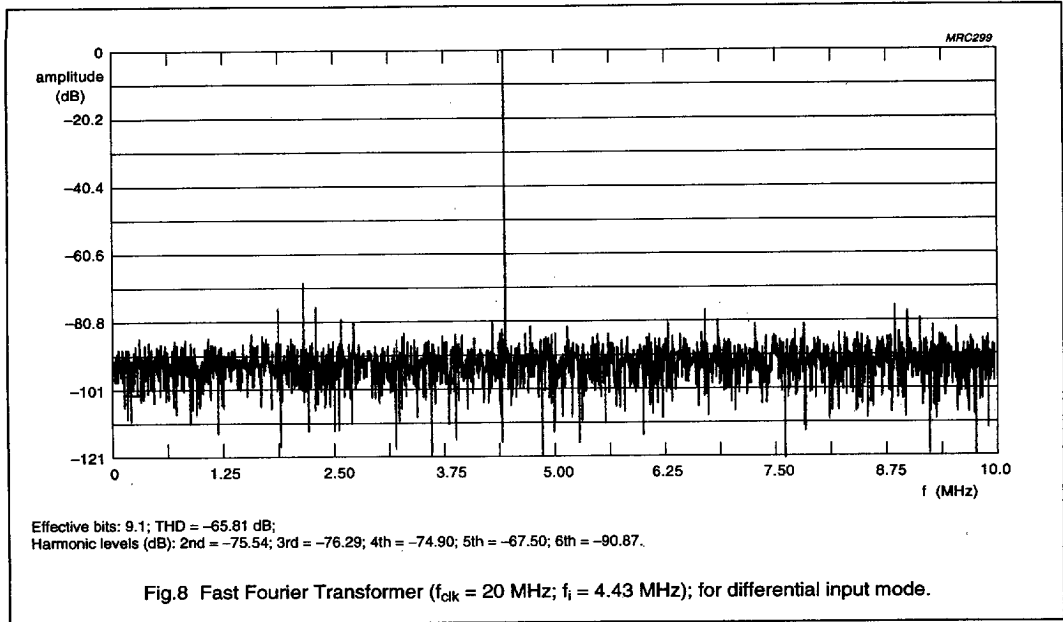
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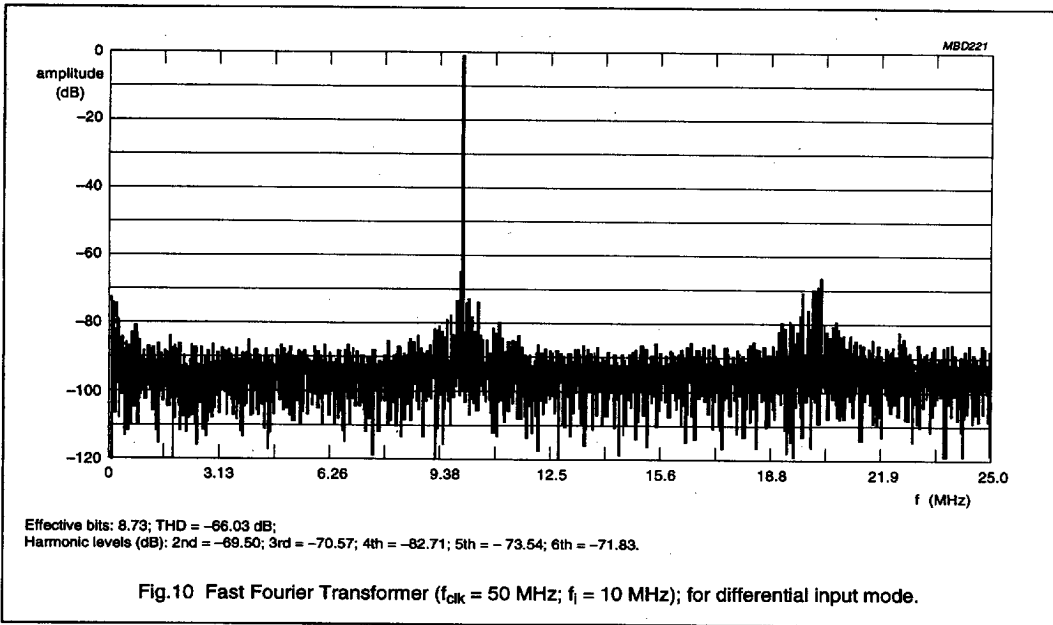
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## INTERNAL PIN CONFIGURATION

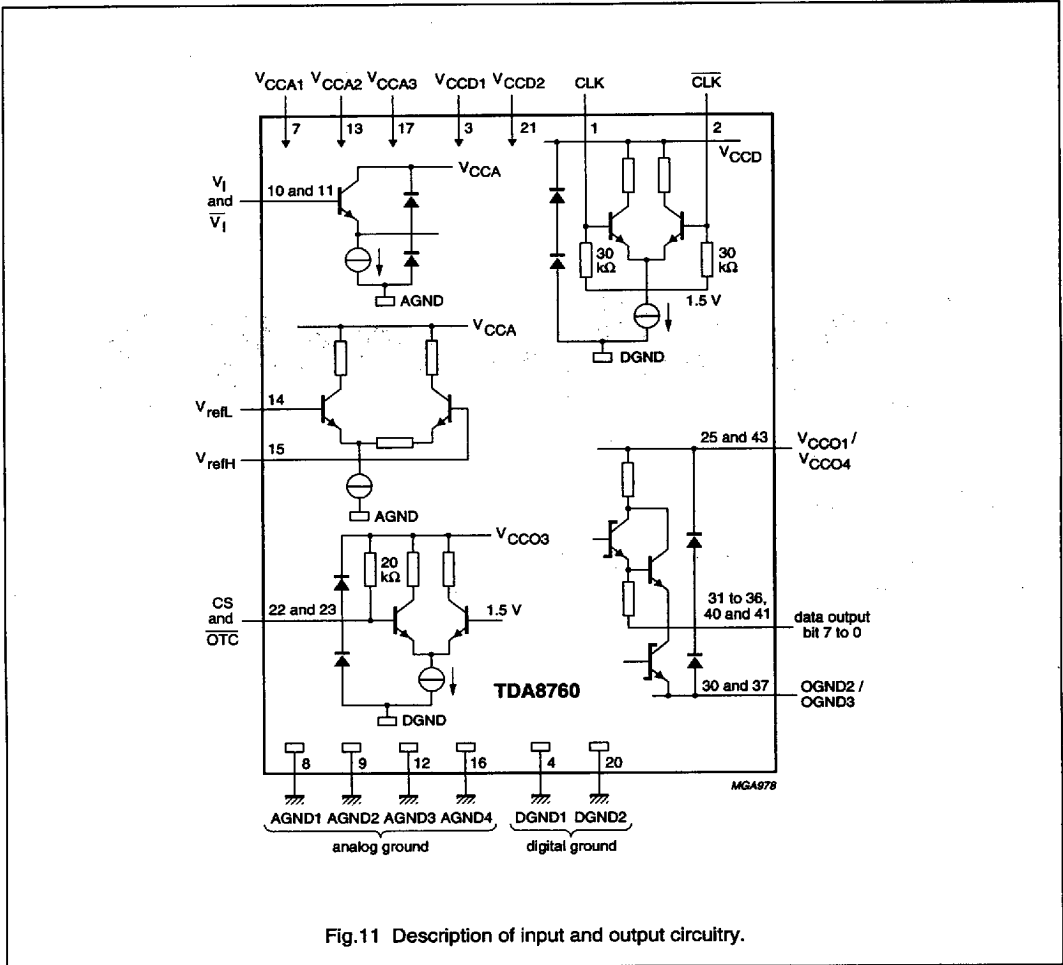
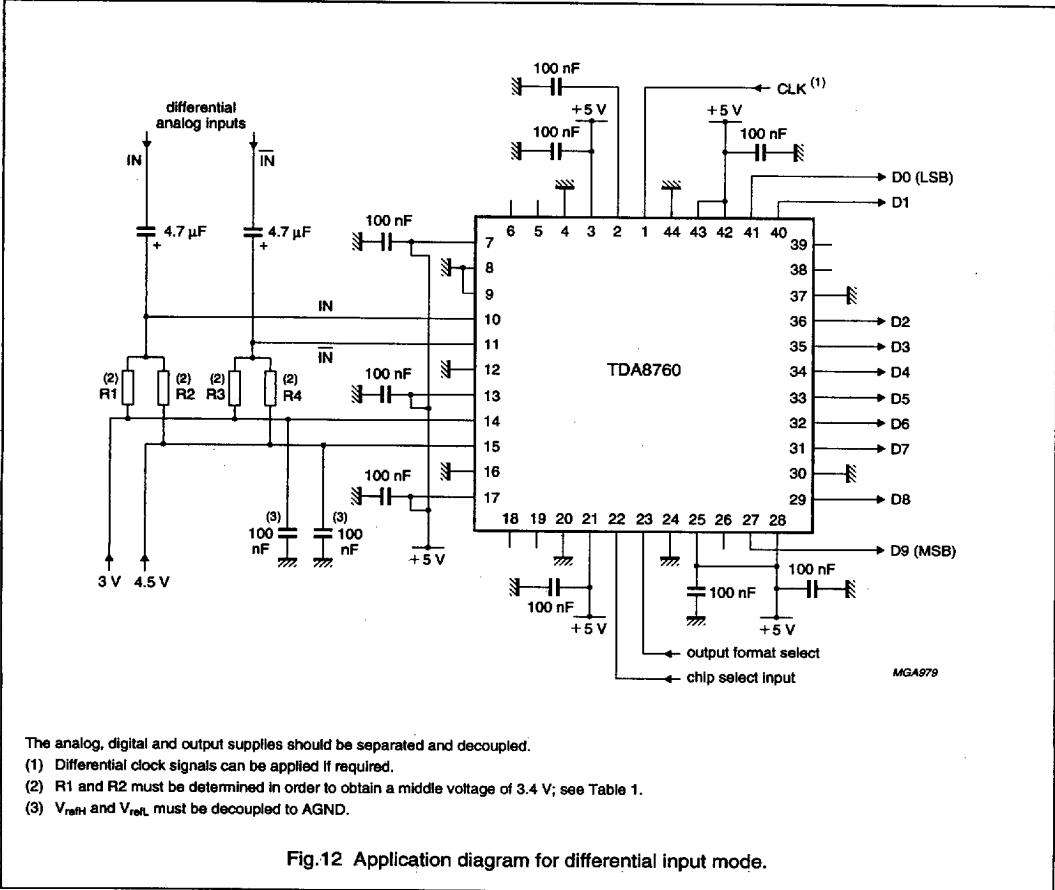


Fig.11 Description of input and output circuitry.

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### APPLICATION INFORMATION



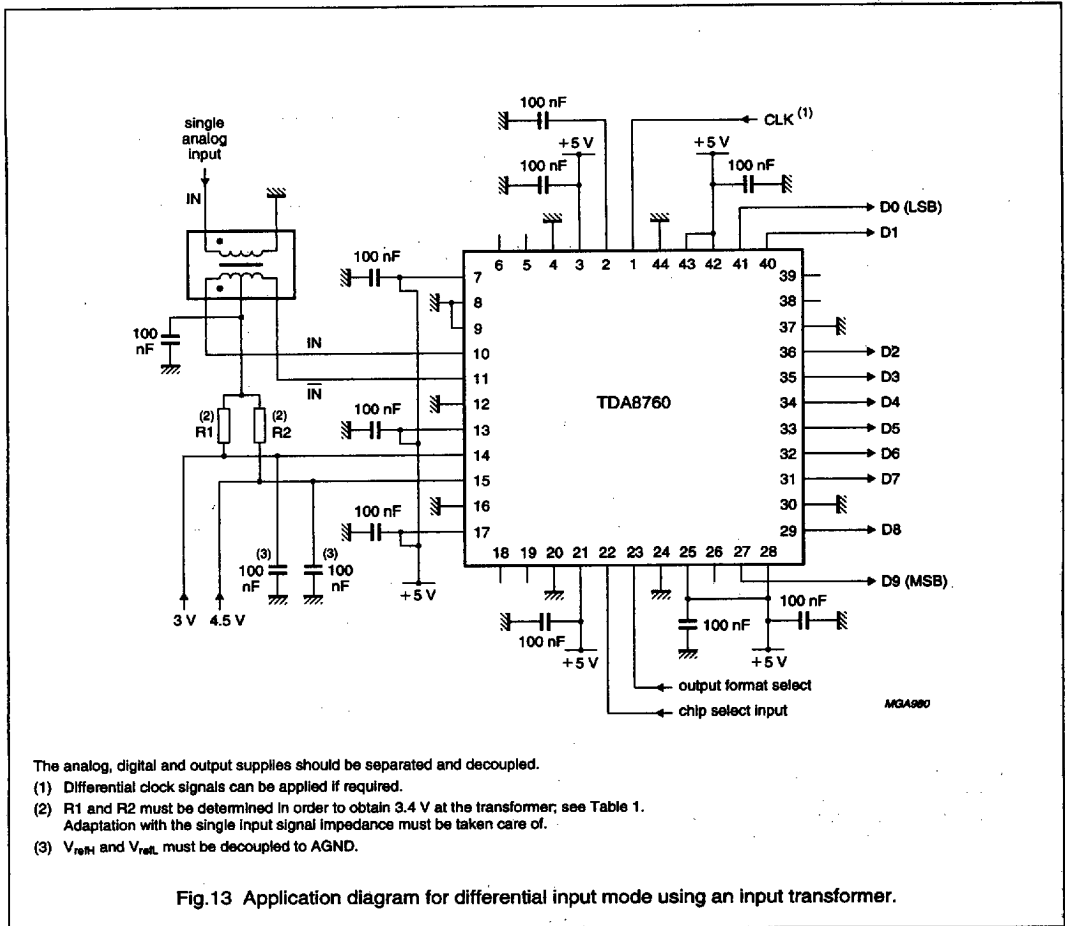
- The analog, digital and output supplies should be separated and decoupled.
- (1) Differential clock signals can be applied if required.
  - (2) R1 and R2 must be determined in order to obtain a middle voltage of 3.4 V; see Table 1.
  - (3)  $V_{\text{refH}}$  and  $V_{\text{refL}}$  must be decoupled to AGND.

Fig.12 Application diagram for differential input mode.



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