

ISL9N305ASK8T

N-Channel Logic Level PWM Optimized UltraFET® Trench Power MOSFET

General Description

This device employs a new advanced trench MOSFET technology and features low gate charge while maintaining low on-resistance.

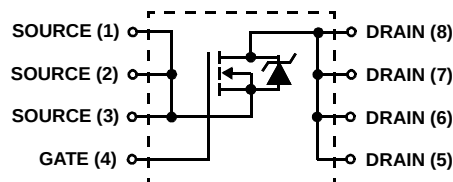
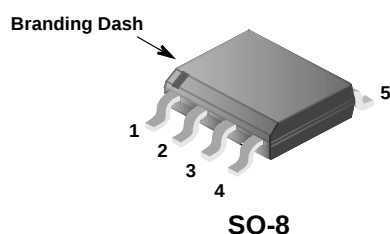
Optimized for switching applications, this device improves the overall efficiency of DC/DC converters and allows operation to higher switching frequencies.

Applications

- DC/DC converters

Features

- Fast switching
- $r_{DS(ON)} = 0.004\Omega$ (Typ), $V_{GS} = 10V$
- $r_{DS(ON)} = 0.0064\Omega$ (Typ), $V_{GS} = 4.5V$
- Q_g (Typ) = 38nC, $V_{GS} = 5V$
- Q_{gd} (Typ) = 11.5nC
- C_{ISS} (Typ) = 4260pF



MOSFET Maximum Ratings $T_A = 25^\circ C$ unless otherwise noted

Symbol	Parameter	Ratings	Units
V_{DSS}	Drain to Source Voltage	30	V
V_{GS}	Gate to Source Voltage	± 20	V
I_D	Drain Current		
	Continuous ($T_A = 25^\circ C$, $V_{GS} = 10V$)	18	A
	Continuous ($T_A = 100^\circ C$, $V_{GS} = 4.5V$)	9	A
	Pulsed	Figure 4	A
P_D	Power dissipation	2.5	W
	Derate above $25^\circ C$	20	mW/ $^\circ C$
T_J, T_{STG}	Operating and Storage Temperature	-55 to 150	$^\circ C$

Thermal Characteristics

$R_{\theta JA}$	FR-4 board with 0.76 in ² (490 mm ²) copper pad at 10 seconds	50	$^\circ C/W$
$R_{\theta JA}$	FR-4 board with 0.054 in ² (34.8 mm ²) copper pad at 1000 seconds	152	$^\circ C/W$
$R_{\theta JA}$	FR-4 board with 0.0115 in ² (7.42 mm ²) copper pad at 1000 seconds	189	$^\circ C/W$

Package Marking and Ordering Information

Device Marking	Device	Reel Size	Tape Width	Quantity
N305ASK8	ISL9N305ASK8T	330mm	12mm	2500units

Electrical Characteristics $T_A = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Typ	Max	Units
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Off Characteristics

$B_{V_{DSS}}$	Drain to Source Breakdown Voltage	$I_D = 250\mu\text{A}$, $V_{GS} = 0\text{V}$	30	-	-	V
I_{DSS}	Zero Gate Voltage Drain Current	$V_{DS} = 25\text{V}$ $V_{GS} = 0\text{V}$ $T_A = 150^\circ\text{C}$	-	-	1	μA
I_{GSS}	Gate to Source Leakage Current	$V_{GS} = \pm 20\text{V}$	-	-	± 100	nA

On Characteristics

$V_{GS(TH)}$	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}$, $I_D = 250\mu\text{A}$	1	-	3	V
$r_{DS(ON)}$	Drain to Source On Resistance	$I_D = 18\text{A}$, $V_{GS} = 10\text{V}$ $I_D = 9\text{A}$, $V_{GS} = 4.5\text{V}$	-	0.004 0.0064	0.005 0.008	Ω

Dynamic Characteristics

C_{ISS}	Input Capacitance	$V_{DS} = 15\text{V}$, $V_{GS} = 0\text{V}$, $f = 1\text{MHz}$	-	4260	-	pF
C_{OSS}	Output Capacitance		-	750	-	pF
C_{RSS}	Reverse Transfer Capacitance		-	340	-	pF
$Q_{g(TOT)}$	Total Gate Charge at 10V	$V_{GS} = 0\text{V}$ to 10V	$V_{DD} = 15\text{V}$ $I_D = 9\text{A}$ $I_g = 1.0\text{mA}$	72	108	nC
$Q_{g(5)}$	Total Gate Charge at 5V	$V_{GS} = 0\text{V}$ to 5V		38	57	nC
$Q_{g(TH)}$	Threshold Gate Charge	$V_{GS} = 0\text{V}$ to 1V		4.1	6.1	nC
Q_{gs}	Gate to Source Gate Charge			10.4	-	nC
Q_{gd}	Gate to Drain "Miller" Charge			11.5	-	nC

Switching Characteristics ($V_{GS} = 4.5\text{V}$)

t_{ON}	Turn-On Time	$V_{DD} = 15\text{V}$, $I_D = 9\text{A}$ $V_{GS} = 4.5\text{V}$, $R_{GS} = 3.9\Omega$	-	-	158	ns
$t_{d(ON)}$	Turn-On Delay Time		-	24	-	ns
t_r	Rise Time		-	81	-	ns
$t_{d(OFF)}$	Turn-Off Delay Time		-	44	-	ns
t_f	Fall Time		-	52	-	ns
t_{OFF}	Turn-Off Time		-	-	145	ns

Switching Characteristics ($V_{GS} = 10\text{V}$)

t_{ON}	Turn-On Time	$V_{DD} = 15\text{V}$, $I_D = 9\text{A}$ $V_{GS} = 10\text{V}$, $R_{GS} = 3.9\Omega$	-	-	100	ns
$t_{d(ON)}$	Turn-On Delay Time		-	12	-	ns
t_r	Rise Time		-	55	-	ns
$t_{d(OFF)}$	Turn-Off Delay Time		-	66	-	ns
t_f	Fall Time		-	50	-	ns
t_{OFF}	Turn-Off Time		-	-	173	ns

Unclamped Inductive Switching

t_{AV}	Avalanche Time	$I_D = 2.2\text{A}$, $L = 3\text{mH}$	145	-	-	μs
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Drain-Source Diode Characteristics

V_{SD}	Source to Drain Diode Voltage	$I_{SD} = 9\text{A}$	-	-	1.25	V
		$I_{SD} = 4\text{A}$	-	-	1.0	V
t_{rr}	Reverse Recovery Time	$I_{SD} = 9\text{A}$, $dI_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	40	ns
Q_{RR}	Reverse Recovered Charge	$I_{SD} = 9\text{A}$, $dI_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	38	nC

Typical Characteristic

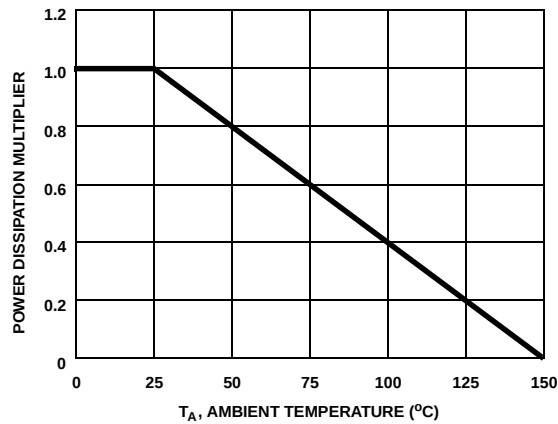


Figure 1. Normalized Power Dissipation vs Ambient Temperature

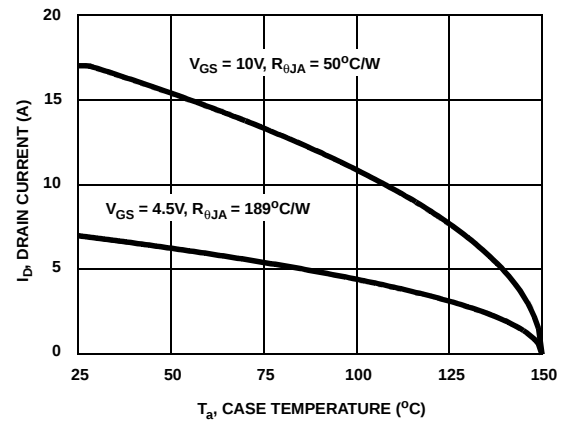


Figure 2. Maximum Continuous Drain Current vs Case Temperature

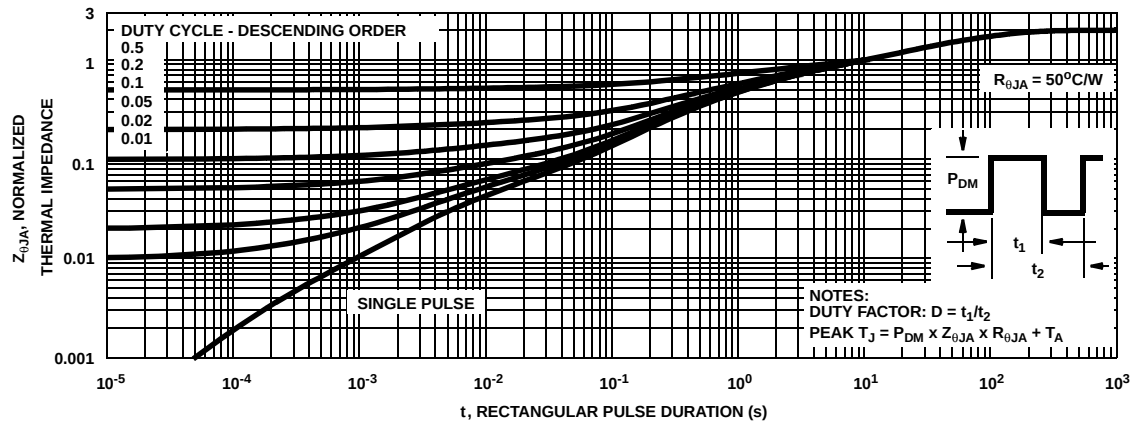


Figure 3. Normalized Maximum Transient Thermal Impedance

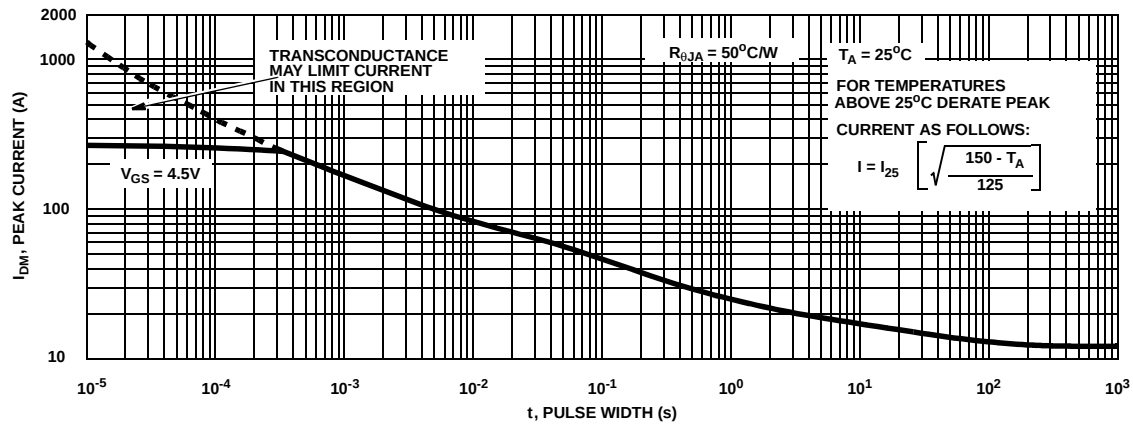


Figure 4. Peak Current Capability

Typical Characteristic (Continued)

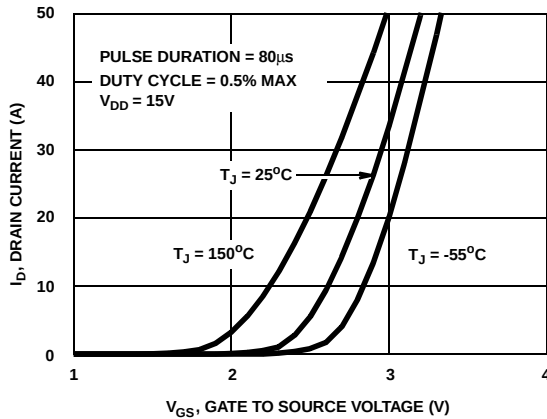


Figure 5. Transfer Characteristics

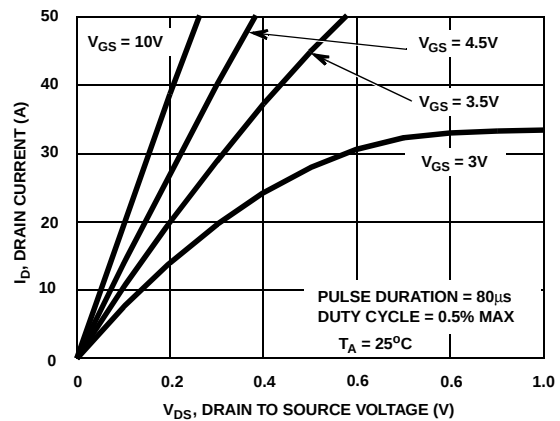


Figure 6. Saturation Characteristics

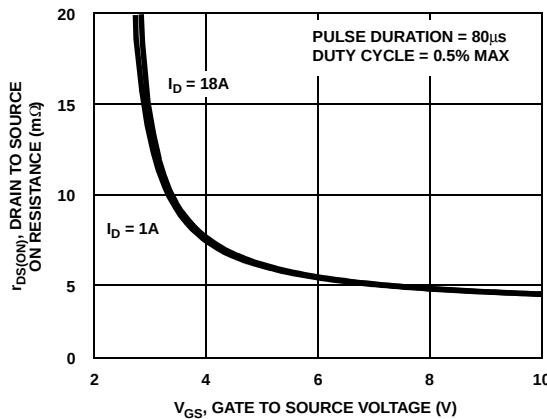


Figure 7. Drain To Source On Resistance vs Gate Voltage And Drain Current

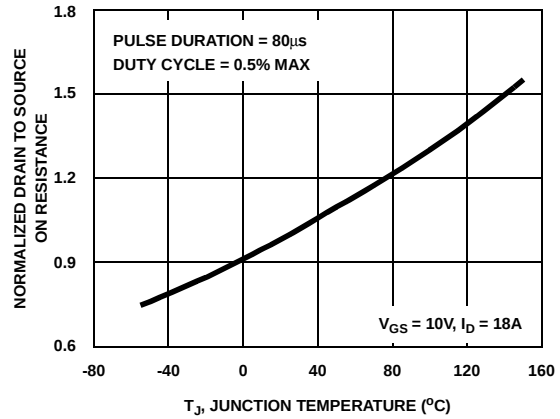


Figure 8. Normalized Drain to Source On Resistance vs Junction Temperature

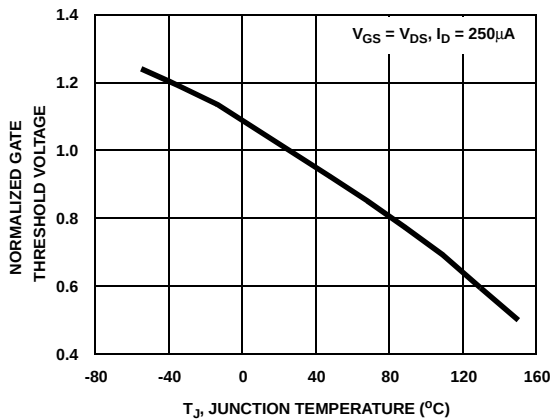


Figure 9. Normalized Gate Threshold Voltage vs Junction Temperature

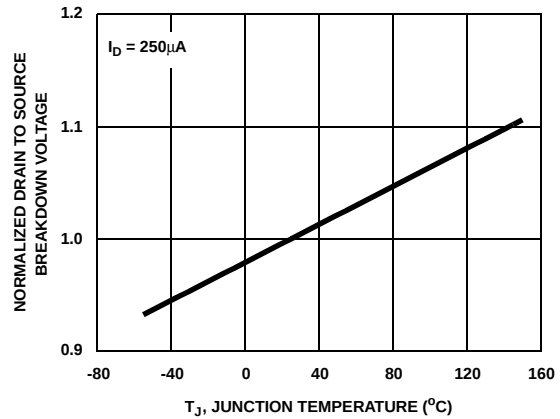


Figure 10. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

Typical Characteristic (Continued)

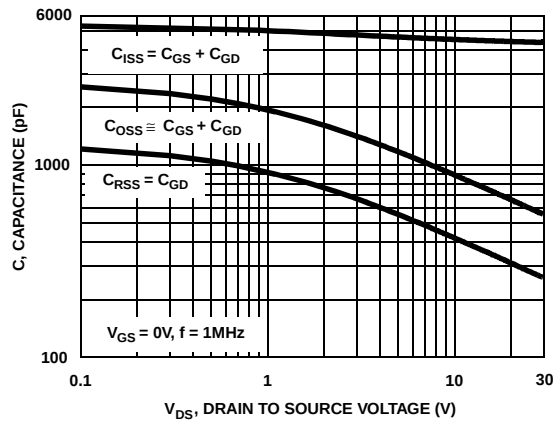


Figure 11. Capacitance vs Drain to Source Voltage

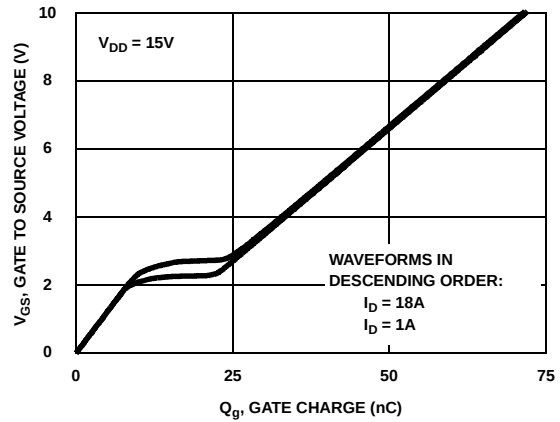


Figure 12. Gate Charge Waveforms for Constant Gate Currents

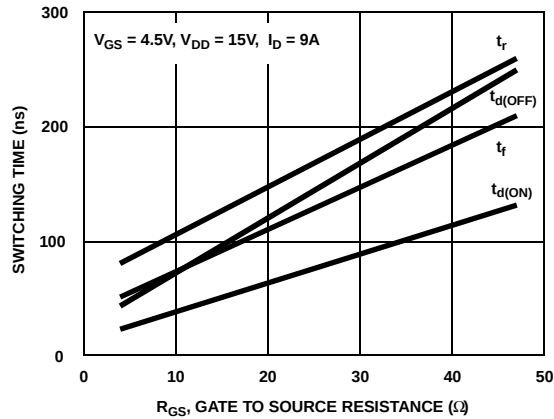


Figure 13. Switching Time vs Gate Resistance

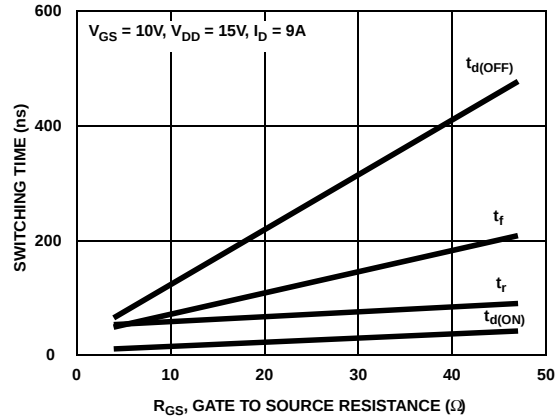


Figure 14. Switching Time vs Gate Resistance

Test Circuits and Waveforms

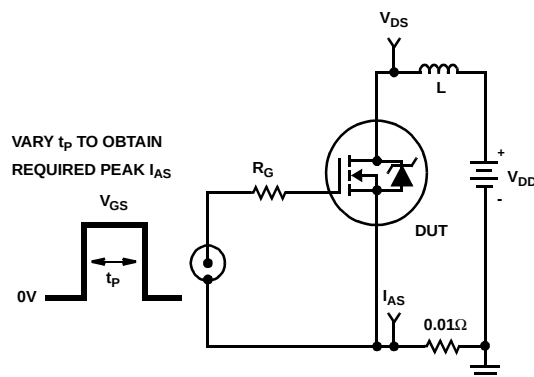


Figure 15. Unclamped Energy Test Circuit

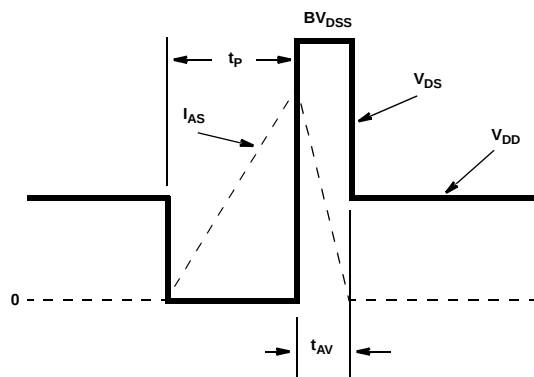


Figure 16. Unclamped Energy Waveforms

Test Circuits and Waveforms (Continued)

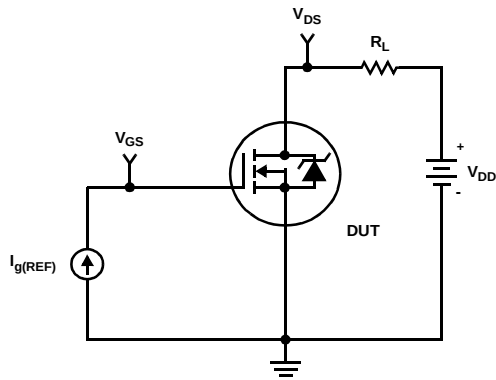


Figure 17. Gate Charge Test Circuit

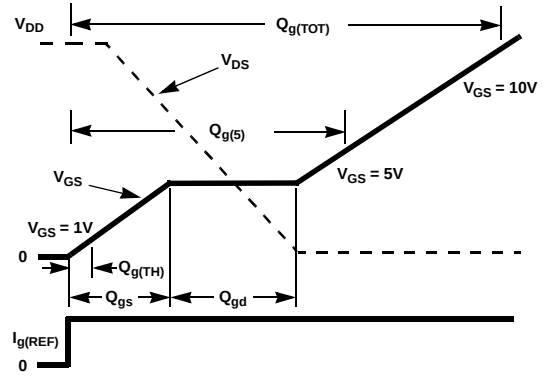


Figure 18. Gate Charge Waveforms

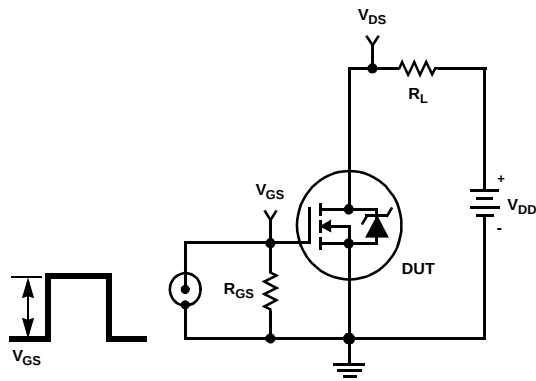


Figure 19. Switching Time Test Circuit

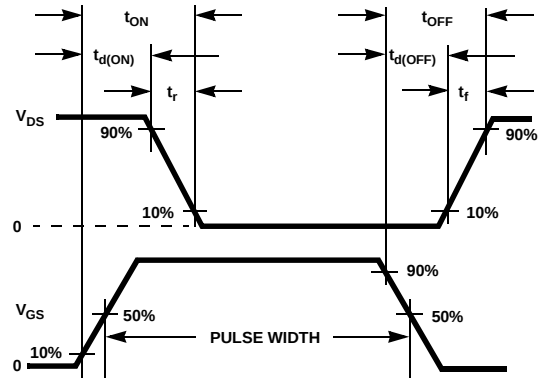


Figure 20. Switching Time Waveforms

Thermal Resistance vs. Mounting Pad Area

The maximum rated junction temperature, T_{JM} , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation, P_{DM} , in an application. Therefore the application's ambient temperature, T_A ($^{\circ}\text{C}$), and thermal resistance $R_{\theta JA}$ ($^{\circ}\text{C/W}$) must be reviewed to ensure that T_{JM} is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\theta JA}} \quad (\text{EQ. 1})$$

In using surface mount devices such as the SO8 package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of P_{DM} is complex and influenced by many factors:

1. Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board.
2. The number of copper layers and the thickness of the board.
3. The use of external heat sinks.
4. The use of thermal vias.
5. Air flow and board orientation.
6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

Fairchild provides thermal information to assist the designer's preliminary application evaluation. Figure 21 defines the $R_{\theta JA}$ for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the Fairchild device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

Displayed on the curve are $R_{\theta JA}$ values listed in the Electrical Specifications table. The points were chosen to

depict the compromise between the copper board area, the thermal resistance and ultimately the power dissipation, P_{DM} .

Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2. $R_{\theta JA}$ is defined as the natural log of the area times a coefficient added to a constant. The area, in square inches is the top copper area including the gate and source pads.

$$R_{\theta JA} = 83.2 - 23.6 \times \ln(\text{Area}) \quad (\text{EQ. 2})$$

The transient thermal impedance ($Z_{\theta JA}$) is also effected by varied top copper board area. Figure 22 shows the effect of copper pad area on single pulse transient thermal impedance. Each trace represents a copper pad area in square inches corresponding to the descending list in the graph. Spice and SABER thermal models are provided for each of the listed pad areas.

Copper pad area has no perceivable effect on transient thermal impedance for pulse widths less than 100ms. For pulse widths less than 100ms the transient thermal impedance is determined by the die and package. Therefore, C THERM1 through C THERM5 and R THERM1 through R THERM5 remain constant for each of the thermal models. A listing of the model component values is available in Table 1.

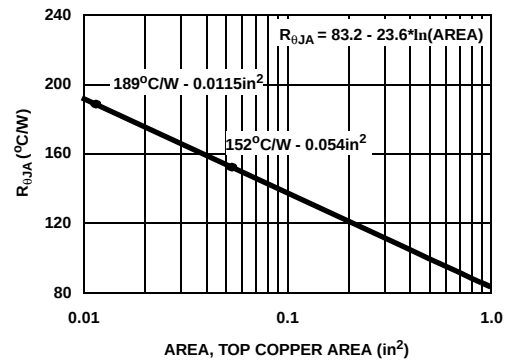


Figure 21. Thermal Resistance vs Mounting Pad Area

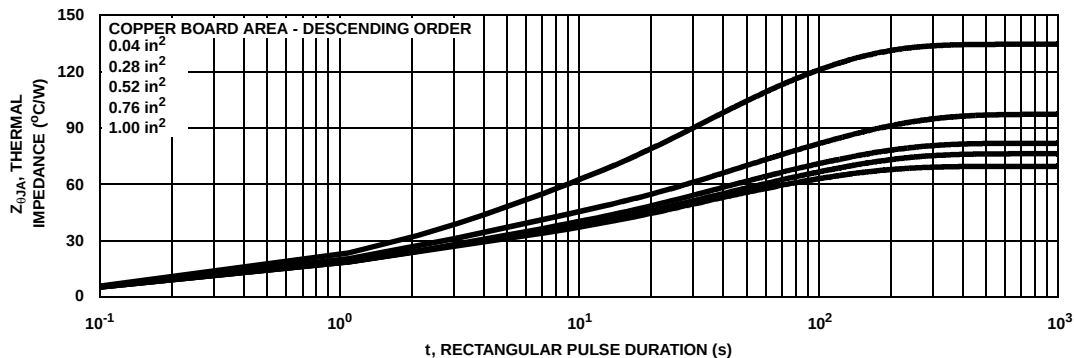


Figure 22. Thermal Impedance vs Mounting Pad Area

PSPICE Electrical Model

.SUBCKT ISL9N305ASK8 2 1 3 ; rev May 2001

CA 12 8 2e-9
 CB 15 14 2.3e-9
 CIN 6 8 4.1e-9

DBODY 7 5 DBODYMOD
 DBREAK 5 11 DBREAKMOD
 DPLCAP 10 5 DPLCAPMOD

EBREAK 11 7 17 18 32.8
 EDS 14 8 5 8 1
 EGS 13 8 6 8 1
 ESG 6 10 6 8 1
 EVTHRES 6 21 19 8 1
 EVTEMP 20 6 18 22 1

IT 8 17 1

LDRAIN 2 5 1e-9
 LGATE 1 9 1.29e-9
 LSOURCE 3 7 1.75e-10

MMED 16 6 8 8 MMEDMOD
 MSTRO 16 6 8 8 MSTROMOD
 MWEAK 16 21 8 8 MWEAKMOD

RBREAK 17 18 RBREAKMOD 1
 RDRAIN 50 16 RDRAINMOD 1e-3
 RGATE 9 20 2.34
 RLDRAIN 2 5 10
 RLGATE 1 9 12.9
 RLSOURCE 3 7 1.75
 RSLC1 5 51 RSLCMOD 1e-6
 RSLC2 5 50 1e3
 RSOURCE 8 7 RSOURCEMOD 3.3e-3
 RVTHRES 22 8 RVTHRESMOD 1
 RVTEMP 18 19 RVTEMPMOD 1

S1A 6 12 13 8 S1AMOD
 S1B 13 12 13 8 S1BMOD
 S2A 6 15 14 13 S2AMOD
 S2B 13 15 14 13 S2BMOD

VBAT 22 19 DC 1

$$\text{ESLC } 51 \ 50 \ \text{VALUE} = \{(V(5,51)/\text{ABS}(V(5,51))) * (\text{PWR}(V(5,51)/(1e-6 * 150), 5))\}$$

.MODEL DBODYMOD D (IS = 2e-11 N = 1.04 RS = 4e-3 TRS1 = 1e-3 TRS2 = 1e-6 XTI=2 CJO = 1.4e-9 TT = 9e-10 M = 0.45 XTI=2)

.MODEL DBREAKMOD D (RS = 0.22 TRS1 = 8e-4 TRS2 = -8.9e-6)

.MODEL DPLCAPMOD D (CJO = 1.27e-9 IS = 1e-30 N = 10 M = 0.46)

.MODEL MMEDMOD NMOS (VTO = 1.87 KP = 10 IS=1e-30 N = 10 TOX = 1 L = 1u W = 1u RG = 2.34)

.MODEL MSTROMOD NMOS (VTO = 2.25 KP = 140 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u)

.MODEL MWEAKMOD NMOS (VTO = 1.5 KP = 0.1 IS = 1e-30 N = 10 TOX = 1 L = 1u W = 1u RG = 23.4 RS = 0.1)

.MODEL RBREAKMOD RES (TC1 = 9.2e-4 TC2 = 0)

.MODEL RDRAINMOD RES (TC1 = 1.2e-2 TC2 = 3e-5)

.MODEL RSLCMOD RES (TC1 = 1e-3 TC2 = 1e-7)

.MODEL RSOURCEMOD RES (TC1 = 1e-3 TC2 = 1e-6)

.MODEL RVTHRESMOD RES (TC1 = -1.6e-3 TC2 = -8e-6)

.MODEL RVTEMPMOD RES (TC1 = -2.9e-3 TC2 = 1e-7)

.MODEL S1AMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -3.0 VOFF= -2.0)

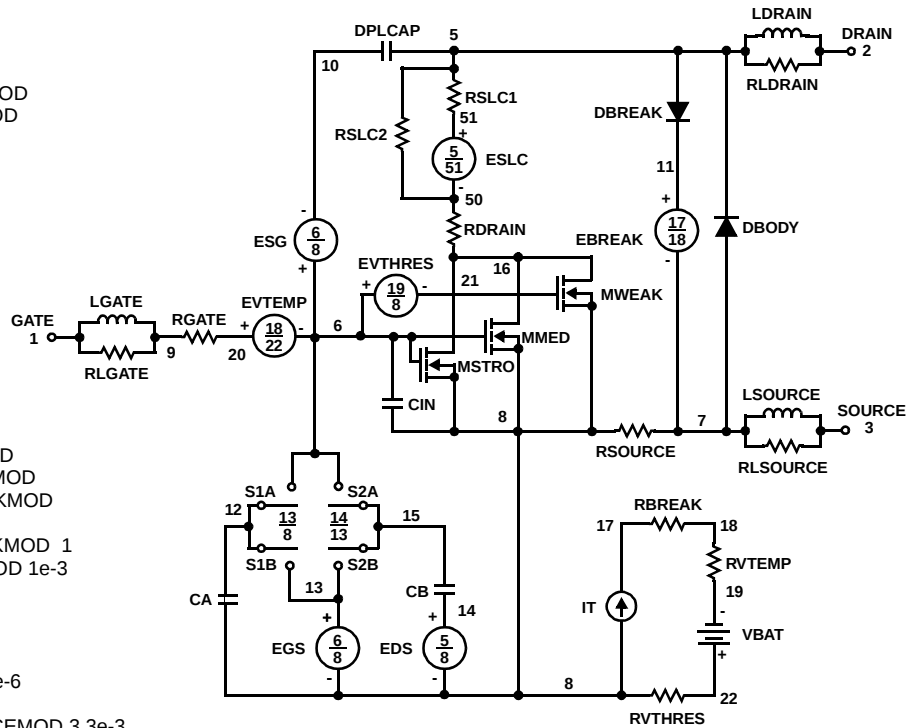
.MODEL S1BMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -2.0 VOFF= -3.0)

.MODEL S2AMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = -0.5 VOFF= 0.2)

.MODEL S2BMOD VSWITCH (RON = 1e-5 ROFF = 0.1 VON = 0.2 VOFF= -0.5)

.ENDS

Note: For further discussion of the PSPICE model, consult **A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options**; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.



SPICE Thermal Model

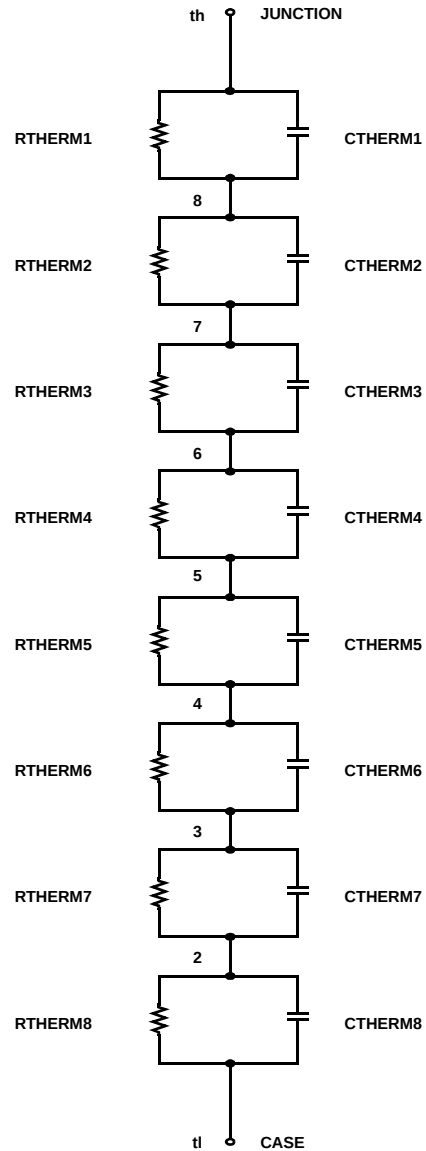
REV 18 May 2001
 ISL9N305ASK8_76T
 Copper Area = 0.76 in²
 C THERM1 th 8 2.0e-3
 C THERM2 8 7 5.0e-3
 C THERM3 7 6 1.0e-2
 C THERM4 6 5 4.0e-2
 C THERM5 5 4 9.0e-2
 C THERM6 4 3 0.2
 C THERM7 3 2 1.0
 C THERM8 2 tl 3.0

R THERM1 th 8 0.1
 R THERM2 8 7 0.5
 R THERM3 7 6 1.0
 R THERM4 6 5 5.0
 R THERM5 5 4 8.0
 R THERM6 4 3 13
 R THERM7 3 2 19
 R THERM8 2 tl 29.7

SABER Thermal Model

Copper Area = 0.76 in²
 template thermal_model th tl
 thermal_c th, tl
 {
 ctherm.ctherm1 th 8 = 2.0e-3
 ctherm.ctherm2 8 7 = 5.0e-3
 ctherm.ctherm3 7 6 = 1.0e-2
 ctherm.ctherm4 6 5 = 4.0e-2
 ctherm.ctherm5 5 4 = 9.0e-2
 ctherm.ctherm6 4 3 = 0.2
 ctherm.ctherm7 3 2 = 1.0
 ctherm.ctherm8 2 tl = 3.0

rtherm.rtherm1 th 8 = 0.1
 rtherm.rtherm2 8 7 = 0.5
 rtherm.rtherm3 7 6 = 1.0
 rtherm.rtherm4 6 5 = 5.0
 rtherm.rtherm5 5 4 = 8.0
 rtherm.rtherm6 4 3 = 13
 rtherm.rtherm7 3 2 = 19
 rtherm.rtherm8 2 tl = 29.7
 }

**TABLE 1. THERMAL MODELS**

COMPONANT	0.04 in ²	0.28 in ²	0.52 in ²	0.76 in ²	1.0 in ²
C THERM6	1.2e-1	1.5e-1	2.0e-1	2.0e-1	2.0e-1
C THERM7	0.5	1.0	1.0	1.0	1.0
C THERM8	1.3	2.8	3.0	3.0	3.0
R THERM6	26	20	15	13	12
R THERM7	39	24	21	19	18
R THERM8	55	38.7	31.3	29.7	25

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ActiveArray™	FACT Quiet Series™	ISOPLANAR™	POP™	Stealth™
Bottomless™	FAST®	LittleFET™	Power247™	SuperSOT™-3
CoolFET™	FAST _r ™	MicroFET™	PowerTrench®	SuperSOT™-6
CROSSVOLT™	FRFET™	MicroPak™	QFET™	SuperSOT™-8
DOVE™	GlobalOptoisolator™	MICROWIRE™	QS™	SyncFET™
EcoSPARK™	GTO™	MSX™	QT Optoelectronics™	TinyLogic™
E ² CMOS™	HiSeC™	MSXPro™	Quiet Series™	TruTranslation™
EnSigna™	I ² C™	OCX™	RapidConfigure™	UHC™
Across the board. Around the world.™		OCXPro™	RapidConnect™	UltraFET®
The Power Franchise™		OPTOLOGIC®	SILENT SWITCHER®	VCX™
Programmable Active Droop™		OPTOPLANAR™	SMART START™	

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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

PRODUCT STATUS DEFINITIONS

Definition of Terms

Datasheet Identification	Product Status	Definition
Advance Information	Formative or In Design	This datasheet contains the design specifications for product development. Specifications may change in any manner without notice.
Preliminary	First Production	This datasheet contains preliminary data, and supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice in order to improve design.
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