ACT6305
Rev0, 13-May-08
Low Input High Efficiency
Synchronous Step-Up DC/DC Converter

## FEATURES

- 1V Typical Startup Voltage
- Up to 94\% Efficiency
- Fully Integrated Power Stage
- Fixed 3.3V/5V or Adjustable Output Voltage
- NanoStandby ${ }^{\text {TM }}$ Shutdown Mode
- Tiny SOT23-5 Package
-RoHS Compliant Package


## APPLICATIONS

- Hand-Held Devices
- One to Three Cell Battery Operated Devices
- PDAs
- Cellular Phones
- Digital Cameras
- GPS


## GENERAL DESCRIPTION

The ACT6305 is a high efficiency synchronous step-up DC/DC converter that consumes just $17 \mu \mathrm{~A}$ of quiescent supply current and is capable of delivering more than 400 mA output current using only three external components. This device achieves a typical startup voltage of just 1 V , making it ideal for applications operating from 1 or 2 alkaline cells.

This device incorporates an internal synchronous rectifier, and achieves excellent peak efficiency of up to $94 \%$ and features a proprietary control scheme that ensures high efficiency over a wide load current range.
The ACT6305 features pin-selectable output voltages of 3.3 V or 5 V , or it can also be set externally to any voltage between 2 V and 5 V . The ACT6305 is available in a tiny SOT23-5 package.

## TYPICAL APPLICATION CIRCUIT




## ORDERING INFORMATION

| PART NUMBER | TEMPERATURE RANGE | PACKAGE | PINS | PACKING METHOD | TOP MARK |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ACT6305UC-T | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ | SOT23-5 | 5 | TAPE \& REEL | HBCA |

All Active-Semi components are RoHS Compliant and with Pb-free plating unless specified differently. The term "Pb-free" means semiconductor products that are in compliance with current RoHS (Restriction of Hazardous Substances) standards.

## PIN CONFIGURATION



## PIN DESCRIPTION

| PIN NUMBER | PIN NAME | PIN DESCRIPTION |
| :---: | :---: | :--- |
| 1 | FB | Output Voltage Selection and Feedback Input. Connect this pin to G for Vout $=5 \mathrm{~V}$, to <br> OUT for Vout $=3.3 V$. Connect FB to a resistor divider to set the output voltage be- <br> tween 2V and 5V. See the Output Voltage Selection section for more information. |
| 2 | G | Ground. |
| 3 | EN | Shutdown Input. Drive EN to OUT or to a logic high for normal operation, drive EN to <br> G or to a logic low to disable the regulator. |
| 4 | LX | Inductor Connection. |
| 5 | OUT | Output. ACT6305 regulates voltage at this pin. OUT also provides power to the IC. |

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## ABSOLUTE MAXIMUM RATINGS ${ }^{\circledR}$

| PARAMETER | VALUE | UNIT |
| :--- | :---: | :---: |
| OUT, EN to G | -0.3 to 6 | V |
| FB, LX to G | -0.3 to $\mathrm{V}_{\text {out }}+0.3$ | V |
| Continuous LX Current | Internally Limited | A |
| Maximum Power Dissipation (derate $5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ above $\mathrm{T}_{\mathrm{A}}=50^{\circ} \mathrm{C}$ ) | 0.53 | W |
| Junction to Ambient Thermal Resistance $\left(\theta_{\mathrm{JA}}\right)$ | 190 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Operating Junction Temperature | -40 to 150 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | -55 to 150 | ${ }^{\circ} \mathrm{C}$ |
| Lead Temperature (Soldering, 10 sec$)$ | 300 | ${ }^{\circ} \mathrm{C}$ |

(1): Do not exceed these limits to prevent damage to the device. Exposure to absolute maximum rating conditions for long periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}_{\text {IN }}=1.5 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right.$, unless otherwise specified.)

| PARAMETER | SYMBOL | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Minimum Operating Input Voltage | $\mathrm{V}_{\text {IN }}$ |  |  | 0.6 |  | V |
| Maximum Operating Input Voltage | $\mathrm{V}_{\text {IN }}$ |  |  |  | 5 | V |
| Startup Voltage | $\mathrm{V}_{\text {IN }}$ | $\mathrm{I}_{\text {LOAD }}=1 \mathrm{~mA}$ |  | 1 | 1.18 | V |
| Output Voltage | $V_{\text {OUt }}$ | FB = OUT; 60 mV PP Ripple | 3.2 | 3.3 | 3.4 | V |
|  |  | FB $=\mathrm{G} ; 40 \mathrm{mV} \mathrm{VPP}$ Ripple | 4.85 | 5 | 5.15 |  |
| Output Voltage Range |  | External Feedback Divider | 2 |  | 5 | V |
| Quiescent Current at OUT | $\mathrm{I}_{\mathrm{Q}}$ | $\mathrm{V}_{\mathrm{FB}}=1.3 \mathrm{~V}$ |  | 17 | 34 | $\mu \mathrm{A}$ |
| Shutdown Supply Current | $\mathrm{I}_{\text {SD }}$ | $\mathrm{EN}=\mathrm{G}$ |  | 0.1 | 1 | $\mu \mathrm{A}$ |
| Startup to Normal Transition Threshold |  |  |  | 1.85 |  | V |
| FB Feedback Voltage | $\mathrm{V}_{\text {FB }}$ | $\mathrm{V}_{\text {Out }}=2 \mathrm{~V}$ to 5 V | 1.15 | 1.180 | 1.21 | V |
| FB Input Current |  | $\mathrm{FB}=1.3 \mathrm{~V}$ |  | 50 |  | nA |
| Power Switch Current Limit | $\mathrm{I}_{\text {LIM }}$ |  | 0.7 | 1 | 1.25 | A |
| Power Switch On Resistance | R ONN | $\mathrm{ILx}^{\text {L }}=100 \mathrm{~mA}$ |  | 0.19 | 0.4 | $\Omega$ |
| Synchronous Rectifier On Resistance | $\mathrm{R}_{\text {ONP }}$ | $\mathrm{ILx}_{\mathrm{Lx}}=-100 \mathrm{~mA}$ |  | 0.21 | 0.4 | $\Omega$ |
| LX Leakage Current |  | $\mathrm{EN}=\mathrm{G}, \mathrm{V}_{\mathrm{LX}}=0 \mathrm{~V}$ or 3.3 V |  | 0.1 | 1 | $\mu \mathrm{A}$ |
| Maximum Switch On-Time | tonmax |  | 3.75 | 5 | 6.25 | $\mu \mathrm{s}$ |
| Minimum Switch Off-Time | $\mathrm{t}_{\text {fFFmin }}$ |  | 0.75 | 1 | 1.25 | $\mu \mathrm{s}$ |
| EN Logic Low Threshold | $\mathrm{V}_{\text {IL }}$ |  |  |  | 0.4 | V |
| EN Logic High Threshold | $\mathrm{V}_{\mathrm{IH}}$ |  | 0.6 |  |  | V |
| EN Input Current |  |  |  |  | 1 | $\mu \mathrm{A}$ |

## TYPICAL PERFORMANCE CHARACTERISTICS

(Typical Application Circuit, $\mathrm{L}=6.8 \mu \mathrm{H}, \mathrm{C} 1=22 \mu \mathrm{~F}$ ceramic, $\mathrm{C} 2=22 \mu \mathrm{~F}$ ceramic, unless otherwise specified.)


Efficiency vs. Constant Current


Startup Voltage vs. Constant Resistance Load


Efficiency vs. Constant Current


Efficiency vs. Constant Current


Startup Voltage vs. Constant Resistance Load


## TYPICAL PERFORMANCE CHARACTERISTICS CONT'D

(Typical Application Circuit, $\mathrm{L}=6.8 \mu \mathrm{H}, \mathrm{C} 1=22 \mu \mathrm{~F}$ ceramic, $\mathrm{C} 2=22 \mu \mathrm{~F}$ ceramic, unless otherwise specified.)


## TYPICAL PERFORMANCE CHARACTERISTICS CONT'D

(Typical Application Circuit, $\mathrm{L}=6.8 \mu \mathrm{H}, \mathrm{C} 1=22 \mu \mathrm{~F}$ ceramic, $\mathrm{C} 2=22 \mu \mathrm{~F}$ ceramic, unless otherwise specified.)

Light Load Operation for Low ESR Cout


CH1: VLx, 2.00V/div
CH2: V out, $100 \mathrm{mV} /$ div
TIME: $100 \mu \mathrm{~s} / \mathrm{div}$

Medium Load Operation for Low ESR Cout


CH1: VLx, $2.00 \mathrm{~V} /$ div
CH2: Vout, $100 \mathrm{mV} /$ div
TIME: $2.50 \mu \mathrm{~s} / \mathrm{div}$

Light Load Operation for Low ESR Cout


CH1: V $\mathrm{Lx}, 2.00 \mathrm{~V} / \mathrm{div}$
CH2: Vout, $50 \mathrm{mV} / \mathrm{div}$
TIME: $25.0 \mu \mathrm{~s} / \mathrm{div}$

Medium Load Operation for Low ESR Cout


CH1: VLx, 2.00V/div
CH2: V
TIME: $2.50 \mu \mathrm{~s} / \mathrm{div}$

Heavy Load Operation for Low ESR Cout


CH1: VLx, $2.00 \mathrm{~V} / \mathrm{div}$
CH2: $\mathrm{V}_{\text {out }}, 50 \mathrm{mV} / \mathrm{div}$
TIME: $2.50 \mu \mathrm{~s} / \mathrm{div}$

Medium Load Operation for Low ESR Cout


CH1: V $\mathrm{Lx}, 2.00 \mathrm{~V} /$ div
CH2: Vout, $100 \mathrm{mV} / \mathrm{div}$
TIME: $2.50 \mu \mathrm{~s} / \mathrm{div}$

## FUNCTIONAL BLOCK DIAGRAM



## FUNCTIONAL DESCRIPTION

As seen in the Functional Block Diagram, the ACT6305 comprises a reference, low input startup circuitry, feedback select, error comparator, control logic and timers, internal power MOSFETs and current comparators.

## Main Control Loop

The ACT6305 control architecture utilizes a series of one-shot and current sense circuits to regulate the output voltage with a very low quiescent supply current of just $17 \mu \mathrm{~A}$. An ON cycle begins when FB falls below the ACT6305's internal reference voltage. During an ON cycle, the N -channel power switch is turned on to connect the LX node to ground, transferring energy from input supply to the inductor. The ON cycle persists or the 1A current limit is reached. When either of these conditions occurs, the N -channel power switch turns off and the OFF cycle begins. During the OFF cycle, the energy stored in the inductor is released into the output capacitor and load through the synchronous rectifier. Once the $1 \mu s$ minimum OFF time expires, another ON cycle begins when the FB voltage drops below the reference voltage.

## Light Load Operation

Under light load conditions, the IC automatically enters a power-saving mode by extending the OFF time. This reduces switching losses and allows the ACT6305 to maintain high efficiency across a very wide load range.

## Low Voltage Startup

The ACT6305 uses proprietary start-up circuitry that allows it to achieve very low voltage startup down to 1V. The ACT6305 remains in low voltage startup mode until $\mathrm{V}_{\text {out }}$ exceeds 1.85 V . The ACT6305 cannot support its full rated load current when operating in low-voltage startup mode. Care should be taken to minimize load current during startup.

## Shutdown

In shutdown mode, all internal circuits including the power switch and the synchronous rectifier are disabled, reducing quiescent supply current to less than $1 \mu \mathrm{~A}$. Note that in shutdown mode, a DC path from the input supply to the output still exists due to the body diode of the synchronous rectifier.

## APPLICATIONS INFORMATION

## Output Voltage Selection

The ACT6305 provides fixed output voltage options of 3.3 V or 5 V , or the output voltage may be adjusted over a 2 V to 5 V range by connecting an external resistive voltage divider. When Output Voltage is set by external resistive voltage divider, add a 47 pF to 150 pF capacitor as feedforward capacitor ( $\mathrm{C}_{\mathrm{FF}}$ ) as shown in Figure 1.

Table 1:

## Output Voltage Programming

| $\mathrm{FB}=\mathrm{OUT}$ | $\mathrm{V}_{\text {OUT }}=3.3 \mathrm{~V}$ |
| :---: | :---: |
| $\mathrm{FB}=\mathrm{G}$ | $\mathrm{V}_{\text {OUT }}=5 \mathrm{~V}$ |
| FB connect to resistive <br> divider | $\mathrm{V}_{\text {OUT }}=1.18 \mathrm{~V}\left(1+\mathrm{R}_{\mathrm{FB} 1} / \mathrm{R}_{\mathrm{FB} 2}\right)$ |

Figure 1:
Output Voltage Setting


## Output Capacitor Selection

A minimum value of output capacitance is required to maintain loop stability and normal operation of the IC. The output capacitor value should be in the range of $22 \mu \mathrm{~F}$ to $100 \mu \mathrm{~F}$. To obtain small output ripple, use a large capacitor with low ESR. Ceramic capacitors should be used for highest performance. If a tantalum capacitor is used, choose only low ESR types and a smaller low ESR capacitor of about $1 \mu \mathrm{~F}$ can be connected in parallel to filter high frequency noise.

## Inductor Selection

For most applications, the inductor value should be in the range of $4.7 \mu \mathrm{H}$ to $22 \mu \mathrm{H}$. Smaller inductors provide faster load transient response and have a smaller physical size, but they also result in higher ripple current and reduce the maximum available output current. Choose inductors with low series resistance to obtain the highest efficiency.

## Board Layout

To reduce noise and increase efficiency, high current traces should be wide and direct, and an extended ground plane should be used. Switching current paths should be laid out as tightly as possible, with the inductor and input and output capacitors located close to the IC in order to reduce electromagnetic radiation.

If an external resistor divider is used, place the resistors very close to the FB and G pins, and keep them away from the high switching current paths.


## PACKAGE OUTLINE

## SOT23-5 PACKAGE OUTLINE AND DIMENSIONS



| SYMBOL | DIMENSION IN <br> MILLIMETERS |  | DIMENSION IN <br> INCHES |  |
| :---: | :---: | :---: | :---: | :---: |
|  | MIN | MAX | MIN | MAX |
|  | 1.050 | 1.250 | 0.041 | 0.049 |
| A1 | 0.000 | 0.100 | 0.000 | 0.004 |
| A2 | 1.050 | 1.150 | 0.041 | 0.045 |
| b | 0.300 | 0.400 | 0.012 | 0.016 |
| c | 0.100 | 0.200 | 0.004 | 0.008 |
| D | 2.820 | 3.020 | 0.111 | 0.119 |
| E | 1.500 | 1.700 | 0.059 | 0.067 |
| E1 | 2.650 | 2.950 | 0.104 | 0.116 |
| e | 0.950 TYP |  | 0.037 TYP |  |
| e1 | 1.800 | 2.000 | 0.071 | 0.079 |
| L | 0.700 REF |  | 0.028 |  |
| R1EF |  |  |  |  |
| $\theta$ | 0.300 | 0.600 | 0.012 |  |
| $0^{\circ}$ |  | $8^{\circ}$ | $0^{\circ}$ |  |
| $8^{\circ}$ |  |  |  |  |

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