

Differential Output Video Dot Clock Generator

General Description

The ICS1567 is a very high performance monolithic PLL frequency synthesizer. Utilizing ICS's advanced CMOS mixed-mode technology, the ICS1567 provides a low cost solution for high-end video clock generation, and for telecom system clock generation.

The ICS1567 has differential video clock outputs (CLK and CLK) that are compatible with industry standard video DACs & RAMDACs. An additional clock output, LD, is provided, whose frequency is divided down from the main clock by a programmable divider.

Operating frequencies are selectable from a pre-programmed (customer-defined) table. An on-chip crystal oscillator for generating the reference frequency is provided on the ICS1567.

Programming of the ICS1567 is accomplished via frequency select pins on the package. The ICS1567 has five lines plus a STROBE pin which permits selection of 32 frequencies. Reset of the pipeline delay on Brooktree RAMDACs is automatically performed on a rising edge of the STROBE line.

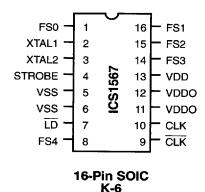
Features

- High frequency operation for extended video modes up to 180 MHz
- Compatible with Brooktree high performance RAMDACs
 - a) Differential output clocks with ECL logic levels
 - Programmable divider modulus for load clock
 - c) Circuitry included for automatic reset of Brooktree RAMDAC pipeline delay
- Low cost eliminates need for multiple ECL crystal clock oscillators in video display systems
- Strobed/Transparent frequency select options
- 32-user selected mask-programmable frequencies
- Fast acquisition of selected frequencies, strobed or nonstrobed
- · Advanced PLL for low phase-jitter
- Dynamic control of VCO sensitivity providing optimized loop gain over entire frequency range
- Small footprint 16-pin wide body (300 mil) SOIC

Applications

- Workstations
- High-resolution PC and MAC displays
- 8514A TMS340X0 systems
- EGA VGA Super VGA video
- Telecom reference clock generation suitable for Sonet, ATM and other data rates up to 155.52Mb.

Pin Configuration



ICS1567RevB090894

4825758 0001471 369

E-51



Block Diagram

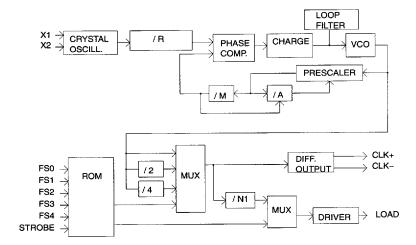


Figure 1

System Schematic

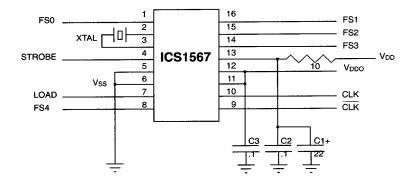
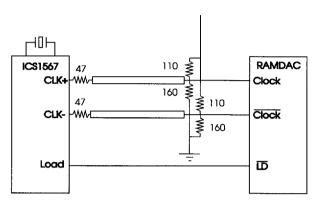


Figure 2



Typical Output Configuration





Notes:

CLK & $\overline{\text{CLK}}$ outputs are pseudo-ECL. Logic low level is set by the ratio of the resistors stacked across the power supply $V_{LO} = (V \text{ supply } \bullet 160)/(110 + 160)$ in the example shown above.

The above values are a good starting point for RAMDAC or clock generator interface.

Figure 3

Pin Description

PIN NUMBER	PIN SYMBOL	TYPE	DESCRIPTION	
1	• FS0	IN	Frequency Select LSB	
2	XTAL1	IN	Crystal Interface/External Oscillator Input	
3	XTAL2	OUT	Crystal Interface	
4	• STROBE	IN	Control For Frequency Select Latch, also performs automatic RAMDAC reset	
5	VSS		Device Ground (Both pins must be connected.)	
6	VSS		Device Ground (Both pins must be connected.)	
7	LD	OUT	Load Output. This output is at CLK frequency divided by N1.	
8	• FS4	IN	Frequency Select MSB	
9	CLK	OUT	Clock Output Inverted	
10	CLK	OUT	Clock Output Non-Inverted	
11	VDDO		Output Stage Power (Both pins must be connected)	
12	VDDO		Output Stage Power (Both pins must be connected)	
13	VDD		PLL System Power	
14	• FS3	IN	Frequency Select	
15	• FS2	IN	Frequency Select	
16	• FS1	IN	Frequency Select	

^{• =} inputs with internal pull-up resistor



Circuit Description

Overview

The ICS1567 is designed to provide the graphics system clock signals required by industry standard RAMDACs. One of 32 pre-programmed (user-definable) frequencies may be selected under digital control. Fully programmable feedback and reference divider capability allow virtually any frequency to be generated, not just simple multiples of the reference frequency. The ICS1567 uses the latest generation of frequency synthesis techniques developed by ICS and is completely suitable for the most demanding video applications.

Digital Inputs

The FS0-FS4 pins and the STROBE pin are used to select the desired operating frequency from the 32 pre-programmed frequencies in the ROM table of the ICS1567. The STROBE pin also controls activation of the pipeline delay RESET function included in the ICS1567 (see PIPELINE DELAY RESET section for details). The FS0-FS4 and STROBE pins are each equipped with a pull-up and will be at a logic HIGH level when not connected.

Transparent Mode - When the STROBE pin is held HIGH, the FS0 through FS4 inputs are transparent; that is, they directly access the ROM table. The synthesizer will output the frequency programmed into the location addressed by the FS0-FS4 pins.

Latched Mode - When the STROBE pin is held LOW, the FS0-FS4 pins are ignored. The synthesizer will output the frequency corresponding to the state of the FS0-FS4 pins when the STROBE pin was last HIGH. In the event that the ICS1567 is powered-up with the STROBE pin held LOW, the synthesizer will output the frequency programmed into address 0 (i.e., the one selected with FS0 through FS4 at a logic LOW level).

Frequency Synthesizer Description

Refer to Figure 1 for a block diagram of the ICS1567. The reference frequency is generated by an on-chip crystal oscillator, or the reference frequency may be applied to the ICS1567 from an external frequency source.

The ICS1567 generates its output frequencies using phase-locked loop techniques. The phase-locked loop (or PLL) is a closed-loop feedback system that drives the output frequency to be ratiometrically related to the reference frequency provided to the PLL. The phase-frequency detector shown in the block diagram drives the VCO to a frequency that will cause the two inputs to the phase-frequency detector to be matched in frequency and phase. This occurs when:

$$F(vco) = \frac{F(XTAL1) \bullet Feedback Divider}{Reference Divider}$$

This expression is exact; that is, the accuracy of the output frequency depends solely on the reference frequency provided to the part (assuming correctly-programmed dividers). The divider programming is one of the functions performed by the ROM look-up table in the ICS1567. The VCO gain is also ROM programmable which permits the ICS1567 to be optimized for best performance at each frequency in the table.

The feedback divider makes use of a dual-modulus prescaler technique that allows construction of a programmable counter to operate at high speeds while still allowing the feedback divider to be programmed in steps of 1. This is an improvement over conventional fixed prescaler architectures that typically impose a factor-of-four penalty (or larger) in this respect.

A post-divider may be inserted between the VCO and the CLK and CLK outputs of the ICS1567. This is useful in generation of lower frequencies, as the VCO has been optimized for high-frequency operation. Different post-divider settings may be used for each frequency in the table.



Load Clock Divider

The ICS1567 has an additional programmable divider that is used to generate the LOAD frequency. The modulus of this divider may be set to 3, 4, 5, 6, 8, or 10. The design of this divider permits the output duty factor to be 50/50, even when an odd modulus is selected.

The selection of the modulus is done by the ROM look-up table. A different modulus may, therefore, be selected for each frequency address.

Pipeline Delay Reset Function

The ICS1567 implements the clocking sequence required to reset the pipeline delay on Brooktree RAMDACs. This sequence is automatically generated by the ICS1567 upon any rising edge of the STROBE line.

When the frequency select inputs (FS0-FS4) are used in a transparent mode, simply lower and raise the STROBE line to activate the function. When the frequency select inputs are latched, simply load the same frequency into the ICS1567 twice.

When changing frequencies, it is advisable to allow 500uSec after the new frequency is selected to activate the reset function. The output frequency of the synthesizer should be stable enough at that point for the RAMDAC to correctly execute its reset sequence.

See Figure 4 for a diagram of the clock sequencing.

Output Stage Description

The CLK and CLK outputs are each connected to the drains of P-Channel MOSFET devices. The source of each of these devices is connected to VDDO. Typical on resistance of each device is 15 Ohms. These outputs will drive the clock and clock of a RAMDAC device when a resistive network equivalent to Figure 3 is utilized.

The LD output is a high-current CMOS type drive whose frequency is controlled by a programmable divider that may be selected for a modulus of 3, 4, 5, 6, 8, or 10. Under control of the ROM, this output may also be suppressed (logic low level) at any frequency select address, if desired.

Application Information

Power Supplies

The ICS1567 has two VSS pins to reduce the effects of package inductance. Both pins are connected to the same potential on the die (the ground bus). BOTH of these pins should connect to the ground plane of the video board as close to the package as is possible.

The ICS1567 has two VDDO pins which are the supply of +5 volt power to all output stages. Again, both VDDO pins connect to the same point on the die. BOTH of these pins should be connected to the power plane (or bus) using standard high-frequency decoupling practice. This decoupling consists of a low series inductance bypass capacitor, using the shortest leads possible, mounted close to the ICS1567.

The VDD pin is the power supply for the synthesizer circuitry and other lower current digital functions. We recommend that RC decoupling or zener regulation be provided for this pin (as shown in the recommended application circuitry). This will allow the PLL to "track" through power supply fluctuations without visible effects.

Crystal Oscillator and Crystal Selection

The ICS1567 has circuitry on-board to implement a Pierce oscillator with the addition of only one external component, a quartz crystal. Pierce oscillators operate the crystal in anticalso called parallel-) resonant mode. See the AC Characteristics for the effective capacitive loading to specify when ordering crystals.

So-called series-resonant crystals may also be used with the ICS1567. Be aware that the oscillation frequency will be slightly higher than the frequency that is stamped on the can (typically 0.005-0.01%).

As the entire operation of the phase-locked loop depends on having a stable reference frequency, we recommend that the crystal be mounted as closely as possible to the package. Avoid routing digital signals or the ICS1567 outputs underneath or near these traces. It is also desirable to ground the crystal can to the ground plane, if possible.



Application Notes (continued) Bus Clock Interface

In some applications, it may be desirable to utilize the bus clock. To do this, connect the clock through a .047uF capacitor to XTAL1 (2) and keep the lead length of the capacitor to XTAL1 (2) to a minimum to reduce noise susceptibility. This input is internally biased at VDD/2. Since TTL compatible clocks typically exhibit a VOH of 3.5V, capacitively coupling the input restores noise immunity. The ICS1567 is not sensitive to the duty cycle of the bus clock; however, the quality of this signal varies considerably with different motherboard designs. As the quality of the bus clock is typically outside the control of the graphics adapter card manufacturer, it is suggested that this signal be buffered on the graphics adapter board. XTAL2 (3) must be left open in this configuration.

ICS1567 Interface

The ICS1567 should be located as close as possible to the video DAC or RAMDAC. Figure 3 illustrates interfacing the ICS1567 to a RAMDAC. The differential output CLOCK drivers are current sourcing only and are designed to drive resistive terminations in a complementary fashion CLK and CLK connections should follow good ECL interconnection practice. Terminating resistors should be as close as possible to the RAMDAC.

Absolute Maximum Ratings

Ambient Temeperature under bias	$T_0 \dots \dots$	0°C to 70°C
Supply Voltage	V_{DD}	-0.5V to +7V
Input Voltage	V_{IN}	$-0.5V$ to $V_{DD} + 0.5V$
Output Voltage	V _{OUT}	$-0.5V$ to $V_{DD} + 0.5V$
Clamp Diode Current		
Output Current per Pin	IOUT	±50mA
Storage Temperature	$T_S\dots\dots\dots$	-85°C to + 150°C
Power Dissipation	P_D	500mW

Values beyond these ratings may damage the device. This device contains circuitry to protect the inputs and outputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid applications of any voltage higher than the maximum rated voltages. For proper operation, it is recommended that V_{IN} and V_{OUT} be constrained to $> = V_{SS}$ and $< = V_{DD}$.

Standard Test Conditions

The characteristics below apply for the following standard test conditions, unless otherwise noted. All voltages are referenced to VSS (OV Ground). Positive current flows into the referenced pin.

Operating Temperature range	0°C to 70°C
Power supply voltage	4.75 to 5.25 Volts



DC Characteristics

PARAMETER	SYMBOL	MIN	MAX	UNITS	CONDITIONS
Input High Voltage	VIH	2.0	$V_{DD} + 0.5$	V	
Input Low Voltage	V _{IL}	Vss -0.5	0.8	V	
Input High Current	IIH		10	uA	$V_{IN} = V_{DD}$
Input Low Current	IIL		-200	uA	$V_{IN} = V_{SS}$
		LO	AD OUTPUT		
Output High Voltage	V _{OH}	2.4		V	I _{OH} = -4.0 mA
Output Low Voltage	Vol		0.4	V	$I_{OL} = 6.0 \text{ mA}$
	N 4	CLO	CK OUTPUTS		
Differential Output Voltage (CLK-CLK)	V _{OD}	1.2		V	See Figure 4
XTALI INPUT					
Input High Voltage	V _{XH}	3.75	$V_{DD} + 0.5$	V	
Input Low Voltage	V_{XL}	VSS -0.5	1.25	V	
Operating Current	I_{DD}		50	mA	Outputs Unloaded
Input Pin Capacitance	Cin		8	pF	$F_C = 1 \text{ MHz}$
Output Pin Capacitance	Cout		12	pF	$F_C = 1 \text{ MHz}$



AC Characteristics

PARAMETER	SYMBOL	MIN	TYP.	MAX	UNITS	NOTES
	CI	K and CLK T	IMING			
Duty Cycle	THIGH	40		60	%	3, 4, 9
Frequency Error				0.5	%	
Rise Time	Tr			2	ns	5, 9
Fall Time	T_{f}			2	ns	5, 9
VCO Frequency	F _{VCO}	20		180	MHz	1
PLL Acquire Time	T _{LOCK}		500		uS	
<u> </u>	1 1 20 20	LD* TIMIN	īG	1	3 20 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	
Duty Cycle	T _{HIGH}	40		60	%	6
Load Frequency	FLOAD			60	MHz	
Rise Time	T_r			2	ns	7, 8
Fall Time	$T_{\mathbf{f}}$			2	ns	7, 8
	REFE	RENCE INPL	T CLOCK			
Crystal Frequency	FXTAL	5		20	MHz	
Crystal Oscillator	CPAR		20		pF	ĺ
Loading Capacitance						
XTAL1 High Time	T _{XHI}	8			ns	2
XTAL1 Low Time	T _{XLO}	8		1	ns	2
Rise Time	Tr			10	ns	2, 7
Fall Time	T_{f}			10	ns	2, 7
		DIGITAL INF	UTS			
Frequency Select Setup Time	1	10			ns	10
Frequency Select Hold Time	2	10			ns	10
Strobe Pulse Width	3	20			ns	10
	PIPI	ELINE DELA	RESET			
Reset Activation	4			2*T _{CLK}	ns	10
Reset Duration	5	4*T _{CLK}			ns	10
Restart Delay	6	-1*Tclk		+1.5*T _{CLK}	ns	10

Notes:

- Use of the post-divider is required for frequencies lower than 20 MHz on CLK and CLK outputs. Use of the
 post-divider is recommended for output frequencies lower than 65 MHz.
- 2. Values for XTAL1 driven by an external clock
- 3. Duty Cycle for Differential Output (CLK- CLK)
- 4. Duty cycle measured at VOD/2 for Differential CLK Output
- 5. Rise and fall time between 20% and 80% of VOD
- 6. Duty cycle measured at 1.4V for TTL I/O
- 7. Rise and fall time between 0.8 and 2.0 VDC for TTL I/O
- 8. Output pin loading = 15 pf
- 9. See Figure 3.
- 10. See Figure 4.



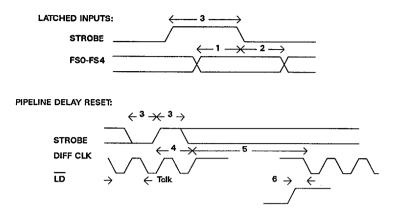


Figure 4



ICS1567 Pattern Request Form

Custom patterns are also available, although a significant volume commitment and/or one-time mask charge will apply. Contact ICS Sales for details.

ICS Part Number	ICS1567- 742	ICS1567- Custom Pattern #1		
Video Clock Address (HEX)	Frequency (MHz)	Frequency (MHz)		
0	112.000			
1	148.000			
2	OFF			
3	135.000			
4	31.500			
5	105.500			
6	78.000			
7	86.000			
8	108.000			
9	120.000			
10	128.000			
11	93.000			
12	112.000			
13	148.000			
14	135.000			
15	89.210			
16	105.500			
17	112.000			
18	25.000			
19	45.000			
20	64.000			
21	75.000			
22	78.000			
23	86.000			
24	103.000			
25	108.000			
26	120.000			
27	127.000			
28	128.000			
29	135,000			
30	112.000			
31	148.000			

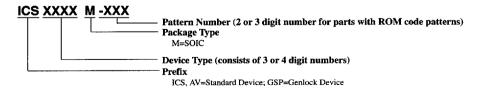
Custom pattern #1 reference frequency = _____

Standard pattern shown above uses 16.000 MHz as the input reference frequency.

Ordering Information

ICS1567M-XXX

Example:



E-60