# 100 mA, 3.0 V, Low Power Low Dropout Voltage Regulator

The LP2950 and LP2951 are micropower voltage regulators that are specifically designed to maintain proper regulation with an extremely low input–to–output voltage differential. These devices feature a very low quiescent bias current of 75  $\mu A$  and are capable of supplying output currents in excess of 100 mA. Internal current and thermal limiting protection is provided.

The LP2951 has three additional features. The first is the Error Output that can be used to signal external circuitry of an out of regulation condition, or as a microprocessor power—on reset. The second feature allows the output voltage to be preset to 5.0 V, 3.3 V or 3.0 V output (depending on the version) or programmed from 1.25 V to 29 V. It consists of a pinned out resistor divider along with direct access to the Error Amplifier feedback input. The third feature is a Shutdown input that allows a logic level signal to turn—off or turn—on the regulator output.

Due to the low input-to-output voltage differential and bias current specifications, these devices are ideally suited for battery powered computer, consumer, and industrial equipment where an extension of useful battery life is desirable. The LP2950 is available in the three pin case 29 and DPAK packages, and the LP2951 is available in the eight pin dual-in-line, SO-8 and Micro-8 surface mount packages. The 'A' suffix devices feature an initial output voltage tolerance  $\pm 0.5\%$ .

#### LP2950 and LP2951 Features:

- Low Quiescent Bias Current of 75 μA
- $\bullet$  Low Input–to–Output Voltage Differential of 50 mV at 100  $\mu A$  and 380 mV at 100 mA
- 5.0 V, 3.3 V or 3.0 V  $\pm$ 0.5% Allows Use as a Regulator or Reference
- Extremely Tight Line and Load Regulation
- Requires Only a 1.0 μF Output Capacitor for Stability
- Internal Current and Thermal Limiting

#### LP2951 Additional Features:

- Error Output Signals an Out of Regulation Condition
- Output Programmable from 1.25 V to 29 V
- Logic Level Shutdown Input

(See Following Page for Device Information.)



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TO-92 Z SUFFIX CASE 29



Pin: 1. Output 2. Ground

3. Input

DPAK DT SUFFIX CASE 369A



#### PIN CONNECTIONS



Pin: 1. Input

2. Ground

3. Output

Heatsink surface (shown as terminal 4 in case outline drawing) is connected to Pin 2.

SO-8 D SUFFIX

D SUFFIX CASE 751



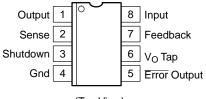
PDIP-8 N SUFFIX CASE 626



Micro-8 DM SUFFIX CASE 846A



#### **PIN CONNECTIONS**



(Top View)

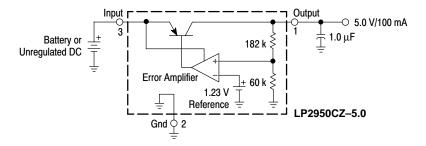
#### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on pages 13 and 14 of this data sheet.

#### **DEVICE INFORMATION**

		Output Voltage				
Package	3.0V	3.3V	5.0V	Adjustable	Operating Junction Temperature Range	
TO-92	LP2950CZ-3.0	LP2950CZ-3.3	LP2950CZ-5.0	Not	$T_{J} = -40^{\circ} \text{ to } +125^{\circ}\text{C}$	
Suffix Z	LP2950ACZ-3.0	LP2950ACZ-3.3	LP2950ACZ-5.0	Available		
DPAK	LP2950CDT-3.0	LP2950CDT-3.3	LP2950CDT-5.0	Not	$T_{J} = -40^{\circ} \text{ to } +125^{\circ}\text{C}$	
Suffix DT	LP2950ACDT-3.0	LP2950ACDT-3.3	LP2950ACDT-5.0	Available		
SO-8	LP2951CD-3.0	LP2951CD-3.3	LP2951CD	LP2951CD	$T_{J} = -40^{\circ} \text{ to } +125^{\circ}\text{C}$	
Suffix D	LP2951ACD-3.0	LP2951ACD-3.3	LP2951ACD	LP2951ACD		
Micro–8	LP2951CDM-3.0	LP2951CDM-3.3	LP2951CDM	LP2951CDM	$T_J = -40^{\circ} \text{ to } +125^{\circ}\text{C}$	
Suffix DM	LP2951ACDM-3.0	LP2951ACDM-3.3	LP2951ACDM	LP2951ACDM		
DIP-8	LP2951CN-3.0	LP2951CN-3.3	LP2951CN	LP2951CN	$T_J = -40^{\circ} \text{ to } +125^{\circ}\text{C}$	
Suffix N	LP2951ACN-3.0	LP2951ACN-3.3	LP2951ACN	LP2951ACN		

LP2950Cx-xx / LP2951Cxx-xx LP2950ACx-xx / LP2951ACxx-xx 1% Output Voltage Precision at  $T_J = 25^{\circ}C$  0.5% Output Voltage Precision at  $T_J = 25^{\circ}C$ 



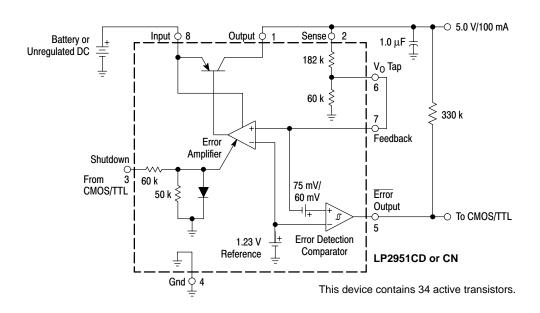


Figure 1. Representative Block Diagrams

#### **MAXIMUM RATINGS** ( $T_A = 25^{\circ}C$ , unless otherwise noted.)

Rating	Symbol	Value	Unit	
Input Voltage	V <sub>CC</sub>	30	Vdc	
Power Dissipation and Thermal Characteristics				
Maximum Power Dissipation	P <sub>D</sub>	Internally Limited	W	
Case 751(SO-8) D Suffix				
Thermal Resistance, Junction-to-Ambient	$R_{ heta JA}$	180	°C/W	
Thermal Resistance, Junction-to-Case	$R_{ heta JC}$	45	°C/W	
Case 369A (DPAK) DT Suffix (Note 1)				
Thermal Resistance, Junction-to-Ambient	$R_{ heta JA}$	92	°C/W	
Thermal Resistance, Junction-to-Case	$R_{ hetaJC}$	6.0	°C/W	
Case 29 (TO-226AA/TO-92) Z Suffix				
Thermal Resistance, Junction-to-Ambient	$R_{ heta JA}$	160	°C/W	
Thermal Resistance, Junction-to-Case	$R_{ heta JC}$	83	°C/W	
Case 626 N Suffix				
Thermal Resistance, Junction-to-Ambient	$R_{ heta JA}$	105	°C/W	
Case 846A (Micro-8) DM Suffix				
Thermal Resistance, Junction-to-Ambient	$R_{ heta JA}$	240	°C/W	
Feedback Input Voltage	$V_{fb}$	-1.5 to +30	Vdc	
Shutdown Input Voltage	$V_{sd}$	-0.3 to +30	Vdc	
Error Comparator Output Voltage	V <sub>err</sub>	-0.3 to +30	Vdc	
Operating Junction Temperature	TJ	-40 to +125	°C	
Storage Temperature Range	T <sub>stg</sub>	-65 to +150	°C	

#### **ELECTRICAL CHARACTERISTICS** ( $V_{in} = V_O + 1.0 \text{ V}, I_O = 100 \text{ }\mu\text{A}, C_O = 1.0 \text{ }\mu\text{F}, T_J = 25 ^{\circ}\text{C}$ [Note 3], unless otherwise noted.)

Characteristic	Symbol	Min	Тур	Max	Unit
Output Voltage, 5.0 V Versions	V <sub>O</sub>				V
$V_{in} = 6.0 \text{ V}, I_{O} = 100 \mu\text{A}, T_{J} = 25^{\circ}\text{C}$					
LP2950C-5.0/LP2951C		4.950	5.000	5.050	
LP2950AC-5.0/LP2951AC		4.975	5.000	5.025	
$T_{J} = -40 \text{ to } +125^{\circ}\text{C}$					
LP2950C-5.0/LP2951C		4.900	-	5.100	
LP2950AC-5.0/LP2951AC		4.940	_	5.060	
$V_{in}$ = 6.0 to 30 V, $I_{O}$ = 100 $\mu A$ to 100 mA, $T_{J}$ = -40 to +125°C					
LP2950C-5.0/LP2951C		4.880	-	5.120	
LP2950AC-5.0/LP2951AC		4.925	_	5.075	
Output Voltage, 3.3 V Versions	Vo				V
$V_{in} = 4.3 \text{ V}, I_{O} = 100 \mu\text{A}, T_{J} = 25^{\circ}\text{C}$					
LP2950C-3.3/LP2951C-3.3		3.267	3.300	3.333	
LP2950AC-3.3/LP2951AC-3.3		3.284	3.300	3.317	
$T_{J} = -40 \text{ to } +125^{\circ}\text{C}$					
LP2950C-3.3/LP2951C-3.3		3.234	_	3.366	
LP2950AC-3.3/LP2951AC-3.3		3.260	_	3.340	
$V_{in}$ = 4.3 to 30 V, $I_{O}$ = 100 $\mu A$ to 100 mA, $T_{J}$ = -40 to +125°C					
LP2950C-3.3/LP2951C-3.3		3.221	_	3.379	
LP2950AC-3.3/LP2951AC-3.3		3.254	_	3.346	
Output Voltage, 3.0 V Versions	Vo				V
$V_{in} = 4.0 \text{ V}, I_O = 100 \mu\text{A}, T_J = 25^{\circ}\text{C}$					
LP2950C-3.0/LP2951C-3.0		2.970	3.000	3.030	
LP2950AC-3.0/LP2951AC-3.0		2.985	3.000	3.015	
$T_{J} = -40 \text{ to } +125^{\circ}\text{C}$					
LP2950C-3.0/LP2951C-3.0		2.940	_	3.060	
LP2950AC-3.0/LP2951AC-3.0		2.964	_	3.036	
$V_{in}$ = 4.0 to 30 V, $I_O$ = 100 $\mu A$ to 100 mA, $T_J$ = -40 to +125°C					
LP2950C-3.0/LP2951C-3.0		2.928	_	3.072	
LP2950AC-3.0/LP2951AC-3.0		2.958		3.042	

- The Junction-to-Ambient Thermal Resistance is determined by PC board copper area per Figure 27.
   ESD data available upon request.
- Low duty pulse techniques are used during test to maintain junction temperature as close to ambient as possible.
   V<sub>O(nom)</sub> is the part number voltage option.
   Noise tests on the LP2951 are made with a 0.01 μF capacitor connected across Pins 7 and 1.

 $\textbf{ELECTRICAL CHARACTERISTICS (continued)} \quad (V_{in} \ = \ V_O \ + \ 1.0 \ V, \ I_O \ = \ 100 \ \mu\text{A}, \ C_O \ = \ 1.0 \ \mu\text{F}, \ T_J \ = \ 25^{\circ}\text{C} \qquad [Note \ 8], \ unless \ = \ 1.0 \ \mu\text{A}, \ C_O \ = \ 1.0 \ \mu\text{F}, \ T_J \ = \ 25^{\circ}\text{C}$ otherwise noted.)

Characteristic	Symbol	Min	Тур	Max	Unit
Line Regulation (V <sub>in</sub> = V <sub>O(nom)</sub> +1.0 V to 30 V) (Note 9)	Reg <sub>line</sub>				%
LP2950C-XX/LP2951C/LP2951C-XX		_	0.08	0.20	
LP2950AC-XX/LP2951AC/LP2951AC-XX		-	0.04	0.10	
Load Regulation (I <sub>O</sub> = 100 μA to 100 mA)	Reg <sub>load</sub>				%
LP2950C-XX/LP2951C/LP2951C-XX		_	0.13	0.20	
LP2950AC-XX/LP2951AC/LP2951AC-XX		_	0.05	0.10	
Dropout Voltage	$V_I - V_O$				mV
$I_{O} = 100 \mu\text{A}$		_	30	80	
I <sub>O</sub> = 100 mA		_	350	450	
Supply Bias Current	I <sub>CC</sub>				
$I_{O} = 100 \mu\text{A}$		_	93	120	μΑ
I <sub>O</sub> = 100 mA		_	4.0	12	mA
Dropout Supply Bias Current ( $V_{in}$ = $V_{O(nom)}$ – 0.5 V, $I_{O}$ = 100 $\mu$ A) (Note 9)	I <sub>CCdropout</sub>	_	110	170	μΑ
Current Limit (V <sub>O</sub> Shorted to Ground)	l <sub>Limit</sub>	_	220	300	mA
Thermal Regulation	Reg <sub>thermal</sub>	_	0.05	0.20	%/W
Output Noise Voltage (10 Hz to 100 kHz) (Note 10)	V <sub>n</sub>				μVrms
$C_L = 1.0 \mu\text{F}$		_	126	_	
$C_L = 100  \mu F$		_	56	_	
LP2951A/LP2951AC ONLY					
Reference Voltage $(T_J = 25^{\circ}C)$	$V_{ref}$				V
LP2951C/LP2951C-XX		1.210	1.235	1.260	
LP2951AC/LP2951AC-XX		1.220	1.235	1.250	
Reference Voltage ( $T_J = -40 \text{ to } +125^{\circ}\text{C}$ )	$V_{ref}$				V
LP2951C/LP2951C-XX		1.200	_	1.270	
LP2951AC/LP2951AC-XX		1.200	_	1.260	
Reference Voltage ( $T_J = -40 \text{ to } +125^{\circ}\text{C}$ )	$V_{ref}$				V
$I_0 = 100 \mu\text{A}$ to 100 mA, $V_{in} = 23$ to 30 V					
LP2951C/LP2951C-XX		1.185	_	1.285	
LP2951AC/LP2951AC—XX		1.190		1.270	
Feedback Pin Bias Current	I <sub>FB</sub>	_	15	40	nA
ERROR COMPARATOR	T .	I		ı	1 .
Output Leakage Current (V <sub>OH</sub> = 30 V)	I <sub>lkg</sub>	_	0.01	1.0	μΑ
Output Low Voltage ( $V_{in} = 4.5 \text{ V}$ , $I_{OL} = 400 \mu\text{A}$ )	V <sub>OL</sub>	_	150	250	mV
Upper Threshold Voltage (V <sub>in</sub> = 6.0 V)	$V_{thu}$	40	45	-	mV
Lower Threshold Voltage (V <sub>in</sub> = 6.0 V)	$V_{thl}$	-	60	95	mV
Hysteresis (V <sub>in</sub> = 6.0 V)	$V_{hy}$	_	15	-	mV
SHUTDOWN INPUT	•	T		1	T
Input Logic Voltage	$V_{shtdn}$				V
Logic "0" (Regulator "On")		0	_	0.7	
Logic "1" (Regulator "Off")		2.0	_	30	
Shutdown Pin Input Current	I <sub>shtdn</sub>				μΑ
$V_{\text{shtdn}} = 2.4 \text{ V}$		_	35	50	
V <sub>shtdn</sub> = 30 V		_	450	600	_
Regulator Output Current in Shutdown Mode	I <sub>off</sub>	_	3.0	10	μΑ
(V <sub>in</sub> = 30 V, V <sub>shtdn</sub> = 2.0 V, V <sub>O</sub> = 0, Pin 6 Connected to Pin 7)					

The Junction—to–Ambient Thermal Resistance is determined by PC board copper area per Figure 27.
 ESD data available upon request.
 Low duty pulse techniques are used during test to maintain junction temperature as close to ambient as possible.
 V<sub>O(nom)</sub> is the part number voltage option.
 Noise tests on the LP2951 are made with a 0.01 μF capacitor connected across Pins 7 and 1.

#### **DEFINITIONS**

**Dropout Voltage** – The input/output voltage differential at which the regulator output no longer maintains regulation against further reductions in input voltage. Measured when the output drops 100 mV below its nominal value (which is measured at 1.0 V differential), dropout voltage is affected by junction temperature, load current and minimum input supply requirements.

**Line Regulation** – The change in output voltage for a change in input voltage. The measurement is made under conditions of low dissipation or by using pulse techniques such that average chip temperature is not significantly affected.

**Load Regulation** – The change in output voltage for a change in load current at constant chip temperature.

**Maximum Power Dissipation** – The maximum total device dissipation for which the regulator will operate within specifications.

**Bias Current** – Current which is used to operate the regulator chip and is not delivered to the load.

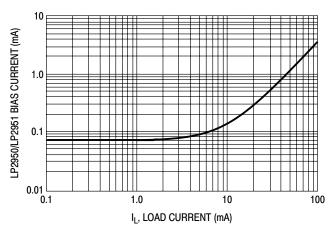


Figure 2. Quiescent Current

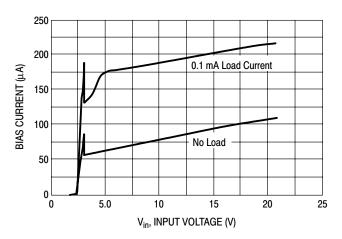


Figure 4. Input Current

**Output Noise Voltage** – The rms ac voltage at the output, with constant load and no input ripple, measured over a specified frequency range.

**Leakage Current** – Current drawn through a bipolar transistor collector–base junction, under a specified collector voltage, when the transistor is "off".

**Upper Threshold Voltage** – Voltage applied to the comparator input terminal, below the reference voltage which is applied to the other comparator input terminal, which causes the comparator output to change state from a logic "0" to "1".

**Lower Threshold Voltage** – Voltage applied to the comparator input terminal, below the reference voltage which is applied to the other comparator input terminal, which causes the comparator output to change state from a logic "1" to "0".

**Hysteresis** – The difference between Lower Threshold voltage and Upper Threshold voltage.

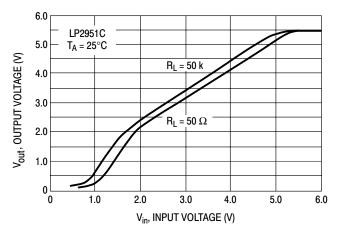


Figure 3. Dropout Characteristics

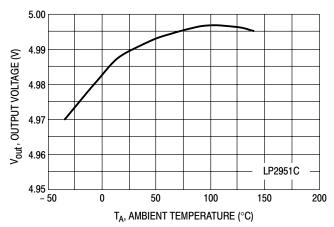


Figure 5. Output Voltage versus Temperature

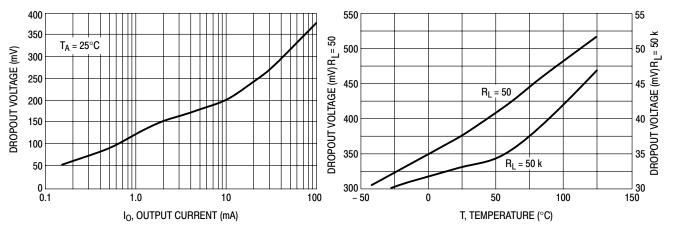
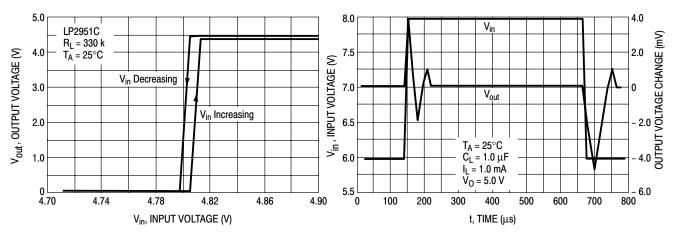


Figure 6. Dropout Voltage versus Output Current

Figure 7. Dropout Voltage versus Temperature



**Figure 8. Error Comparator Output** 

**Figure 9. Line Transient Response** 

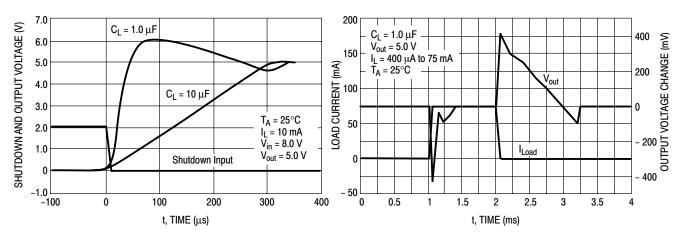


Figure 10. LP2951 Enable Transient

Figure 11. Load Transient Response

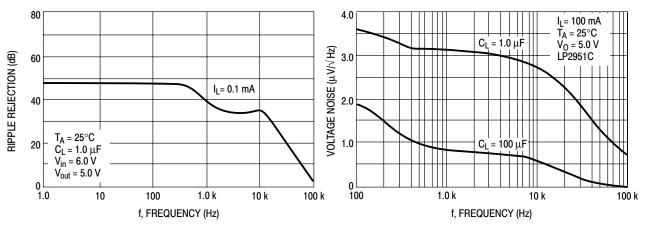


Figure 12. Ripple Rejection

Figure 13. Output Noise

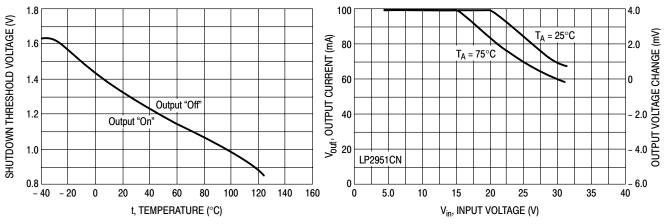


Figure 14. Shutdown Threshold Voltage versus Temperature

Figure 15. Maximum Rated Output Current

#### **APPLICATIONS INFORMATION**

#### Introduction

The LP2950/LP2951 regulators are designed with internal current limiting and thermal shutdown making them user–friendly. Typical application circuits for the LP2950 and LP2951 are shown in Figures 18 through 26.

These regulators are not internally compensated and thus require a 1.0  $\mu F$  (or greater) capacitance between the LP2950/LP2951 output terminal and ground for stability. Most types of aluminum, tantalum or multilayer ceramic will perform adequately. Solid tantalums or appropriate multilayer ceramic capacitors are recommended for operation below 25°C.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to  $0.33~\mu F$  for currents less than 10~mA, or  $0.1~\mu F$  for currents below 1.0~mA. Using the 8–pin versions at voltages less than 5.0~V operates the error amplifier at lower values of gain, so that more output capacitance is needed for stability. For the worst case operating condition of a 100~mA load at 1.23~V output (Output Pin 1 connected to the feedback Pin 7) a minimum capacitance of  $3.3~\mu F$  is recommended.

The LP2950 will remain stable and in regulation when operated with no output load. When setting the output voltage of the LP2951 with external resistors, the resistance values should be chosen to draw a minimum of  $1.0~\mu A$ .

A bypass capacitor is recommended across the LP2950/LP2951 input to ground if more than 4 inches of wire connects the input to either a battery or power supply filter capacitor.

Input capacitance at the LP2951 Feedback Pin 7 can create a pole, causing instability if high value external resistors are used to set the output voltage. Adding a 100 pF capacitor between the Output Pin 1 and the Feedback Pin 7 and increasing the output filter capacitor to at least 3.3  $\mu F$  will stabilize the feedback loop.

#### **Error Detection Comparator**

The comparator switches to a positive logic low whenever the LP2951 output voltage falls more than approximately 5.0% out of regulation. This value is the comparator's designed—in offset voltage of 60 mV divided by the 1.235 V internal reference. As shown in the representative block diagram. This trip level remains 5.0% below normal regardless of the value of regulated output voltage. For example, the error flag trip level is 4.75 V for a normal 5.0 V regulated output, or 9.50 V for a 10 V output voltage.

Figure 2 is a timing diagram which shows the  $\overline{\text{ERROR}}$  signal and the regulated output voltage as the input voltage

to the LP2951 is ramped up and down. The ERROR signal becomes valid (low) at about 1.3 V input. It goes high when the input reaches about 5.0 V (V<sub>out</sub> exceeds about 4.75 V). Since the LP2951's dropout voltage is dependent upon the load current (refer to the curve in the Typical Performance Characteristics), the input voltage trip point will vary with load current. The output voltage trip point does not vary with load.

The error comparator output is an open collector which requires an external pull–up resistor. This resistor may be returned to the output or some other voltage within the system. The resistance value should be chosen to be consistent with the 400  $\mu A$  sink capability of the error comparator. A value between 100 k and 1.0  $M\Omega$  is suggested. No pull–up resistance is required if this output is unused.

When operated in the shutdown mode, the error comparator output will go high if it has been pulled up to an external supply. To avoid this invalid response, the error comparator output should be pulled up to  $V_{out}$  (see Figure 16).

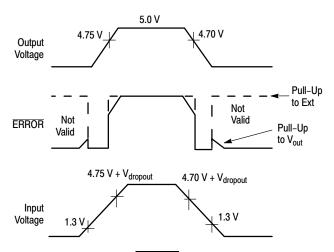


Figure 16. ERROR Output Timing

#### Programming the Output Voltage (LP2951)

The LP2951CX may be pin-strapped for the nominal fixed output voltage using its internal voltage divider by tying Pin 1 (output) to Pin 2 (sense) and Pin 7 (feedback) to Pin 6 (5.0 V tap). Alternatively, it may be programmed for any output voltage between its 1.235 reference voltage and its 30 V maximum rating. An external pair of resistors is required, as shown in Figure 17.

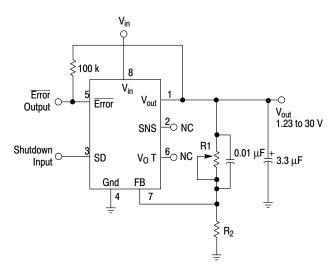


Figure 17. Adjustable Regulator

The complete equation for the output voltage is:

$$V_{out} = V_{ref} (1 + R1/R2) + I_{FB} R1$$

where  $V_{ref}$  is the nominal 1.235 V reference voltage and  $I_{FB}$  is the feedback pin bias current, nominally -20 nA. The minimum recommended load current of 1.0  $\mu$ A forces an upper limit of 1.2 M $\Omega$  on the value of R2, if the regulator must work with no load.  $I_{FB}$  will produce a 2% typical error in  $V_{out}$  which may be eliminated at room temperature by adjusting R1. For better accuracy, choosing R2 = 100 k reduces this error to 0.17% while increasing the resistor program current to 12  $\mu$ A. Since the LP2951 typically draws 75  $\mu$ A at no load with Pin 2 open circuited, the extra 12  $\mu$ A of current drawn is often a worthwhile tradeoff for eliminating the need to set output voltage in test.

#### **Output Noise**

In many applications it is desirable to reduce the noise present at the output. Reducing the regulator bandwidth by increasing the size of the output capacitor is the only method for reducing noise on the 3 lead LP2950. However, increasing the capacitor from 1.0  $\mu F$  to 220  $\mu F$  only decreases the noise from 430  $\mu V$  to 160  $\mu V$ rms for a 100 kHz bandwidth at the 5.0 V output.

Noise can be reduced fourfold by a bypass capacitor across R1, since it reduces the high frequency gain from 4 to unity. Pick

$$C_{Bypass} \approx \frac{1}{2\pi R1 \times 200 \text{ Hz}}$$

or about 0.01  $\mu$ F. When doing this, the output capacitor must be increased to 3.3  $\mu$ F to maintain stability. These changes reduce the output noise from 430  $\mu$ V to 126  $\mu$ Vrms for a 100 kHz bandwidth at 5.0 V output. With bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

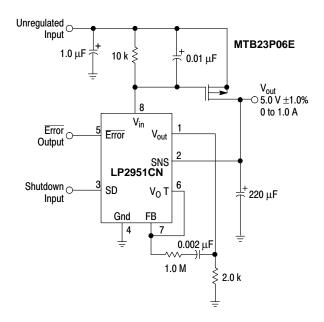


Figure 18. 1.0 A Regulator with 1.2 V Dropout

#### **TYPICAL APPLICATIONS**

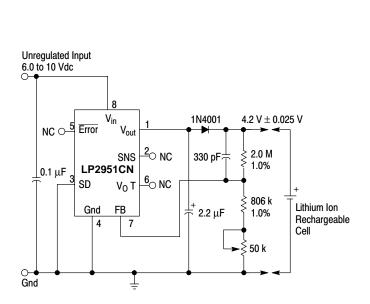


Figure 19. Lithium Ion Battery Cell Charger

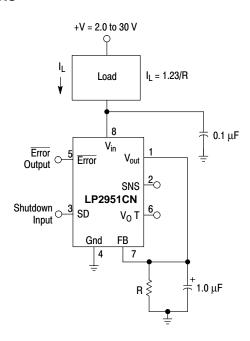


Figure 20. Low Drift Current Sink

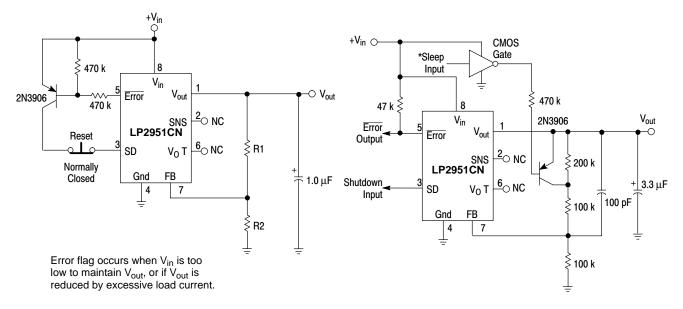
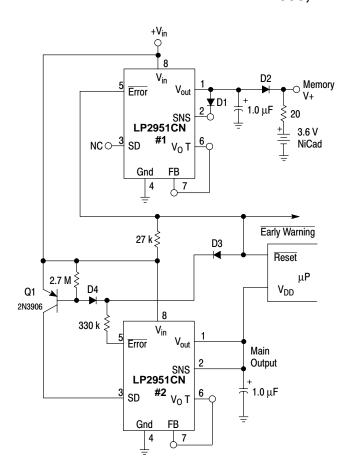


Figure 21. Latch Off When Error Flag Occurs

Figure 22. 5.0 V Regulator with 2.5 V Sleep Function



All diodes are 1N4148.

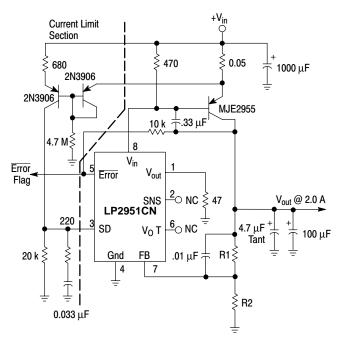
Early Warning flag on low input voltage.

Main output latches off at lower input voltages.

Battery backup on auxiliary output.

Operation: Regulator #1's  $V_{out}$  is programmed one diode drop above 5.0 V. Its error flag becomes active when  $V_{in} \leq 5.7$  V. When  $V_{in}$  drops below 5.3 V, the error flag of regulator #2 becomes active and via Q1 latches the main output "off". When  $V_{in}$  again exceeds 5.7 V, regulator #1 is back in regulation and the early warning signal rises, unlatching regulator #2 via D3.

Figure 23. Regulator with Early Warning and Auxiliary Output



 $V_{out} = 1.25V (1.0 + R1/R2)$ 

For 5.0 V output, use internal resistors. Wire Pin 6 to 7, and wire Pin 2 to +V  $_{\rm out}$  Bus.

Figure 24. 2.0 A Low Dropout Regulator

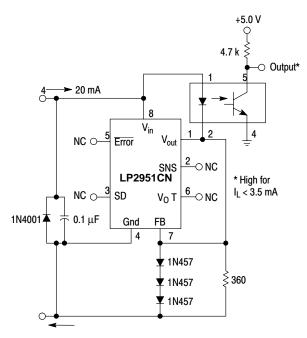


Figure 25. Open Circuit Detector for 4.0 to 20 mA Current Loop

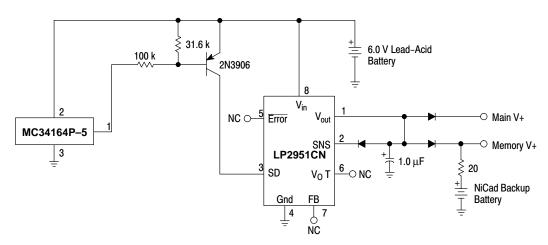


Figure 26. Low Battery Disconnect

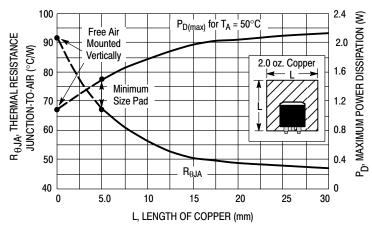


Figure 27. DPAK Thermal Resistance and Maximum Power Dissipation versus P.C.B. Copper Length

#### **ORDERING INFORMATION (LP2950)**

Part Number	Output Voltage (Volts)	Tolerance (%)	Package	Shipping
LP2950CZ-3.0	3.0	1.0	TO-92	2000 Units / Bag
LP2950ACZ-3.0	3.0	0.5	TO-92	2000 Units / Bag
LP2950CZ-3.3	3.3	1.0	TO-92	2000 Units / Bag
LP2950ACZ-3.3	3.3	0.5	TO-92	2000 Units / Bag
LP2950CZ-3.3RA	3.3	1.0	TO-92	2000 Units / Tape & Reel
LP2950ACZ-3.3RA	3.3	0.5	TO-92	2000 Units / Tape & Reel
LP2950CZ-5.0	5.0	1.0	TO-92	2000 Units / Bag
LP2950ACZ-5.0	5.0	0.5	TO-92	2000 Units / Bag
LP2950CZ-5.0RA	5.0	1.0	TO-92	2000 Units / Tape & Reel
LP2950CZ-5.0RP	5.0	1.0	TO-92	2000 Units / Ammo Pack
LP2950ACZ-5.0RA	5.0	0.5	TO-92	2000 Units / Tape & Reel
LP2950CDT-3.0	3.0	1.0	DPAK	75 Units / Rail
LP2950CDT-3.0RK	3.0	1.0	DPAK	2500 Units / Tape & Reel
LP2950ACDT-3.0	3.0	0.5	DPAK	75 Units / Rail
LP2950CDT-3.3	3.3	1.0	DPAK	75 Units / Rail
LP2950CDT-3.3RK	3.3	1.0	DPAK	2500 Units / Tape & Reel
LP2950ACDT-3.3	3.3	0.5	DPAK	75 Units / Rail
LP2950CDT-5.0	5.0	1.0	DPAK	75 Units / Rail
LP2950CDT-5.0RK	5.0	1.0	DPAK	2500 Units / Tape & Reel
LP2950ACDT-5.0	5.0	0.5	DPAK	75 Units / Rail
LP2950ACDT-5.0RK	5.0	0.5	DPAK	2500 Units / Tape & Reel

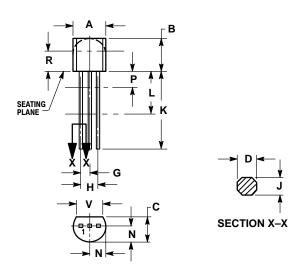
#### **ORDERING INFORMATION (LP2951)**

Part Number	Output Voltage (Volts)	Tolerance (%)	Package	Shipping
LP2951CD-3.0	3.0 or Adj.	1.0	SO-8	98 Units / Rail
LP2951CD-3.0R2	3.0 or Adj.	1.0	SO-8	2500 Units / Tape & Reel
LP2951ACD-3.0	3.0 or Adj.	0.5	SO-8	98 Units / Rail
LP2951ACD-3.0R2	3.0 or Adj.	0.5	SO-8	2500 Units / Tape & Reel
LP2951CD-3.3	3.3 or Adj.	1.0	SO-8	98 Units / Rail
LP2951CD-3.3R2	3.3 or Adj.	1.0	SO-8	2500 Units / Tape & Reel
LP2951ACD-3.3	3.3 or Adj.	0.5	SO-8	98 Units / Rail
LP2951ACD-3.3R2	3.3 or Adj.	0.5	SO-8	2500 Units / Tape & Reel
LP2951CD	5.0 or Adj.	2.0	SO-8	98 Units / Rail
LP2951CDR2	5.0 or Adj.	2.0	SO-8	2500 Units / Tape & Reel
LP2951ACD	5.0 or Adj.	1.2	SO-8	98 Units / Rail
LP2951ACDR2	5.0 or Adj.	1.2	SO-8	2500 Units / Tape & Reel
LP2951CDM-3.0R2	3.0 or Adj.	1.0	Micro-8	2500 Units / Tape & Reel
LP2951ACDM-3.0R2	3.0 or Adj.	0.5	Micro-8	2500 Units / Tape & Reel
LP2951CDM-3.3R2	3.3 or Adj.	1.0	Micro-8	2500 Units / Tape & Reel
LP2951ACDM-3.3R2	3.3 or Adj.	0.5	Micro-8	2500 Units / Tape & Reel
LP2951CDMR2	5.0 or Adj.	2.0	Micro-8	2500 Units / Tape & Reel
LP2951ACDMR2	5.0 or Adj.	1.2	Micro-8	2500 Units / Tape & Reel
LP2951CN-3.0	3.0 or Adj.	1.0	DIP-8	50 Units / Rail
LP2951ACN-3.0	3.0 or Adj.	0.5	DIP-8	50 Units / Rail
LP2951CN-3.3	3.3 or Adj.	1.0	DIP-8	50 Units / Rail
LP2951ACN-3.3	3.3 or Adj.	0.5	DIP-8	50 Units / Rail
LP2951CN	5.0 or Adj.	2.0	DIP-8	50 Units / Rail
LP2951ACN	5.0 or Adj.	1.2	DIP-8	50 Units / Rail

#### **PACKAGE DIMENSIONS**

#### TO-226AA/TO-92 **Z SUFFIX**

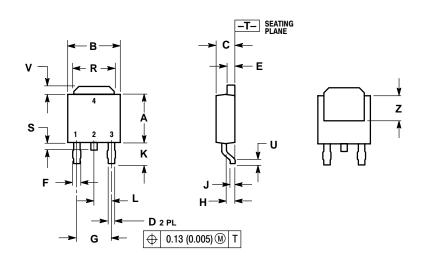
PLASTIC PACKAGE CASE 29-11 **ISSUE AL** 



- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
  2. CONTROLLING DIMENSION: INCH.
  3. CONTOUR OF PACKAGE BEYOND DIMENSION R IS UNCONTROLLED.
  4. LEAD DIMENSION IS UNCONTROLLED IN P AND BEYOND DIMENSION K MINIMUM.

	INC	HES	MILLIN	IETERS
DIM	MIN	MAX	MIN	MAX
Α	0.175	0.205	4.45	5.20
В	0.170	0.210	4.32	5.33
С	0.125	0.165	3.18	4.19
D	0.016	0.021	0.407	0.533
G	0.045	0.055	1.15	1.39
Н	0.095	0.105	2.42	2.66
J	0.015	0.020	0.39	0.50
K	0.500		12.70	
L	0.250		6.35	
N	0.080	0.105	2.04	2.66
P		0.100		2.54
R	0.115		2.93	
٧	0.135		3.43	

#### **DPAK DT SUFFIX** PLASTIC PACKAGE CASE 369A-13 **ISSUE AB**

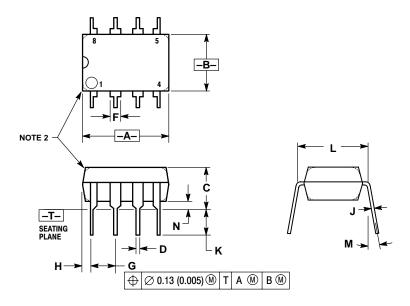


- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982. 2. CONTROLLING DIMENSION: INCH.

	INC	HES	MILLIN	IETERS
DIM	MIN	MAX	MIN	MAX
Α	0.235	0.250	5.97	6.35
В	0.250	0.265	6.35	6.73
С	0.086	0.094	2.19	2.38
D	0.027	0.035	0.69	0.88
E	0.033	0.040	0.84	1.01
F	0.037	0.047	0.94	1.19
G	0.180	BSC	4.58 BSC	
Н	0.034	0.040	0.87	1.01
J	0.018	0.023	0.46	0.58
K	0.102	0.114	2.60	2.89
L	0.090	BSC	2.29	BSC
R	0.175	0.215	4.45	5.46
S	0.020	0.050	0.51	1.27
U	0.020		0.51	
٧	0.030	0.050	0.77	1.27
Z	0.138		3.51	

#### **PACKAGE DIMENSIONS**

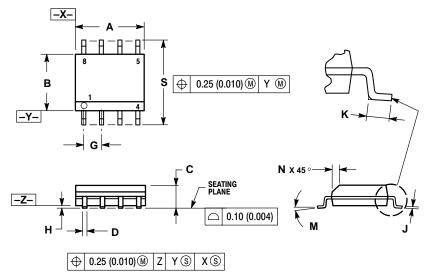
#### PDIP-8 **N SUFFIX** PLASTIC PACKAGE CASE 626-05 ISSUE L



- NOTES: 1. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
- 2. PACKAGE CONTOUR OPTIONAL (ROUND OR SQUARE CORNERS).
  3. DIMENSIONING AND TOLERANCING PER ANSI
- Y14.5M, 1982.

	MILLIN	IETERS	INC	HES
DIM	MIN	MAX	MIN	MAX
Α	9.40	10.16	0.370	0.400
В	6.10	6.60	0.240	0.260
С	3.94	4.45	0.155	0.175
D	0.38	0.51	0.015	0.020
F	1.02	1.78	0.040	0.070
G	2.54	BSC	0.100 BSC	
Н	0.76	1.27	0.030	0.050
J	0.20	0.30	0.008	0.012
K	2.92	3.43	0.115	0.135
L	7.62 BSC		0.300	BSC
M		10°		10°
N	0.76	1.01	0.030	0.040

#### **SO-8 D SUFFIX** PLASTIC PACKAGE CASE 751-07 ISSUE W



- DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

- Y14.5M, 1982.

  CONTROLLING DIMENSION: MILLIMETER.

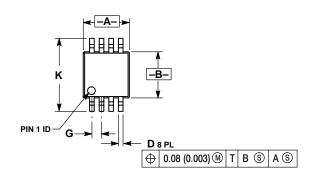
  DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.

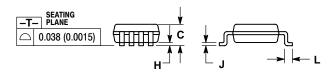
  MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER
- 4. MAXIMUM MULD FROTTICE.
  SIDE.
  5. DIMENSION D DOES NOT INCLUDE DAMBAR
  PROTRUSION. ALLOWABLE DAMBAR
  PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN
  EXCESS OF THE D DIMENSION AT MAXIMUM
  MATERIAL CONDITION.

	MILLIN	IETERS	INC	HES
DIM	MIN	MAX	MIN	MAX
Α	4.80	5.00	0.189	0.197
В	3.80	4.00	0.150	0.157
С	1.35	1.75	0.053	0.069
D	0.33	0.51	0.013	0.020
G	1.27	7 BSC	0.050 BSC	
Н	0.10	0.25	0.004	0.010
J	0.19	0.25	0.007	0.010
K	0.40	1.27	0.016	0.050
M	0 °	8 °	0 °	8 °
N	0.25	0.50	0.010	0.020
S	5.80	6.20	0.228	0.244

#### **PACKAGE DIMENSIONS**

#### Micro-8 **DM SUFFIX** PLASTIC PACKAGE CASE 846A-02 ISSUE E





- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
  2. CONTROLLING DIMENSION: MILLIMETER.
  3. DIMENSION A DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 (0.006) PER SIDE.
  4. DIMENSION B DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 (0.010) PER SIDE.
  - PER SIDE.

	MILLIN	IETERS	INC	HES
DIM	MIN	MAX	MIN	MAX
Α	2.90	3.10	0.114	0.122
В	2.90	3.10	0.114	0.122
С		1.10		0.043
D	0.25	0.40	0.010	0.016
G	0.65	BSC	0.026	BSC
Н	0.05	0.15	0.002	0.006
J	0.13	0.23	0.005	0.009
K	4.75	5.05	0.187	0.199
L	0.40	0.70	0.016	0.028

# **Notes**

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