# Triple Port PHY (Physical Layer) for 25.6 and 51.2 Mbps ATM Networks

PRELIMINARY IDT77V1253

#### **FEATURES**

- Performs the PHY-Transmission Convergence (TC) and Physical Media Dependent (PMD) Sublayer functions for three 25.6 Mbps ATM channels
- Compliant to ATM Forum (af-phy-040.000) and ITU-T I.432.5 specifications for 25.6 Mbps physical interface
- Also operates at 51.2Mbps
- UTOPIA Level 1, UTOPIA Level 2, or DPI-4 Interface
- 3-Cell Transmit & Receive FIFOs
- · LED Interface for status signalling
- · Supports UTP Category 3 physical media
- Interfaces to standard magnetics
- Low-Power CMOS
- 3.3V supply with 5V tolerant inputs
- 144-pin PQFP Package (28 x 28 mm)

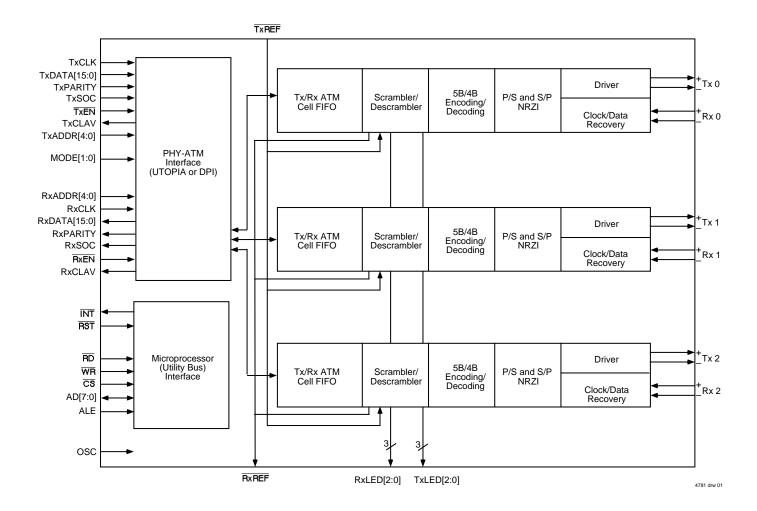
#### DESCRIPTION

The IDT77V1253 is a member of IDT's family of products supporting Asynchronous Transfer Mode (ATM) data communications and networking. The IDT77V1253 implements the physical layer for 25.6 Mbps ATM, connecting three serial copper links (UTP Category 3) to one ATM layer device such as a SAR or a switch ASIC. The IDT77V1253 also operates at 51.2 Mbps, and is well suited to backplane driving applications.

The 77V1253-to-ATM layer interface is selectable as one of three options: 16-bit UTOPIA Level 2, 8-bit UTOPIA Level 1 Multi-PHY, or triple 4-bit DPI (Data Path Interface).

The IDT77V1253 is fabricated using IDT's state-of-the-art CMOS technology, providing the highest levels of integration, performance and reliability, with the low-power consumption characteristics of CMOS.

#### **FUNCTIONAL BLOCK DIAGRAM - UTOPIA LEVEL 2 MODE**



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**NOVEMBER 1998** 

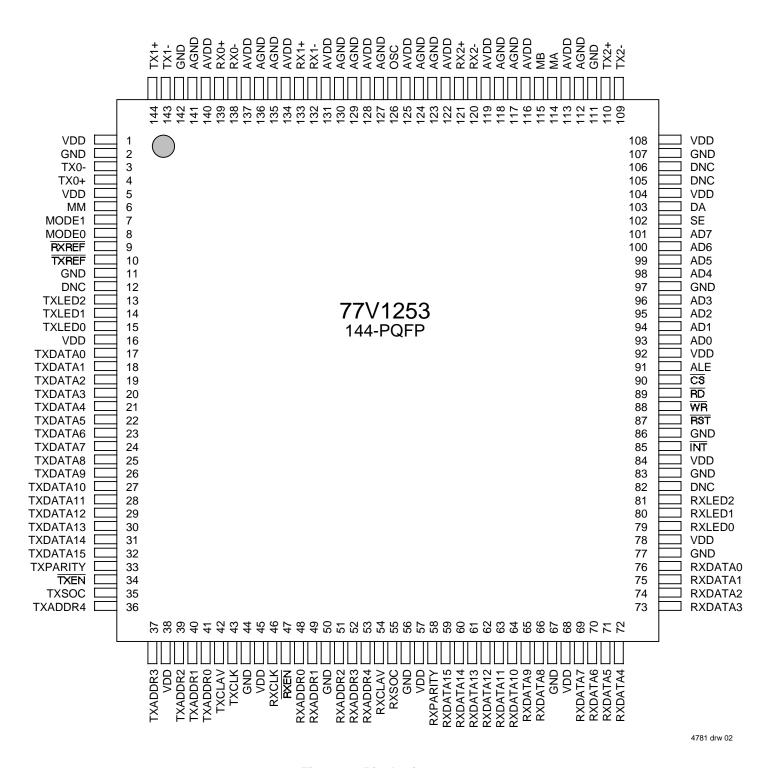


Figure 1. Pin Assignments

## **TABLE 1. SIGNAL DESCRIPTIONS**

#### **LINE SIDE SIGNALS**

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
RX0+, -	139, 138	In	Port 0 positive and negative receive differential input pair
RX1+, -	133, 132	In	Port 1 positive and negative receive differential input pair
RX2+, -	121, 120	In	Port 2 positive and negative receive differential input pair
TX0+, -	4, 3	Out	Port 0 positive and negative transmit differential output pair
TX1+, -	144, 143	Out	Port 1 positive and negative transmit differential output pair
TX2+, -	110, 109	Out	Port 2 positive and negative transmit differential output pair

#### **UTILITY BUS SIGNALS**

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
AD[7:0]	101, 100, 99, 98 96, 95, 94, 93	In/Out	Utility bus address/data bus. The address input is sampled on the falling edge of ALE. Data is output on this bus when a read is performed. Input data is sampled at the completion of a write operation.
ALE	91	In	Utility bus address latch enable. Asynchronous input. An address on the AD bus is sampled on the falling edge of ALE. ALE may be either high low when the AD bus is being used for data.
<u>CS</u>	90	In	Utility bus asynchronous chip select.
RD	89	In	Utility bus read enable. Active low asynchronous input. After latching an address, a read is performed by deasserting WR and asserting RD and CS.
WR	88	In	Utility bus write enable. Active low asynchronous input. After latching an address, a write is performed by deasserting $\overline{\text{RD}}$ , placing data on the AD bus, and asserting $\overline{\text{WR}}$ and $\overline{\text{CS}}$ . Data is sampled when $\overline{\text{WR}}$ or $\overline{\text{CS}}$ is deasserted.

#### **MISCELLANEOUS SIGNALS**

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION	
DA	103	In	Reserved signal. This input must be connected to logic low.	
DNC	12, 82, 105, 106	Out	Do Not Connect. Do not connect these pins to anything external to the chip. They must remain open.	
ĪNT	85	Out open drain	Interrupt. INT is an open-drain output, driven low to indicate an interrupt. Once low, INT remains low until the interrupt status in the appropriate Interrupt Status Register is read. Interrupt sources are programmable via the Interrupt Mask Registers.	
MA	114	In	Reserved signal. This input must be connected to logic low.	
MB	115	In	Reserved signal. This input must be connected to logic low.	
MM	6	In	Reserved signal. This input must be connected to logic high.	
MODE[1:0]	7, 8	In	Mode Selects. They determine the configuration of the PHY/ATM interface. 00 = UTOPIA Level 2. 01 = UTOPIA Level 1. 10 = DPI. 11 is reserved.	
OSC	126	In	TTL line rate clock source, driven by a 100 ppm oscillator. 32 MHz for 25.6 Mbps; 64 MHz for 51.2 Mbps.	
RST	87	In	Reset. Active low asynchronous input resets all control logic, counters and FIFOs. A reset must be performed after power up prior to normal operation of the part.	
RXLED[2:0]	81, 80, 79	Out	Receive LED drivers. Driven low for 2 <sup>23</sup> RCLK or DPICLK cycles, beginning with RXSOC when that port receives a good (non-null and non-errored) cell. Drives 8 mA both high and low. One per port.	

## **TABLE 1. SIGNAL DESCRIPTIONS (continued)**

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
RXREF	9	Out	Receive Reference. Active low, synchronous to OSC. RXREF pulses low for a programmable number of clock cycles when an X_8 command byte is received. Register 0x40 is programmed to indicate which port is referenced.
SE	102	In	Reserved signal. This input must be connected to logic low.
TXLED[2:0]	13, 14, 15	Out	Ports 2 thru 0 Transmit LED driver. Goes low for 2 <sup>23</sup> TCLK or DPICLK cycles, beginning with TXSOC when this port receives a cell for transmission. 8mA drive current both high and low. One per port.
TXREF	10	In	Transmit Reference. Synchronous to OSC. When this pin is asserted, an X_8 command byte is inserted into the transmit data stream. Logic for this signal is programmed in register 0x40. Typical application is WAN timing.

#### **POWER SUPPLY SIGNALS**

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
AGND	112, 117, 118, 123,124,127, 129,130,135, 136, 141	-	Analog ground. AGND supply a ground reference to the analog portion of the chip, which sources a more constant current than the digital portion.
AVDD	113, 116, 119, 122, 125, 128, 131, 134, 137, 140	-	Analog power supply. $3.3\pm0.3 \text{V}$ AVDD supply power to the analog portion of the chip, which draws a more constant current than the digital portion.
GND	2, 11, 44, 50, 56 67, 77, 83, 86, 97, 107, 111, 142	-	Digital Ground
VDD	1, 5, 16, 38, 45 57, 68, 78, 84, 92, 104, 108	-	Digital power supply. 3.3 ±0.3V

## 16-BIT UTOPIA 2 SIGNALS (MODE[1:0] = 00)

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
RXADDR[4:0]	53, 52, 51, 49, 48	In	Utopia 2 Receive Address Bus. This bus is used in polling and selecting the receive port. The port addresses are defined in bits [4:0] of the Enhanced Control Registers.
RXCLAV	54	Out	Utopia 2 Receive Cell Available. Indicates the cell available status of the addressed port. It is asserted when a full cell is available for retrieval from the receive FIFO. When none of the three ports is addressed, RXCLAV is high impedance.
RXCLK	46	In	Utopia 2 Receive Clock. This is a free running clock input.
RXDATA[15:0]	59, 60, 61, 62, 63, 64, 65, 66, 69, 70, 71, 72, 73, 74, 75, 76	Out	Utopia 2 Receive Data. When one of the three ports is selected, the 77V1253 transfers received cells to an ATM device across this bus. Also see RXPARITY.
RXEN	47	In	Utopia 2 Receive Enable. Driven by an ATM device to indicate its ability to receive data across the RXDATA bus.
RXPARITY	58	Out	Utopia 2 Receive Data Parity. Odd parity over RXDATA[15:0].
RXSOC	55	Out	Utopia 2 Receive Start of Cell. Asserted coincident with the first word of data for each cell on RXDATA.
TXADDR[4:0]	36, 37, 39, 40, 41	In	Utopia 2 Transmit Address Bus. This bus is used in polling and selecting the transmit port. The port addresses are defined in bits [4:0] of the Enhanced Control Registers.
TXCLAV	42	Out	Utopia 2 Transmit Cell Available. Indicates the availability of room in the transmit FIFO of the addressed port for a full cell. When none of the three ports is addressed, TXCLAV is high impedance.

## TABLE 1. SIGNAL DESCRIPTIONS (continued)

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
TXCLK	43	In	Utopia 2 Transmit Clock. This is a free running clock input.
TXDATA[15:0]	32, 31, 30, 29, 28, 27, 26, 25, 24, 23, 22, 21, 20, 19, 18, 17	In	Utopia 2 Transmit Data. An ATM device transfers cells across this bus to the 77V1253 for transmission. Also see TXPARITY.
TXEN	34	In	Utopia 2 Transmit Enable. Driven by an ATM device to indicate it is transmitting data across the TXDATA bus.
TXPARITY	33	In	Utopia 2 Transmit Data Parity. Odd parity across TXDATA[15:0].  Parity is checked and errors are indicated in the Interrupt Status Registers, as enabled in the Master Control Registers. No other action is taken in the event of an error. Tie high or low if unused.
TXSOC	35	In	Utopia 2 Transmit Start of Cell. Asserted coincident with the first word of data for each cell on TXDATA.

## 8-BIT UTOPIA LEVEL 1 SIGNALS (MODE[1:0] = 01)

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
RXCLAV[2:0]	65, 66, 54	Out	Utopia 1 Receive Cell Available. Indicates the cell available status of the respective port. It is asserted when a full cell is available for retrieval from the receive FIFO.
RXCLK	46	In	Utopia 1 Receive Clock. This is a free running clock input.
RXDATA[7:0]	69, 70, 71, 72, 73, 74, 75, 76	Out	Utopia 1 Receive Data. When one of the three ports is selected, the 77V1253 transfers received cells to an ATM device across this bus. Bit 5 in the Diagnostic Control Registers determines whether RXDATA tri-states when RXEN[2:0] are high. Also see RXPARITY.
RXEN[2:0]	49, 48, 47	In	Utopia 1 Receive Enable. Driven by an ATM device to indicate its ability to receive data across the RXDATA bus. One for each port.
RXPARITY	58	Out	Utopia 1 Receive Data Parity. Odd parity over RXDATA[7:0].
RXSOC	55	Out	Utopia 1 Receive Start of Cell. Asserted coincident with the first word of data for each cell on RXDATA. Tri-statable as determined by bit 5 in the Diagnostic Control Registers.
TXCLAV[2:0]	40, 41, 42	Out	Utopia 1 Transmit Cell Available. Indicates the availability of room in the transmit FIFO of the respective port for a full cell.
TXCLK	43	In	Utopia 1 Transmit Clock. This is a free running clock input.
TXDATA[7:0]	24, 23, 22, 21, 20, 19, 18, 17	In	Utopia 1 Transmit Data. An ATM device transfers cells across this bus to the 77V1253 for transmission. Also see TXPARITY.
TXEN[2:0]	26, 25, 34	In	Utopia 1 Transmit Enable. Driven by an ATM device to indicate it is transmitting data across the TXDATA bus. One for each port.
TXPARITY	33	In	Utopia 1 Transmit Data Parity. Odd parity across TXDATA[7:0].  Parity is checked and errors are indicated in the Interrupt Status Registers, as enabled in the Master Control Registers. No other action is taken in the event of an error. Tie high or low if unused.
TXSOC	35	In	Utopia 1 Transmit Start of Cell. Asserted coincident with the first word of data for each cell on TXDATA.

## DPI MODE SIGNALS (MODE[1:0] = 10)

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
DPICLK	43	In	DPI Source Clock for Transmit. This is the free-running clock used as the source to generate Pn_TCLK.
Pn_RCLK	51, 49, 48	ln	DPI Port 'n' Receive CLock. Pn_RCLK is cycled to indicate that the interfacing device is ready to receive a nibble of data on Pn_RD[3:0] of port 'n'.
Pn_RD[3:0]	63, 64, 65, 66, 69, 70, 71, 72, 73, 74, 75, 76	Out	DPI Port 'n' Receive Data. Cells received on port 'n' are passed to the interfacing device across this bus. Each port has its own dedicated bus.

## **TABLE 1. SIGNAL DESCRIPTIONS (continued)**

SIGNAL NAME	PIN NUMBER	I/O	SIGNAL DESCRIPTION
Pn_RFRM	58, 54, 55	Out	DPI Port 'n' Receive Frame. Pn_RFRM is asserted for one cycle immediately preceding the transfer of each cell on Pn_RD[3:0].
Pn_TCLK	39, 40, 41	Out	DPI Port 'n' Transmit Clock. Pn_TCLK is derived from DPICLK, and is cycled when the respective port is ready to accept another 4 bits of data on Pn_TD[3:0].
Pn_TD[3:0]	28, 27, 26, 25, 24, 23, 22, 21, 20, 19, 18, 17	In	DPI Port 'n' Transmit Data. Cells are passed across this bus to the PHY for transmission on port 'n'. Each port has its own dedicated bus.
Pn_TFRM	36, 33, 34, 35	In	DPI Port 'n' Transmit Frame. Start of cell signal which is asserted for one cycle immediately preceding the first 4 bits of each cell on Pn_TD[3:0].

TABLE 2. SIGNAL ASSIGNMENT AS A FUNCTION OF PHY/ATM INTERFACE MODE

SIGNAL NAME	PIN NUMBER	16-BIT UTOPIA 2	8-BIT UTOPIA 1	DPI
		MODE[1,0] = 00	MODE[1,0] = 01	MODE[1,0] = 10
VDD	1			
GND	2			
TX0-	3			
TX0+	4			
VDD	5			
MM	6			
MODE1	7			
MODE0	8			
RXREF	9			
TXREF	10			
GND	11			
DNC	12			
TXLED2	13			
TXLED1	14			
TXLED0	15			
VDD	16			
TXDATA0	17	TXDATA0	TXDATA0	P0_TD[0]
TXDATA1	18	TXDATA1	TXDATA1	P0_TD[1]
TXDATA2	19	TXDATA2	TXDATA2	P0_TD[2]
TXDATA3	20	TXDATA3	TXDATA3	P0_TD[3]
TXDATA4	21	TXDATA4	TXDATA4	P1_TD[0]
TXDATA5	22	TXDATA5	TXDATA5	P1_TD[1]
TXDATA6	23	TXDATA6	TXDATA6	P1_TD[2]
TXDATA7	24	TXDATA7	TXDATA7	P1_TD[3]
TXDATA8	25	TXDATA8	TXEN[1]	P2_TD[0]
TXDATA9	26	TXDATA9	TXEN[2]	P2_TD[1]
TXDATA10	27	TXDATA10	see note 2	P2_TD[2]
TXDATA11	28	TXDATA11	see note 2	P2_TD[3]
TXDATA12	29	TXDATA12	see note 2	see note 2

TABLE 2. SIGNAL ASSIGNMENT AS A FUNCTION OF PHY/ATM INTERFACE MODE (cont.)

SIGNAL NAME	PIN MUMBER	16-BIT UTOPIA 2	8-BIT UTOPIA 1	DPI	
TXDATA13	30	TXDATA13	see note 2	see note 2	
TXDATA14	31	TXDATA14	see note 2	see note 2	
TXDATA15	32	TXDATA15	see note 2	see note 2	
TXPARITY	33	TXPARITY	TXPARITY	P2_TFRM	
TXEN	34	TXEN	TXEN[0]	P1_TFRM	
TXSOC	35	TXSOC	TXSOC	P0_TFRM	
TXADDR4	36	TXADDR4	see note 2	see note 2	
TXADDR3	37	TXADDR3	see note 2	see note 2	
VDD	38				
TXADDR2	39	TXADDR2	see note 1	P2_TCLK	
TXADDR1	40	TXADDR1	TXCLAV[2]	P1_TCLK	
TXADDR0	41	TXADDR0	TXCLAV[1]	P0_TCLK	
TXCLAV	42	TXCLAV	TXCLAV[0]	see note 1	
TXCLK	43	TXCLK	TXCLK	DPICLK	
GND	44				
VDD	45				
RXCLK	46	RXCLK	RXCLK	see note 2	
RXEN	47	RXEN	RXEN[0]	see note 2	
RXADDR0	48	RXADDR0	RXEN[1]	P0_RCLK	
RXADDR1	49	RXADDR1	RXEN[2]	P1_RCLK	
GND	50				
RXADDR2	51	RXADDR2	see note 2	P2_RCLK	
RXADDR3	52	RXADDR3	see note 2	see note 2	
RXADDR4	53	RXADDR4	see note 2	see note 2	
RXCLAV	54	RXCLAV	RXCLAV[0]	P1_RFRM	
RXSOC	55	RXSOC	RXSOC	P0_RFRM	
GND	56				
VDD	57				
RXPARITY	58	RXPARITY	RXPARITY	P2_RFRM	
RXDATA15	59	RXDATA15	see note 1	see note 1	
RXDATA14	60	RXDATA14	see note 1	see note 1	
RXDATA13	61	RXDATA13	see note 1	see note 1	
RXDATA12	62	RXDATA12	see note 1	see note 1	
RXDATA11	63	RXDATA11	see note 1	P2_RD[3]	
RXDATA10	64	RXDATA10	see note 1	P2_RD[2]	
RXDATA9	65	RXDATA9	RXCLAV[2]	P2_RD[1]	
RXDATA8	66	RXDATA8	RXCLAV[1]	P2_RD[0]	
GND	67				
VDD	68				
RXDATA7	69	RXDATA7	RXDATA7	P1_RD[3]	
RXDATA6	70	RXDATA6	RXDATA6	P1_RD[2]	
RXDATA5	71	RXDATA5	RXDATA5	P1_RD[1]	
RXDATA4	72	RXDATA4	RXDATA4	P1_RD[0]	
RXDATA3	73	RXDATA3	RXDATA3	P0_RD[3]	
RXDATA2	74	RXDATA2	RXDATA2	P0_RD[2]	
RXDATA1	75	RXDATA1	RXDATA1	P0_RD[1]	

## TABLE 2. SIGNAL ASSIGNMENT AS A FUNCTION OF PHY/ATM INTERFACE MODE (cont.)

SIGNAL NAME	PIN MUMBER	16-BIT UTOPIA 2	8-BIT UTOPIA 1	DPI
RXDATA0	76	RXDATA0	RXDATA0	P0_RD[0]
GND	77			
VDD	78			
RXLED0	79			
RXLED1	80			
RXLED2	81			
DNC	82			
GND	83			
VDD	84			
ĪNT	85			
GND	86			
RST	87			
WR	88			
RD	89			
CS	90			
ALE	91			
VDD	92			
AD0	93			
AD0	94			
AD0	95			
AD0	96			
GND	97			
AD0	98			
AD0	99			
AD0	100			
AD0	101			
SE	102			
DA	103			
VDD	104			
DNC	105			
DNC	106			
GND	107			
VDD	108			
TX2-	109			
TX2+	110			
GND	111			
AGND	112			
AVDD	113			
MA	114			
MB	115			
AVDD	116			
AGND	117			
AGND	118			
AVDD	119			
RX2-	120			
RX2+	121			
<b>ΓΛ</b> Δ+	121	<u> </u>		1

## TABLE 2. SIGNAL ASSIGNMENT AS A FUNCTION OF PHY/ATM INTERFACE MODE (cont.)

SIGNAL NAME	PIN MUMBER	16-BIT UTOPIA 2	8-BIT UTOPIA 1	DPI
AVDD	122			
AGND	123			
AGND	124			
AVDD	125			
OSC	126			
AGND	127			
AVDD	128			
AGND	129			
AGND	130			
AVDD	131			
RX1-	132			
RX1+	133			
AVDD	134			
AGND	135			
AGND	136			
AVDD	137			
RX0-	138			
RX0+	139			
AVDD	140			
AGND	141			
GND	142			
TX1-	143			
TX1+	144			

#### NOTES:

- This output signal is unused in this mode. It must be left unconnected.
   This input signal is unused in this mode. It must be connected to either logic high or logic low.

#### 77V1253 OVERVIEW

The 77V1253 is a three-port implementation of the physical layer standard for 25.6Mbps ATM network communications as defined by ATM Forum document af-phy-040.000 and ITU-T I.432.5. The physical layer is divided into a Physical Media Dependent sub layer (PMD) and Transmission Convergence (TC) sub layer. The PMD sub layer includes the functions for the transmitter, receiver and clock recovery for operation across 100 meters of category 3 unshielded twisted pair (UTP) cable. This is referred to as the Line Side Interface. The TC sub layer defines the line coding, scrambling, data framing and synchronization.

On the other side, the 77V1253 interfaces to an ATM layer device (such as a switch core or SAR). This cell level interface is configurable as either 8-bit Utopia Level 1 Multi-PHY, 16-bit Utopia Level 2, or as three 4-bit DPI interfaces, as determined by two MODE pins. This is referred to as the PHY-ATM Interface. The pinout and front page block diagram are based on the Utopia 2 configuration. Table 2 shows the corresponding pin functions for the other two modes, and Figures 2 and 3 show functional block diagrams.

The 77V1253 is based on the 77105, and maintains significant register compatibility with it. The 77V1253, however, has

additional register features, and also duplicates most of its registers to provide significant independence between the three ports.

Access to these status and control registers is through the utility bus. This is an 8-bit muxed address and data bus, controlled by a conventional asynchronous read/write handshake.

Additional pins permit insertion and extraction of an 8 kHz timing marker, and provide LED indication of receive and transmit status.

#### **OPERATION AT 51.2 Mbps**

In addition to operation at the standard rate of 25.6 Mbps, the 77V1253 is also specified to operate at 51.2 Mbps. Except for the doubled bit rate, all other aspects of operation are identical to the 25.6 Mbps mode. The data rate is determined by the frequency of the clock applied to the OSC input. OSC is 32 MHz for the 25.6 Mbps line rate, and 64 MHz for the 51.2 Mbps line rate. All ports operate at the same frequency.

See page 30 for recommended line magnetics. Magnetics for 51.2 Mbps operation have a higher bandwidth than magnetics optimized for 25.6 Mbps.

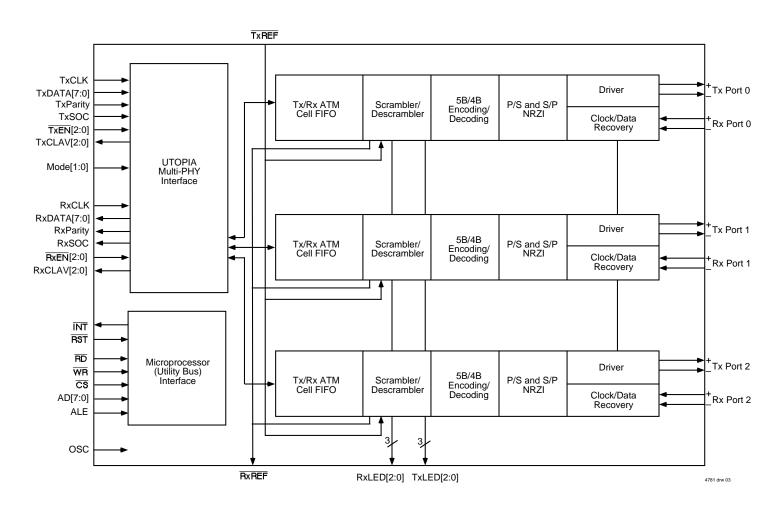


Figure 2. Block Diagram for Utopia Level 1 configuration (MODE[1:0] = 01)

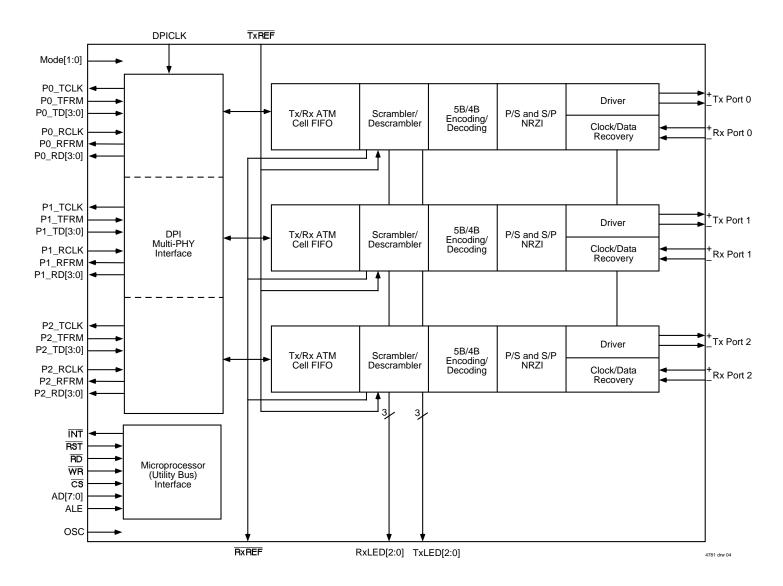


Figure 3. Block Diagram for DPI configuration (MODE[1:0] = 10)

#### **FUNCTIONAL DESCRIPTION**

#### TRANSMISSION CONVERGENCE (TC) SUB LAYER

#### Introduction

The TC sub layer defines the line coding, scrambling, data framing and synchronization. Under control of a switch interface or Segmentation and Reassembly (SAR) unit, the 25.6Mbps ATM PHY accepts a 53-byte ATM cell, scrambles the data, appends a command byte to the beginning of the cell, and encodes the entire 53 bytes before transmission. These data transformations ensure that the signal is evenly distributed across the frequency spectrum. In addition, the serialized bit stream is NRZI coded. An 8 kHz timing sync pulse may be used for isochronous communications.

#### **Data Structure and Framing**

Each 53-byte ATM cell is preceded with a command byte. This byte is distinguished by an escape symbol followed by one of 17 encoded symbols. Together, this byte forms one of seventeen possible command bytes. Three command bytes are defined:

X\_X (read: 'escape' symbol followed by another 'escape'): Start-of-cell with scrambler/descrambler reset.

- 2. **X\_4** ('escape' followed by '4'): Start-of-cell without scrambler/descrambler reset.
- 3. **X\_8** ('escape' followed by '8'): 8kHz timing marker. This command byte is generated when the 8kHz sync pulse is detected, and has priority over all line activity (data or command bytes). It is transmitted immediately when the sync pulse is detected. When this occurs during a cell transmission, the data transfer is temporarily interrupted on an octet boundary, and the X\_8 command byte is inserted. This condition is the only allowed interrupt in an otherwise contiguous transfer.

Below is an illustration of the cell structure and command byte usage:

 $\{X_X\} \{53\text{-byte ATM cell}\} \{X_4\} \{53\text{-byte ATM } \{X_8\} \text{ cell}\} ...$ 

In the above example, the first ATM cell is preceded by the X\_X start-of-cell command byte which resets both the transmitter-scrambler and receiver-descrambler pseudo-random nibble generators (PRNG) to their initial states. The following cell illustrates the insertion of a start-of-cell command without scrambler/descrambler reset. During this cell's transmission, an 8kHz timing sync pulse triggers insertion of the X\_8 8kHz timing marker command byte.

## **FUNCTIONAL BLOCK DIAGRAM (Continued)**

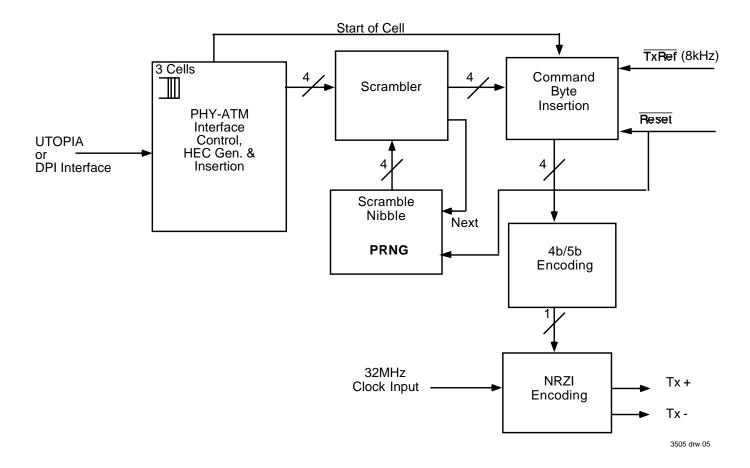


Figure 4. TC Transmit Block Diagram

#### **Transmission Description**

Refer to Figure 4 on the previous page. Cell transmission begins with the PHY-ATM Interface. An ATM layer device transfers a cell into the 77V1253 across the Utopia or DPI transmit bus. This cell enters a 3-cell deep transmit FIFO. Once a complete cell is in the FIFO, transmission begins by passing the cell, four bits (MSB first) at a time to the 'Scrambler'.

The 'Scrambler' takes each nibble of data and exclusive-ORs them against the 4 high order bits (X(t), X(t-1), X(t-2), X(t-3)) of a 10 bit pseudo-random nibble generator (PRNG). Its function is to provide the appropriate frequency distribution for the signal across the line.

The PRNG is clocked every time a nibble is processed, regardless of whether the processed nibble is part of a data or command byte. Note however that only data nibbles are scrambled. The entire command byte (X\_C) is NOT scrambled before it's encoded (see diagram for illustration). The PRNG is based upon the following polynomial:

$$X^{10} + X^7 + 1$$

With this polynomial, the four output data bits (D3, D2, D1, D0) will be generated from the following equations:

D3 = d3 xor X(t-3)D2 = d2 xor X(t-2)D1 = d1 xor X(t-1)D0 = d0 xor X(t)

The following nibble is scrambled with X(t+4), X(t+3), X(t+2), and X(t+1).

A scrambler lock between the transmitter and receiver occurs each time an  $X_X$  command is sent. An  $X_X$  command is initiated only at the beginning of a cell transfer after the PRNG has cycled through all of its states ( $2^{10}$  - 1 = 1023 states). The first valid ATM data cell transmitted after power on will also be accompanied with an  $X_X$  command byte. Each time an  $X_X$  command byte is sent, the first nibble after the last escape (X) nibble is XOR'd with 1111b (PRNG = 3FFx).

Because a timing marker command (X\_8) may occur at any time, the possibility of a reset PRNG start-of-cell command and a timing marker command occurring consecutively does exist (e.g. X\_X\_X\_8). In this case, the detection of the last two consecutive escape (X) nibbles will cause the PRNG to reset to its initial 3FFx state. Therefore, the PRNG is clocked only after the first nibble of the second consecutive escape pair.

Once the data nibbles have been scrambled using the PRNG, the nibbles are further encoded using a 4b/5b process. The 4b/5b scheme ensures that an appropriate number of signal transitions occur on the line. A total of seventeen 5-bit symbols are used to represent the sixteen 4-bit data nibbles and the one escape (X) nibble. The table below lists the 4-bit data with their corresponding 5-bit symbols:

Data	Symbol
0000	10101
0100	00111
1000	10010
1100	10111

<u>Data</u> 0001 0101	<u>Symbol</u> 01001 01101
1001	11001
1101	11101

<u>Data</u> 0010	Symbol 01010
0110	01110
1010	11010
1110	11110

<u>Data</u>	<u>Symbol</u>
0011	01011
0111	01111
1011	11011
1111	11111

ESC(X) = 00010 3217 tbl 01

This encode/decode implementation has several very desirable properties. Among them is the fact that the output data bits can be represented by a set of relatively simple symbols:

- Run length is limited to <= 5;
- Disparity never exceeds +/- 1.

On the receiver, the decoder determines from the received symbols whether a timing marker command (X\_8) or a start-of-cell command was sent (X\_X or X\_4). If a start-of-cell command is detected, the next 53 bytes received are decoded and forwarded to the descrambler. (See TC Receive Block Diagram, Figure 4).

The output of the 4b/5b encoder provides serial data to the NRZI encoder. The NRZI code transitions the wire voltage each time a '1' bit is sent. This, together with the previous encoding schemes guarantees that long run lengths of either '0' or '1's are prevented. Each symbol is shifted out with its most significant bit sent first.

When no cells are available to transmit, the 77V1253 keeps the line active by continuing to transmit valid symbols. But it does not transmit another start-of-cell command until it has another cell for transmission. The 77V1253 never generates idle cells.

#### **Transmit HEC Byte Calculation/Insertion**

Byte #5 of each ATM cell, the HEC (Header Error Control) is calculated automatically across the first 4 bytes of the cell header, depending upon the setting of bit 5 of registers 0x03, 0x13 and 0x23. This byte is then either inserted as a replacement of the fifth byte transferred to the PHY by the external system, or the cell is transmitted as received. A third operating mode provides for insertion of "Bad" HEC codes which may aid in communication diagnostics. These modes are controlled by the LED Driver and HEC Status/Control Registers.

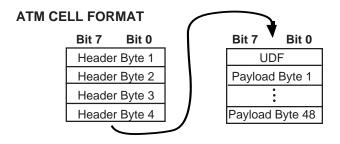
UDF = User Defined Field (or HEC)

#### **Receiver Description**

The receiver side of the TC sublayer operates like the transmitter, but in reverse. The data is NRZI decoded before each symbol is reassembled. The symbols are then sent to the 5b/4b decoder, followed by the Command Byte Interpreter, De-Scrambler, and finally through a FIFO to the UTOPIA or DPI interface to an ATM Layer device.

Note that although the IDT77V1253 can detect symbol and HEC errors, it does not attempt to correct them.

Upon reset or the re-connect, each port's receiver is typically not symbol-synchronized. When not symbol-synchro-



nized, the receiver will indicate a significant number of bad symbols, and will deassert the Good Signal Bit as described below. Synchronization is established immediately once that port receives an Escape symbol, usually as part of the startof-cell command byte preceding the first received cell.

The IDT77V1253 monitors line conditions and can provide an interrupt if the line is deemed 'bad'. The Interrupt Status Registers (registers 0x01, 0x11 and 0x21) contain a Good Signal Bit (bit 6, set to 0 = Bad signal initially) which shows the status of the line per the following algorithm:

#### To declare 'Good Signal' (from "Bad" to "Good"):

There is an up-down counter that counts from 7 to 0 and is initially set to 7. When the clock ticks for 1,024 cycles (32MHz clock, 1,024 cycles = 204.8 symbols) and no "bad symbol" has been received, the counter decreases by one. However, if at least one "bad symbol" is detected during these 1,024 clocks, the counter is increased by one, to a maximum of 7. The Good Signal Bit is set to 1 when this counter reaches 0. The Good Signal Bit could be set to 1 as quickly as 1,433 symbols (204.8 x 7) if no bad symbols have been received.

## **FUNCTIONAL BLOCK DIAGRAM (Continued)**

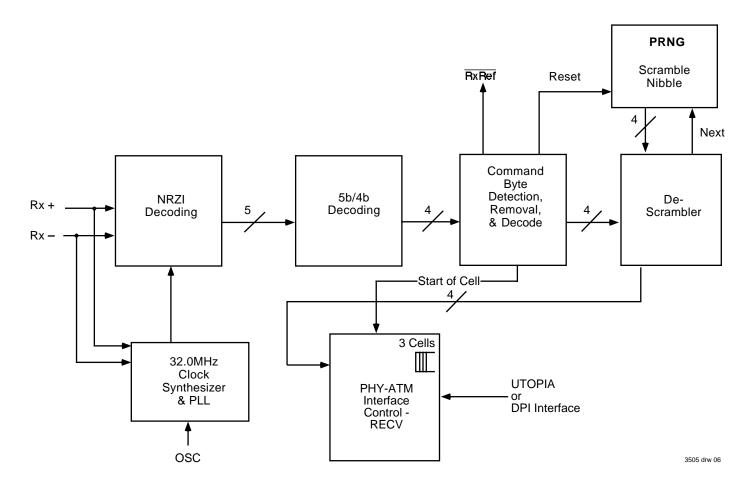


Figure 5. TC Receive Block Diagram

#### To declare 'Bad Signal' (from "Good" to "Bad"):

The same up-down counter counts from 0 to 7 (being at 0 to provide a "Good" status). When the clock ticks for 1,024 cycles (32MHz clock, 1,024 cycles = 204.8 symbols) and there is at least one "bad symbol", the counter increases by one. If it detects all "good symbols" and no "bad symbols" in the next time period, the counter decreases by one. The "Bad Signal" is declared when the counter reaches 7. The Good Signal Bit could be set to 0 as quickly as 1,433 symbols (204.8 x 7) if at least one "bad symbol" is detected in each of seven consecutive groups of 204.8 symbols.

#### 8kHz Timing Marker

The 8kHz timing marker, described earlier, is a completely optional feature which is essential for some applications requiring synchronization for voice or video, and unnecessary for other applications. Figure 6 shows the options available for generating and receiving the 8kHz timing marker.

The source of the marker is programmable in the RXREF and TXREF Control Register (0x40). Each port is individually programmable to either a local source or a looped remote source. The local source is TXREF, which is an 8kHz clock of virtually any duty cycle. When unused, TXREF should be tied high. Also note that it is not limited to 8kHz, should a different frequency be desired. When looped, a received X\_8 command byte causes one to be generated on the transmit side.

A received X\_8 command byte causes the 77V1253 to issue a negative pulse on RXREF. The source channel of the marker is programmable.

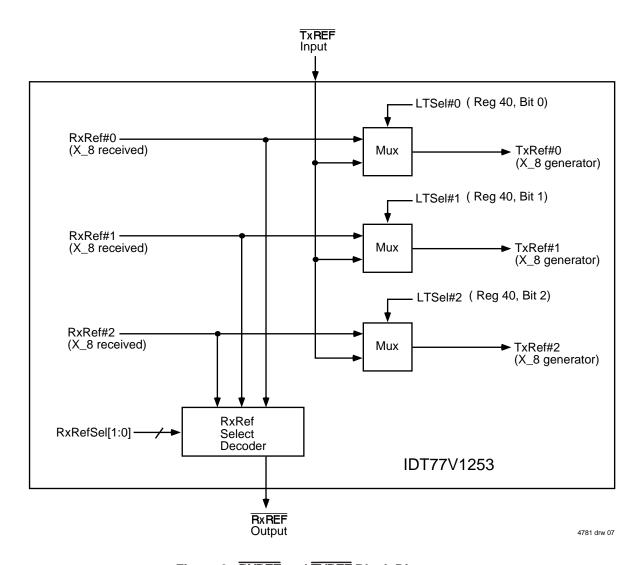


Figure 6. RXREF and TXREF Block Diagram

#### PHY-ATM INTERFACE

The 77V1253 PHY offers three choices in interfacing to ATM layer devices such as segmentation and reassembly (SAR) and switching chips. MODE[1:0] are used to select the configuration of this interface, as shown in the table below.

MODE[1:0]	PHY-ATM Interface configuration
00	one 16-bit UTOPIA Level 2 port
01	one 8-bit UTOPIA Level 1 (Multi-PHY)port
10	three 4-bit Data Path Interface (DPI) ports

UTOPIA is a Physical Layer to ATM Layer interface standardized by the ATM Forum. It has separate transmit and receive channels and specific handshaking protocols. UTOPIA Level 2 has dedicated address signals for both the transmit and receive directions that allow the ATM layer device to specify with which of the four PHY channels it is communicating. UTOPIA Level 1 does not have address signals. Instead, key handshaking signals are duplicated so that each channel has its own signals. In both versions of UTOPIA, all channels share a single transmit data bus and a single receive data bus for data transfer.

DPI is a low-pincount Physical Layer to ATM Layer interface. The low-pincount characteristic allows the 77V1253 to incorporate three separate DPI 4-bit ports, one for each of the three serial ports. As with the UTOPIA interfaces, the transmit and receive directions have their own data paths and handshaking.

#### **UTOPIA LEVEL 2 INTERFACE OPTION**

The 16-bit Utopia Level 2 interface operates as defined in ATM Forum document af-phy-0039. This PHY-ATM interface is selected by setting the MODE[1:0] pins both low.

This mode is configured as a single 16-bit data bus in the transmit (ATM-to-PHY) direction, and a single 16-bit data bus in the receive (PHY-to-ATM) direction. In addition to the data bus, each direction also includes a single optional parity bit, an address bus, and several handshaking signals. The UTOPIA address of each channel is determined by bits 4 to 0 in the Enhanced Control Registers. Please note that the transmit bus and the receive bus operate completely independently. The Utopia 2 signals are summarized below:

TXDATA[15:0] TXPARITY TXSOC TXADDR[4:0] TXEN TXCLAV TXCLK	ATM to PHY ATM to PHY ATM to PHY ATM to PHY ATM to PHY PHY to ATM ATM to PHY
RXDATA[15:0]	PHY to ATM
RXPARITY	PHY to ATM
RXSOC	PHY to ATM
RXADDR[4:0]	ATM to PHY
RXEN	ATM to PHY
RXCLAV	PHY to ATM

RXCLK ATM to PHY

The ATM device starts by polling the PHY ports on the Utopia 2 bus to determine if any of them has room to accept a cell for transmission (TXCLAV), or has a receive cell available to pass on to the ATM device (RXCLAV). To poll, the ATM device drives an address (TXADDR or RXADDR) then observes TXCLAV or RXCLAV on the next cycle of TXCLK or RXCLK. A port must tri-state TXCLAV and RXCLAV except when it is addressed.

If TXCLAV or RXCLAV is asserted, the ATM device may select that port, then transfer a cell to or from it. Selection of a port is performed by driving the address of the desired port while TXEN or RXEN is high, then driving TXEN or RXEN low. When TXEN is driven low, TXSOC (start of cell) is driven high to indicate that the first 16 bits of the cell are being driven on TXDATA. The ATM device may chose to temporarily suspend transfer of the cell by deasserting TXEN. Otherwise, TXEN remains asserted as the next 16 bits are driven onto TXDATA with each cycle of TXCLK.

In the receive direction, the ATM device selects a port if it wished to receive the cell that the port is holding. It does this by asserting  $\overline{\text{RXEN}}$ . The PHY then transfers the data 16 bits each clock cycle, as determined by  $\overline{\text{RXEN}}$ . As in the transmit direction, the ATM device may suspend transfer by deasserting  $\overline{\text{RXEN}}$  at any time. Note that the PHY asserts RXSOC coincident with the first 16 bits of each cell.

TXPARITY and RXPARITY are parity bits for the corresponding 16-bit data fields. Odd parity is used.

Figures 8 through 13 may be referenced for Utopia 2 bus examples.

Because this interface transfers an even number of bytes, rather than the ATM standard of 53 bytes, a filler byte is inserted between the 5-byte header and the 48-byte payload. This is shown in Figure 7.

	Bit 15	Bit 0
First	Header byte 1	Header byte 2
	Header byte 3	Header byte 4
	Header byte 5	stuff byte
	Payload byte 1	Payload byte 2
	Payload byte 3	Payload byte 4
	Payload byte 5	Payload byte 6
	Payload byte 45	Payload byte 46
Last	Payload byte 47	Payload byte 48
	L	

3505 drw 08

Figure 7. Utopia Level 2 Data Format and Sequence

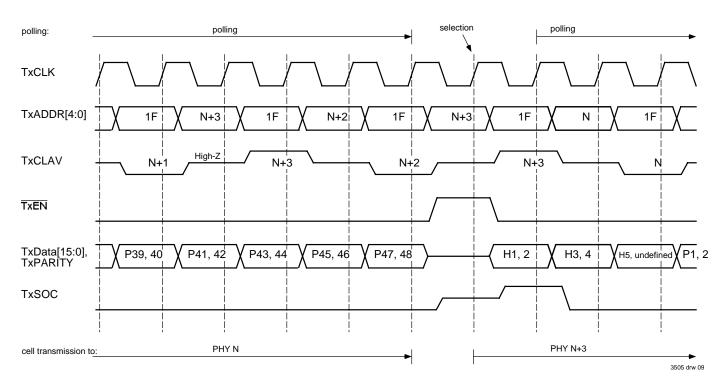


Figure 8. Utopia 2 Transmit Handshake - Back to Back Cells

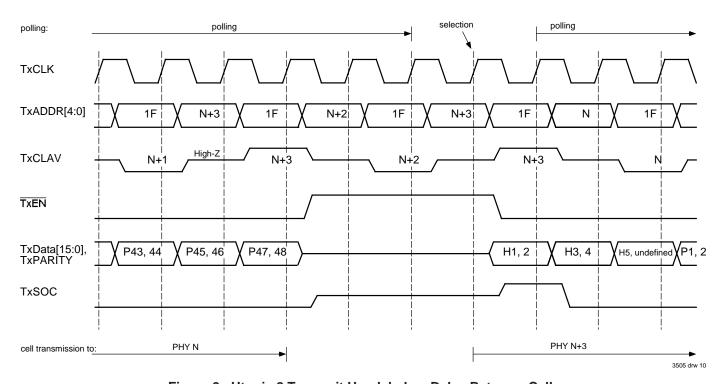


Figure 9. Utopia 2 Transmit Handshake - Delay Between Cells

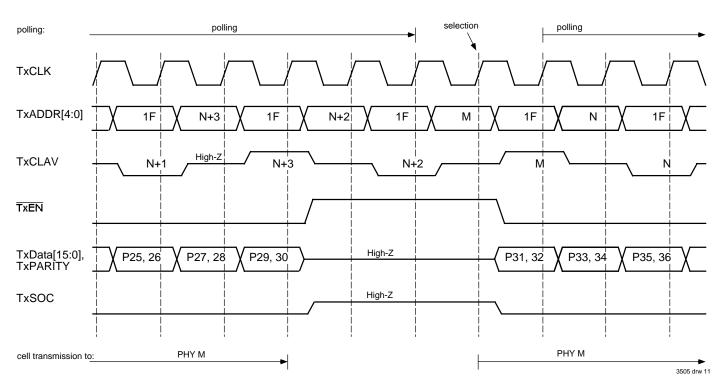


Figure 10. Utopia 2 Transmit Handshake - Transmission Suspended

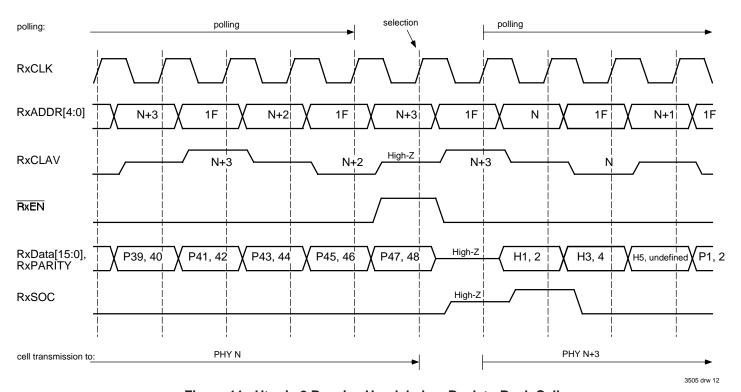


Figure 11. Utopia 2 Receive Handshake - Back to Back Cells

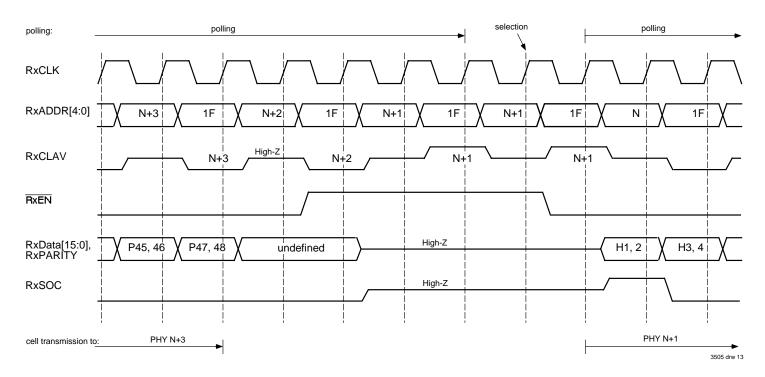


Figure 12. Utopia 2 Receive Handshake - Delay Between Cells

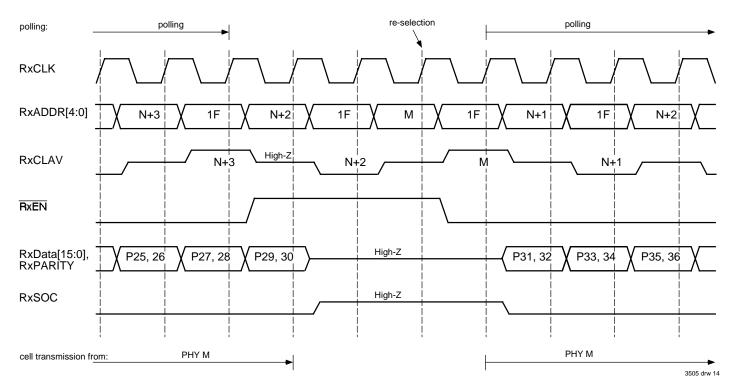


Figure 13. Utopia 2 Receive Handshake - Suspended Transfer of Data

#### **UTOPIA LEVEL 1 MULTI-PHY INTERFACE OPTION**

The UTOPIA Level 1 MULTI-PHY interface operates as defined in ATM Forum document af-phy-0017 and clarified in af-phy-0039. Utopia Level 1 is essentially the same as Utopia Level 2, but without the addressing, polling and selection features. Instead of addressing, it utilizes separate TxCLAV, TxEN, RxCLAV and RxEN signals for each channel of the 77V1254. There are just one each of the TxSOC and RxSOC signals, which are shared across all three channels.

In addition to Utopia Level 2's cell mode transfer protocol, Utopia Level 1 also offers the option of a byte mode protocol. Bit 1 of the Master Control Registers is used to program whether the UTOPIA Level 1 bus is in cell mode or byte mode. In byte mode, the PHY is allowed to control data transfer on a byte-by-byte basis via the TXCLAV and RXCLAV signals. In cell mode, TXCLAV and RXCLAV are ignored once the transfer of a cell has begun. In every other way the two modes are identical. Cell mode is the default configuration and is the one described later.

The Utopia 1 signals are summarized below:

TXDATA[7:0]	ATM to PHY
TXPARITY	ATM to PHY
TXSOC	ATM to PHY
TXEN[2:0]	ATM to PHY
TXCLAV[2:0]	PHY to ATM
TXCLK	ATM to PHY
RXDATA[7:0]	PHY to ATM
RXDATA[7:0] RXPARITY	PHY to ATM PHY to ATM
RXPARITY	PHY to ATM
RXPARITY RXSOC	PHY to ATM PHY to ATM

Transmit and receive both utilize free running clocks, which are inputs to the 77V1253. All Utopia signals are timed to these clocks.

In the transmit direction, the PHY first asserts TXCLAV (transmit cell available) to indicate that it has room in its

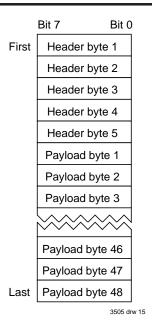


Figure 14. Utopia 1 Data Format and Sequence

transmit FIFO to accept at least one 53-byte ATM cell. When the ATM layer device is ready to begin passing the cell, it asserts  $\overline{\text{TXEN}}$  (transmit enable) and TXSOC (start of cell), coincident with the first byte of the cell on TXDATA.  $\overline{\text{TXEN}}$  remains asserted for the duration of the cell transfer, but the ATM device may deassert  $\overline{\text{TXEN}}$  at any time once the cell transfer has begun, but data is transfered only when  $\overline{\text{TXEN}}$  is asserted.

In the receive direction,  $\overline{\text{RXEN}}$  indicates when the ATM device is prepared to receive data. As with transmit, it may be asserted or deasserted at any time. The PHY asserts RXCLAV to indicate that it has an entire cell to transfer.

In both transmit and receive, TXSOC and RXSOC (start of cell) is asserted for one clock, coincident with the first byte of each cell. Odd parity is utilized across each 8-bit data field.

Figure 14 shows the data sequence for an ATM cell over Utopia Level 1, and Figures 15 to 21 are examples of the Utopia Level 1 handshake.

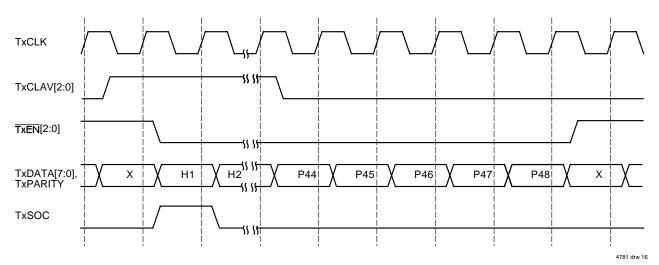


Figure 15. Utopia 1 Transmit Handshake - Single Cell

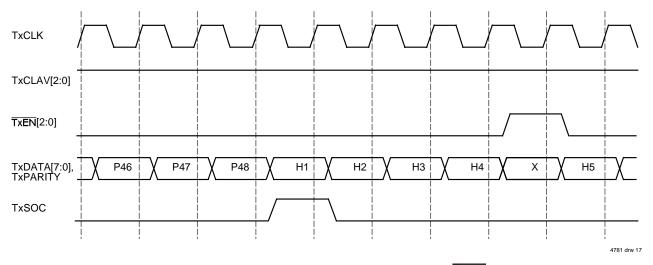


Figure 16. Utopia 1 Transmit Handshake - Back-to-Back Cells, and TXEN Suspended Transmission

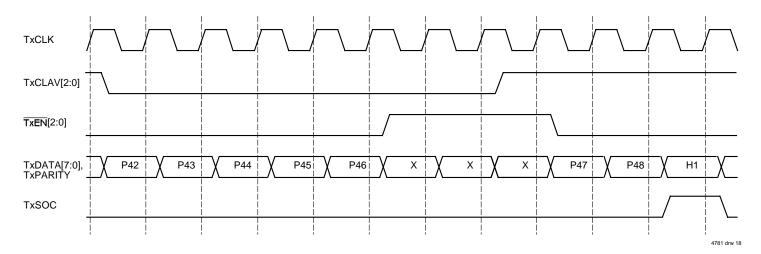


Figure 17. Utopia 1 Transmit Handshake - TXEN Suspended Transmission and Back-to-Back Cells (Byte Mode Only)

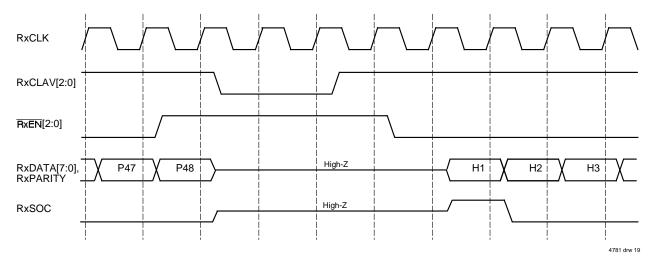


Figure 18. Utopia 1 Receive Handshake - Delay Between Cells

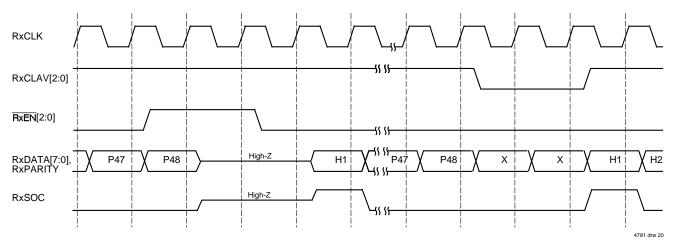


Figure 19. Utopia 1 Receive Handshake - RXEN and RXCLAV Control

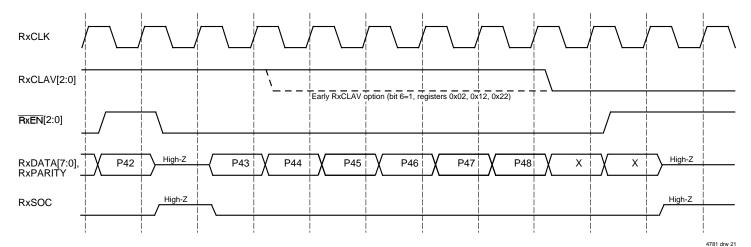


Figure 20. Utopia 1 Receive Handshake - RXCLAV Deassertion

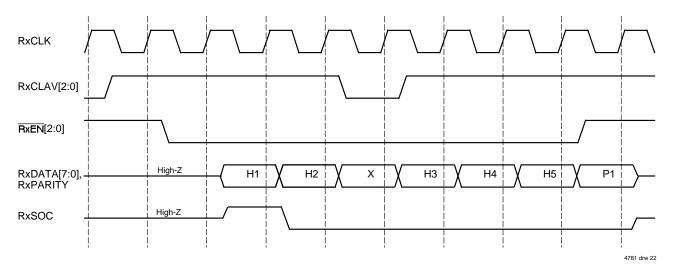


Figure 21. Utopia 1 Receive Handshake - RXCLAV Suspended Transfer (Byte Mode Only)

#### **DPI INTERFACE OPTION**

The DPI interface is relatively new and worth additional description. The biggest difference between the DPI configurations and the UTOPIA configurations is that each channel has its own DPI interface. Each interface has a 4-bit data path, a clock and a start-of-cell signal, for both the transmit direction and the receive direction. Therefore, each signal is point-to-point, and none of these signals has high-Z capability. Additionally, there is one master DPI clock input (DPICLK) into the 77V1253 which is used as a source for the DPI transmit clock outputs. DPI is a cell-based transfer scheme like Utopia Level 2, whereas UTOPIA Level 1 transfers can be either byte- or cell-based.

Another unique aspect of DPI is that it is a symmetrical interface. It is as easy to connect two PHYs back-to-back as it is to connect a PHY to a switch fabric chip. In contrast, Utopia is asymmetrical. Note that for the 77V1253, we are using the "transmit" and "receive" nomenclature in the naming of the DPI signals, whereas other devices may use more generic "in" and "out" nomenclature for their DPI signals.

The DPI signals are summarized below, where "Pn\_" refers to the signals for channel number "n":

DPICLK	input to PHY
Pn_TCLK	PHY to ATM
Pn_TD[3:0]	ATM to PHY
Pn_TFRM	ATM to PHY
Pn_RCLK	ATM to PHY
Pn_RD[3:0]	PHY to ATM
Pn_RFRM	PHY to ATM

In the transmit direction (ATM to PHY), the ATM layer device asserts start-of-cell signal (Pn\_TFRM) for one clock cycle, one clock prior to driving the first nibble of the cell on Pn\_TD[3:0]. Once the ATM side has begun sending a cell, it is prepared to send the entire cell without interruption. The 77V1253 drives the transmit DPI clocks (Pn\_TCLK) back to the ATM device, and can modulate (gap) it to control the flow of data. Specifically, if it cannot accept another nibble, the 77V1253 disables Pn\_TCLK and does not generate another rising edge until it has room for the nibble. Pn\_TCLK are derived from the DPICLK free running clock source.

The DPI protocol is exactly symmetrical in the receive direction, with the 77V1253 driving the data and start-of-cell signals while receiving Pn\_RCLK as an input.

The DPI data interface is four bits, so the 53 bytes of an ATM cell are transferred in 106 cycles. Figure 22 shows the sequence of that data transfer. Figures 23 through 30 show how the handshake operates.

	Bit 3 Bit 0	
First	Header byte 1, (8:5)	
	Header byte 1, (4:1)	
	Header byte 2, (8:5)	
	Header byte 2, (4:1)	
	Header byte 3, (8:5)	
	Header byte 3, (4:1)	
	Header byte 4, (8:5)	
	Header byte 4, (4:1)	
	Header byte 5, (8:5)	
	Header byte 5, (4:1)	
	Payload byte 1, (8:5)	
	Payload byte 1, (4:1)	
		1
	Payload byte 47, (8:5)	7
	Payload byte 47, (4:1)	
	Payload byte 48, (8:5)	
Last	Payload byte 48, (4:1)	
	3505 dr	w 23

Figure 22. DPI Data Format and Sequence

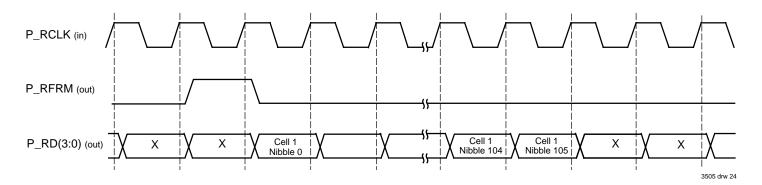


Figure 23. DPI Receive Handshake - One Cell Received

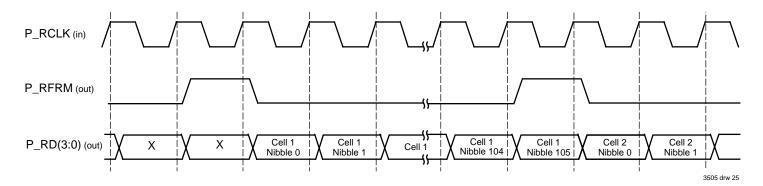


Figure 24. DPI Receive Handshake - Back-to-Back Cells

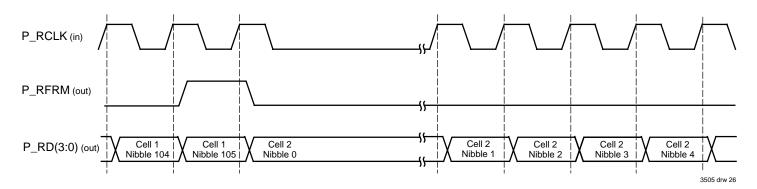


Figure 25. DPI Receive Handshake - ATM Layer Device Suspends Transfer

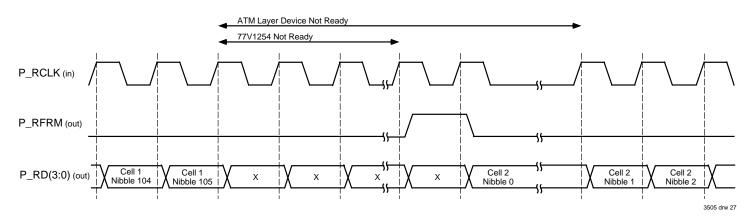


Figure 26. DPI Receive Handshake - Neither Device Ready

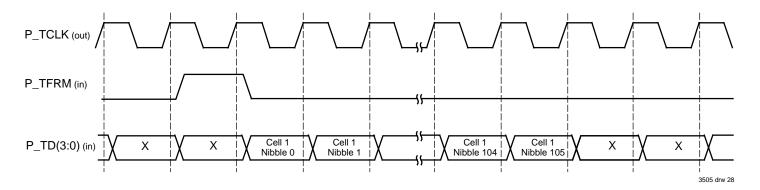


Figure 27. DPI Transmit Handshake - One Cell for Transmission

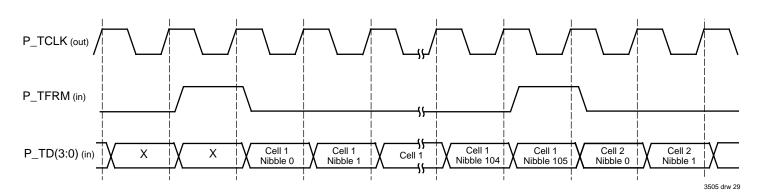


Figure 28. DPI Transmit Handshake - Back-to-Back Cells for Transmission

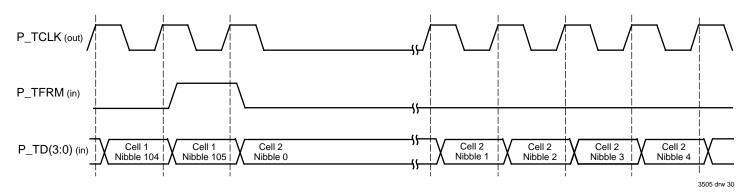


Figure 29. DPI Transmit Handshake - 77V1254 Transmit FIFO Full

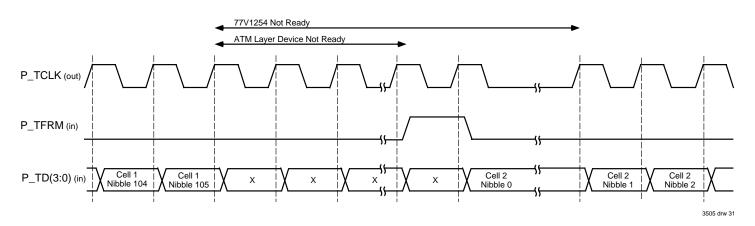


Figure 30. DPI Transmit Handshake - Neither Device Ready

## CONTROL AND STATUS INTERFACE UTILITY BUS

The Utility Bus is a byte-wide interface that provides access to the registers within the IDT77V1253. These registers are used to select desired operating characteristics and functions, and to communicate status to external systems.

The Utility Bus is implemented using a multiplexed address and data bus (AD[7:0]) where the register address is latched via the Address Latch Enable (ALE) signal.

The Utility Bus interface is comprised of the following pins: AD[7:0], ALE,  $\overline{CS}$ ,  $\overline{RD}$ ,  $\overline{WR}$ 

#### **Read Operation**

Refer to the Utility Bus timing waveforms in Figures 42 - 43. A register read is performed as follows:

- 1. Initial condition:
  - RD, WR, CS not asserted (logic 1)
  - ALE not asserted (logic 0)
- 2. Set up register address:
  - place desired register address on AD[7:0]
  - set ALE to logic 1;
  - latch this address by setting ALE to logic 0.
- 3. Read register data:
  - Remove register address data from AD[7:0]
  - assert  $\overline{\text{CS}}$  by setting to logic 0;
  - assert RD by setting to logic 0
- wait minimum pulse width time (see AC specifications)

#### **Write Operation**

A register write is performed as described below:

- 1. Initial condition:
  - RD, WR, CS not asserted (logic 1)
  - ALE not asserted (logic 0)
- 2. Set up register address:
  - place desired register address on AD[7:0]
  - set ALE to logic 1;
  - latch this address by setting ALE to logic 0.
- 3. Write data:
  - place data on AD[7:0]
  - assert  $\overline{\text{CS}}$  by setting to logic 0;
  - assert WR (logic 0) for minimum time

(according to timing specification); reset WR to logic 1 to complete register write cycle.

#### **INTERRUPT OPERATIONS**

The IDT77V1253 provides a variety of selectable interrupt and signalling conditions which are useful both during 'normal' operation, and as diagnostic aids. Refer to the Status and Control Register List section.

Overall interrupt control is provided via bit 0 of the Master Control Registers. When this bit is cleared (set to 0), interrupt signalling is prevented on the respective port. The Interrupt Mask Registers allow individual masking of different interrupt sources. Additional interrupt signal control is provided by bit 5 of the Master Control Registers. When this bit is set (=1),

receive cell errors will be flagged via interrupt signalling and all other interrupt conditions are masked. These errors include:

- Bad receive HEC
- Short (fewer than 53 bytes) cells
- Received cell symbol error

Normal interrupt operations are performed by setting bit 0 and clearing bit 5 in the Master Control Registers.  $\overline{\text{INT}}$  (pin 85) will go to a low state when an interrupt condition is detected. The external system should then interrogate the 77V1253 to determine which one (or more) conditions caused this flag, and reset the interrupt for further occurrences. This is accomplished by reading the Interrupt Status Registers. Decoding the bits in these bytes will tell which error condition caused the interrupt. Reading these registers also:

- clears the (sticky) interrupt status bits in the registers that are read
- resets INT

This leaves the interrupt system ready to signal an alarm for further problems.

#### LED CONTROL AND SIGNALLING

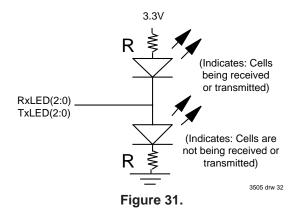
The LED outputs provide bi-directional LED drive capability of 8 mA. As an example, the RxLED outputs are described in the truth table:

STATE	PIN VOLTAGE
Cells being received	Low
Cells not being received	High

As illustrated in the following drawing (Figure 31), this could be connected to provide for a two-LED condition indicator. These could also be different colors to provide simple status indication at a glance. (The minimum value for R should be  $330\Omega$ , but a value closer to 1 k $\Omega$  is recommended).

#### **TxLED Truth Table**

STATE	PIN VOLTAGE
Cells being transmitted	Low
Cells not being transmitted	High



#### **DIAGNOSTIC FUNCTIONS**

#### 1. LOOPBACK

There are two loopback modes supported by the 77V1253. The loopback mode is controlled via bits 1 and 0 of the Diagnostic Control Registers:

Bit 1	Bit 0	MODE
0	0	Normal operating mode
1	0	PHY Loopback
1	1	Line Loopback

#### **Normal Mode**

This mode, Figure 32, supports normal operating conditions: data to be transmitted is transferred to the TC, where it is queued and formatted for transmission by the PMD. Receive data from the PMD is decoded along with its clock for transfer to the receiving "upstream system".

#### **PHY Loopback**

As Figure 33 illustrates below, this loopback mode provides a connection within the PHY from the transmit PHY-ATM interface to the PHY-ATM receive interface. Note that while this mode is operating, no data is forwarded to or received from the line interface.

#### Line Loopback

Figure 34 might also be called "remote loopback" since it provides for a means to test the overall system, including the line. Since this mode will probably be entered under direction from another system (at a remote location), receive data is also decoded and transferred to the upstream system to allow it to listen for commands. A common example would be a command asking the upstream system to direct the TC to leave this loopback state, and resume normal operations.

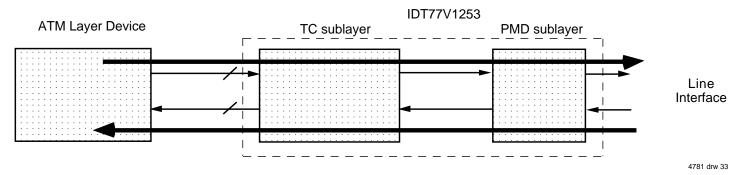


Figure 32. Normal Mode

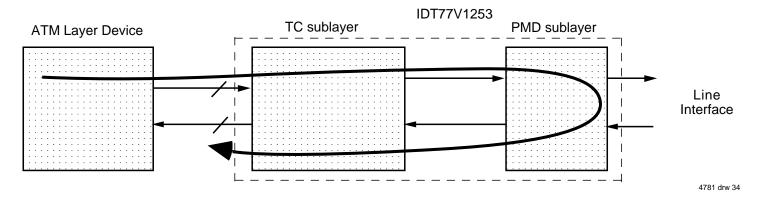


Figure 33. PHY Loopback

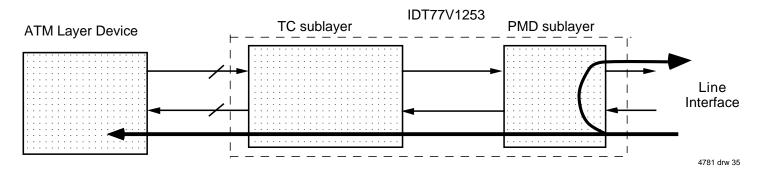


Figure 34. Line Loopback

#### 2. COUNTERS

Several condition counters are provided to assist external systems (e.g. software drivers) in evaluating communications conditions. It is anticipated that these counters will be polled from time to time (user selectable) to evaluate performance. A separate set of registers exists for each channel of the PHY.

- Symbol Error Counters
  - 8 bits
  - counts all invalid 5-bit symbols received
- Transmit Cell Counters
  - 16 bits
  - counts all transmitted cells
- Receive Cell Counters
  - 16 bits
  - counts all received cells, excluding idle cells and HEC errored cells
- Receive HEC Error Counters
  - 5 bits
  - counts all HEC errors received

The TxCell and RxCell counters are sized (16 bits) to provide a full cell count (without roll over) if the counter is read once/second. The Symbol Error counter and HEC Error counter were given sufficient size to indicate exact counts for low error-rate conditions. If these counters overflow, a gross condition is occurring, where additional counter resolution does not provide additional diagnostic benefit.

#### **Reading Counters**

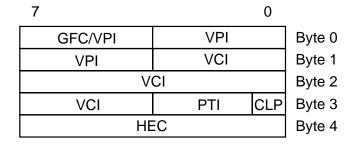
1. Decide which counter value is desired. Write to the Counter Select Register(s) (0x06, 0x16 and 0x26) to the bit location corresponding to the desired counter. This loads the High and Low Byte Counter Registers with the selected counter's value, and resets this counter to zero.

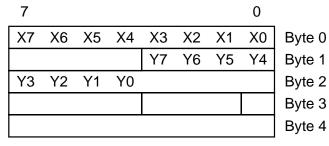
NOTE: Only one counter may be enabled at any time in each of the Counter Select Registers.

2. Read the Counter Registers (0x04, 0x14 or 0x24 (low byte)) and (0x05, 0x15 or 0x25 (high byte)) to get the value. Further reads may be accomplished in the same manner by writing to the Counter Select Registers.

#### VPI/VCI SWAPPING

For compatibility with IDT's SwitchStar products (77V400 and 77V500), the 77V1253 has the ability to swap parts of the VPI/VCI address space in the header of receive cells. This function is controlled by the VPI/VCI Swap bits, which are bit 5 of the Enhanced Control Registers (0x08, 0x18 and 0x28). The portions of the VPI/VCI that are swapped are shown below. Bits X(7:0) are swapped with Y(7:0) when the VPI/VCI Swap bit is set and the chip is in DPI mode.





3505 drw 51

## LINE SIDE (SERIAL) INTERFACE

Each of the four ports has two pins for differential serial transmission, and two pins for differential serial receiving.

#### PHY TO MAGNETICS INTERFACE

A standard connection to  $100\Omega$  and  $120\Omega$  unshielded twisted pair cabling is shown in Figure 35. Note that the transmit signal is somewhat attenuated in order to meet the launch amplitude specified by the standards. The receive circuitry is designed to attenuate low frequencies in order to

compensate for the high frequency attenuation of the cable.

Also, the receive circuitry biases the positive and negative RX inputs to slightly different voltages. This is done so that the receiver does not receive false signals in the absence of a real signal. This can be important because the 77V1253 does not disable error detection or interrupts when an input signal is not present.

When connecting to UTP at 51.2 Mbps, it is necessary to use magnetics with sufficient bandwidth. Such a device can also operate satisfactorily at 25.6 Mbps.

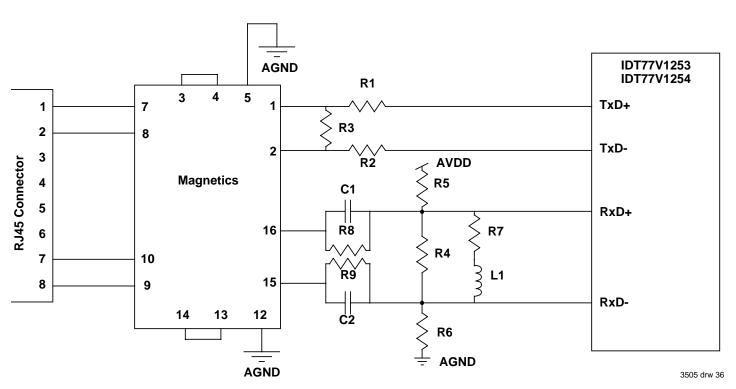


Figure 35. Recommended Connection to Magnetics

Table 3
Analog Component Values

Component	Value	Tolerance
R1	47Ω	±5%
R2	47Ω	±5%
R3	620Ω	±5%
R4	110Ω	±5%
R5	2700Ω	±5%
R6	2700Ω	±5%
R7	82Ω	±5%
R8	33Ω	±5%
R9	33Ω	±5%
C1	470pF	±20%
C2	470pF	±20%
L1	3.3μΗ	±20%

#### **Magnetics Modules for 25.6 Mbps**

Pulse PE-67583 or R4005	(619) 674-8100
TDK TLA-6M103	(847) 803-6100
Valor SF1153	(800) 318-2567

#### **Magnetics Modules for 51.2 Mbps**

Pulse R4005 (619) 674-8100

#### STATUS AND CONTROL REGISTER LIST

The 77V1253 has 28 registers that are accessible through the utility bus. Each of the three ports has 9 registers dedicated to that port. There is only one register (0x40) which is not port specific.

For those register bits which control operation of the Utopia interface, the operation of the Utopia interface is determined by the registers corresponding to the port which is selected at that particular time. For consistent operation, the Utopia control bits should be programmed the same for all three ports, except for the Utopia 2 port addresses in the Enhanced Control Registers.

		Register Ac	Idress	
Register Name	Port 0	Port 1	Port 2	All Ports
Master Control Registers	0x00	0x10	0x20	
Interrupt Status Registers	0x01	0x11	0x21	
Diagnostic Control Registers	0x02	0x12	0x22	
LED Driver and HEC Status/Control	0x03	0x13	0x23	
Low Byte Counter Register [7:0]	0x04	0x14	0x24	
High Byte Counter Register [15:8]	0x05	0x15	0x25	
Counter Registers Read Select	0x06	0x16	0x26	
Interrupt Mask Registers	0x07	0x17	0x27	
Enhanced Control Registers	0x08	0x18	0x28	
RXREF and TXREF Control Register				0x40

#### Nomenclature

"Reserved" register bits, if written, should always be written "0"

R/W = register may be read and written via the utility bus

R-only or W-only = register is read-only or write-only

sticky = register bit is cleared after the register containing it is read; all sticky bits are read-only

"0" = 'cleared' or 'not set'

"1" = 'set'

#### **MASTER CONTROL REGISTERS**

Addresses: 0x00, 0x10, 0x20

Bit	Туре	Initial State	Function	
7	R/W	0	Reserved	
6	R/W	1 = discard errored cells	Discard Receive Error Cells  On receipt of any cell with an error (e.g. short cell, invalid command mnemonic, receive HEC error (if enabled)), this cell will be discarded and will not enter the receive FIFO.	
5	R/W	0 = all interrupts	Enable Cell Error Interrupts Only If Bit 0 in this register is set (Interrupts Enabled), setting of this bit enables only "Received Cell Error" (as defined in bit 6) to trigger interrupt line.	
4	R/W	0 = disabled	Transmit Data Parity Check Directs TC to check parity of TxDATA against parity bit located in TxPARITY.	
3	R/W	1 = discard idle cells	Discard Received Idle Cells  Directs TC to discard received idle (VPI/VCI = 0) cells from PMD without signaling external systems.	
2	R/W	0 = not halted	Halt Tx Halts transmission of data from TC to PMD and forces both TxD signals low.	
1	R/W	0 = cell mode	UTOPIA Level 1 mode select: 0 = cell mode, 1 = byte mode. Not applicable for Utopia 2 or DPI modes.	
0	R/W	1 = enable interrupts	Enable Interrupt Pin (Interrupt Mask Bit) Enables interrupt output pin (pin 85). If cleared, pin is always high and interrupt is masked. If set, an interrupt will be signaled by setting the interrupt pin to "0".	

## **INTERRUPT STATUS REGISTERS**

Addresses: 0x01, 0x11, 0x21

Bit	Туре	Initial State	Function	
7		Reserved		
6	R	0 = Bad Signal	Good Signal Bit See definitions on pages 14 and 15 1 = Good Signal 0 = Bad Signal	
5	sticky	0	HEC error cell received Set when a HEC error is detected on a received cell.	
4	sticky	0	"Short Cell" Received Interrupt signal which flags received cells with fewer than 53 bytes. This condition is detected when receiving Start-of-Cell command bytes with fewer than 53 bytes between them.	
3	sticky	0	Transmit Parity Error  If Bit 4 of Register 0x00 / 0x10 / 0x20 is set (Transmit Data Parity Check), this interrupt flags a transmit data parity error condition. Odd parity is used.	
2	sticky	0	Receive Signal Condition change This interrupt is set when the received 'signal' changes either from 'bad to good' or from 'good to bad'.	
1	sticky	0	Received Symbol Error Set when an undefined 5-bit symbol is received.	
0	sticky	0	Receive FIFO Overflow Interrupt which indicates when the receive FIFO has filled and cannot accept additional data.	

## **DIAGNOSTIC CONTROL REGISTERS**

Addresses: 0x02, 0x12, 0x22

Bit	Туре	Initial State	Function		
7	R/W	0=normal	Force TxCLAV deassert (applicable only in Utopia 1 and 2 modes) Used during line loopback mode to prevent upstream system from continuing to send data to the 77V1253. Not applicable in DPI mode.		
6	R/W	0=UTOPIA	RxCLAV Operation Select (for Utopia 1 mode)  The UTOPIA standard dictates that during cell mode operation, if the receive FIFO no longer has a complete cell available for transfer from PHY, RxCLAV is deasserted following transfer of the last byte out of the PHY to the upsteam system. With this bit set, early deassertion of this signal will occur coincident with the end of Payload byte 44 (as in octet mode for TxCLAV). This provides early indication to the upstream system of this impending condition.  0 = "Standard UTOPIA RxCLAV"  1 = "Cell mode = Byte mode"		
5	R/W	1 = tri-state	Single/Multi-PHY configuration select (applicable and writable only in Utopia 1 mode)  0 = single: Never tri-state RxDATA, RxPARITY and RxSOC  1 = Multi-PHY mode: Tri-state RxDATA, RxPARITY and RxSOC when RxEN = 1		
4	R/W	0=normal	RFLUSH = Clear Receive FIFO This signal is used to tell the TC to flush (clear) all data in the receive FIFO. The TC signals this completion by clearing this bit.		
3	R/W	0=normal	Insert Transmit Payload Error Tells TC to insert cell payload errors in transmitted cells. This can be used to test error detection and recovery systems at destination station, or, under loopback control, at the local receiving station. This payoad error is accomplished by flipping bit 0 of the last cell payload byte.		
2	R/W	0=normal	Insert Transmit HEC Error Tells TC to insert HEC error in Byte 5 of cell. This can be used to test error detection and recovery systems in down-stream switches, or, under loopback control, the local receiving station. This HEC error is accomplished by flipping bit 0 of the HEC byte.		
1,0	R/W	00=normal	Loopback Control bit# 1 0 0 0 Normal mode (receive from network) 1 0 PHY Loopback 1 1 Line Loopback		

#### LED DRIVER AND HEC STATUS/CONTROL REGISTERS

Addresses: 0x03, 0x13, 0x23

Bit	Туре	Initial State	Function		
7		0	Reserved		
6	R/W	0 = enable checking	Disable Receive HEC Checking (HEC Enable)  When not set, the HEC is calculated on first 4 bytes of received cell, and compared against the 5th byte. When set ( = 1), the HEC byte is not checked.		
5	R/W	0 = enable calculate & replace	Disable Xmit HEC Calculate & Replace  When set, the 5th header byte of cells queued for transmit is not replaced with the HEC calculated across the first four bytes of that cell.		
4,3	R/W	00 = 1 cycle	RxREF Pulse Wide bit #	th Select  The select of the s	DSC cycles DSC cycles
2	R	1 = empty	FIFO Status	1=TxFIFO empty	0=TxFIFO not empty
1	R	0	TxLED Status	1=Cell Received	0=Cell NOT Received
0	R	0	RxLED Status	1=Cell Received	0=Cell NOT Received

## **LOW BYTE COUNTER REGISTERS [7:0]**

Addresses: 0x04, 0x14, 0x24

Bit	Туре	Initial State	Function
[7:0]	R	0x00	Provides low-byte of counter value selected via registers 0x06, 0x16, 0x26 and 0x36.

## **HIGH BYTE COUNTER REGISTERS [15:8]**

Addresses: 0x05, 0x15, 0x25

Bit	Туре	Initial State	Function
[7:0]	R	0x00	Provides high-byte of counter value selected via register 0x06, 0x16, 0x26 and 0x36.

## **COUNTER SELECT REGISTERS**

Addresses: 0x06, 0x16, 0x26

Bit	Туре	Initial State	Function
7	_	_	Reserved
6	_	_	Reserved
5	_	_	Reserved
4	_	_	Reserved
3	W	0	Symbol Error Counter
2	W	0	TxCell Counter
1	W	0	RxCell Counter
0	W	0	Receive HEC Error Counter

NOTE: Only one bit may set at any time for proper operation

#### INTERRUPT MASK REGISTERS

Addresses: 0x07, 0x17, 0x27

Bit	Туре	Initial State	Function
7		0	Reserved
6		0	Reserved
5	R/W	0 = interrupt enabled	HEC Error Cell
4	R/W	0 = interrupt enabled	Short Cell Error
3	R/W	0 = interrupt enabled	Transmit Parity Error
2	R/W	0 = interrupt enabled	Receive Signal Condition Change
1	R/W	0 = interrupt enabled	Received Cell Symbol Error
0	R/W	0 = interrupt enabled	Receive FIFO Overflow

Note: When set to "1", these bits mask the corresponding interrupts going to the interrupt pin ( $\overline{\text{INT}}$ ). When set to "0", the interrupts are unmasked. These interrupts correspond to the interrupt status bits in the Interrupt Status Registers.

#### **ENHANCED CONTROL REGISTERS**

Addresses: 0x08, 0x18, 0x28

Bit	Туре	Initial State	Function
7	R/W	0 = not reset	Individual Port Software Reset  1 = Reset. This bit is not self clearing; must write 0 to exit reset.
6	R/W	0 = OSC	Transmit Line Clock (or Loop Timing Mode) When set to 0, the OSC input is used as the transmit line clock. When set to 1, the recovered receive clock is used as the transmit line clock.
5	R/W	0 = no swap	VPI/VCI Swap DPI mode only. Receive direction only. See description on page 29.
4-0	R/W	Port 0 (Reg 0x08): 00000 Port 1 (Reg 0x18): 00001 Port 2 (Reg 0x28): 00010	Utopia 2 Port Address When operating in Utopia 2 Mode, these register bits determine the Utopia 2 port addresses.

## **RXREF AND TXREF CONTROL REGISTER**

Address: 0x40

Bit	Туре	Initial State	Function
7, 6	R/W	$00 = \overline{\text{RxREF0}} \text{ (Port 0)}$	RxREF Source Select Selects which of the three ports (0 to 2) is the source of RxREF.
5	R/W	0 = not reset	Master Software Reset 1 = Reset. This bit is not self clearing; must write 0 to exit reset.
4-3		0	Reserved
2-0	R/W	0000 = not looped	RXREF to TXREF Loop Select  When set to 0, TxREF is used to generate X_8 timing marker commands.  When set to 1, TxREF input is ignored, and received X_8 timing commands are looped back and added to the transmit stream of that same port. See Figure 6.  bit 2: port 2 bit 1: port 1 bit 0: port 0

## **ABSOLUTE MAXIMUM RATINGS(1)**

Symbol	Rating	Value	Unit
VTERM	Terminal Voltage with respect to GND	-0.5 to +5.5	V
TBIAS	Temperature Under Bias	-55 to +125	°C
Тѕтс	Storage Temperature	-55 to +125	°C
Іоит	DC Output Current	50	mA

NOTE:

3139 tbl 02

## RECOMMENDED DC OPERATING CONDITIONS

Symbol	Parameter	Min.	Тур.	Max.	Unit
VDD	Digital Supply Voltage	3.0	3.3	3.6	V
GND	Digital Ground Voltage	0	0	0	V
VIH	Input High Voltage	2.0	_	5.25	V
VIL	Input Low Voltage	-0.3	_	0.8	V
AVDD	Analog Supply Voltage	3.0	3.3	3.6	V
AGND	Analog Ground Voltage	0	0	0	V
VDIF	VDD - AVDD	-0.5	0	0.5	V

## RECOMMENDED OPERATING TEMPERATURE and SUPPLY VOLTAGE

Grade	Ambient Temperature	GND, AGND	VDD, AVDD
Commercial	0°C to 70°C	0V	3.3V± 0.3V
Industrial	-40°C to +85°C	0V	3.3V± 0.3V

## CAPACITANCE (TA = $+25^{\circ}$ C, f = 1MHz)

Symbol	Parameter	Conditions	Max.	Unit
CIN <sup>(1)</sup>	Input Capacitance	VIN = 0V	10	pF
Cio <sup>(1)</sup>	I/O Capacitance	VOUT = 0V	10	pF

NOTES:

3139 tbl 05

1. Characterized values, not currently tested.

## DC ELECTRICAL CHARACTERISTICS (All pins except TX+/- and RX+/-)

Symbol	Parameter	Test Conditions	Min.	Max	Unit
ILI	Input Leakage Current	$Gnd \le VIN \le VDD$	<b>-</b> 5	5	μΑ
ILO	I/O (as input) Leakage Current	$Gnd \le VIN \le VDD$	-10	10	μΑ
Vон1 <sup>(1)</sup>	Output Logic "1" Voltage	IOH = -2  mA, VDD = min.	2.4	_	V
VoH2 <sup>(2)</sup>	Output Logic "1" Voltage	IOH = -8  mA, VDD = min.	2.4	_	V
VoL <sup>(3)</sup>	Output Logic "0" Voltage	IOL = 8 mA, VDD= min.		0.4	V
IDD1 <sup>(4, 5)</sup>	Digital Power Supply Current - VDD	OSC = 32 MHz, all outputs unloaded	_	140	mA
IDD2 <sup>(5)</sup>	Analog Power Supply Current - AVDD	OSC = 32 MHz, all outputs unloaded		140	mA

#### NOTES:

- 1. For AD[7:0] pins only.
- 2. For all output pins except AD[7:0],  $\overline{INT}$  and TX+/-.
- 3. For all output pins except TX+/-.
- 4. Add 15mA for each TX+/- pair that is driving a load
- 5. Total supply current is the sum of IDD1 and IDD2.

## DC ELECTRICAL CHARACTERISTICS (TX+/- output pins only)

Symbol	Parameter	Test Conditions	Min.	Max	Unit
VOH1	Output Logic "1" Voltage	IOH = -20 mA	VDD - 0.5V		V
Vol	Output Logic "0" Voltage	IOL = 20 mA	_	0.5	V

Stresses greater than those listed under ABSOLUTE MAXIMUM RAT-INGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

## **UTOPIA Level 2 Bus Timing Parameters**

Symbol	Parameter	Min.	Max.	Units
t1	TxCLK Fequency	0.2	50	MHz
t2	TxCLK Duty Cycle (% of t1)	40	60	%
t3	TxDATA[15:0], TxPARITY Setup Time to TxCLK	7	_	ns
t4	TxDATA[15:0], TxPARITY Hold Time toTxCLK	2	_	ns
t5	TxADDR[4:0] Setup Time to TxCLK	7	_	ns
t6	TxADDR[4:0] Hold Time to TxCLK	2	_	ns
t7	TxSOC, TxEN Setup Time to TxCLK	7	_	ns
t8	TxSOC, TxEN Hold Time to TxCLK	2	_	ns
t9	TxCLK to TxCLAV High-Z	2	10	ns
t10	TxCLK to TxCLAV Low-Z (min) and Valid (max)	2	14	ns
t12	RxCLK Fequency	0.2	50	MHz
t13	RxCLK Duty Cycle (% of t12)	40	60	%
t14	RXEN Setup Time to RXCLK	7	_	ns
t15	RXEN Hold Time to RxCLK	2	_	ns
t16	RxADDR[4:0] Setup Time to RxCLK	7	_	ns
t17	RxADDR[4:0] Hold Time to RxCLK	2	_	ns
t18	RxCLK to RxCLAV High-Z	2	14	ns
t19	RxCLK to RxCLAV Low-Z (min) and Valid (max)	2	10	ns
t20	RxCLK to RxSOC High-Z	2	10	ns
t21	RxCLK to RxSOC Low-Z (min) and Valid (max)	2	14	ns
t22	RxCLK to RxDATA, RxPARITY High-Z	2	10	ns
t23	RxCLK to RxDATA, RxPARITY Low-Z (min) and Valid (max)	2	14	ns

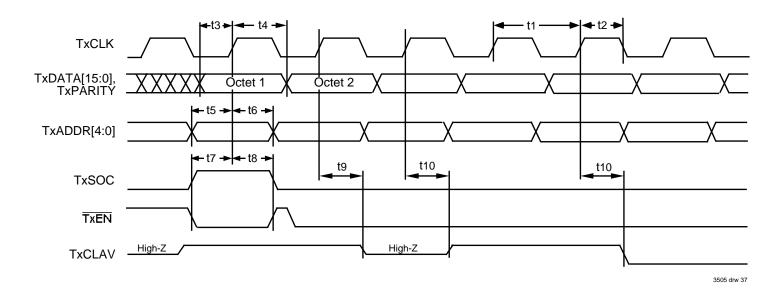


Figure 36. UTOPIA Level 2 Transmit

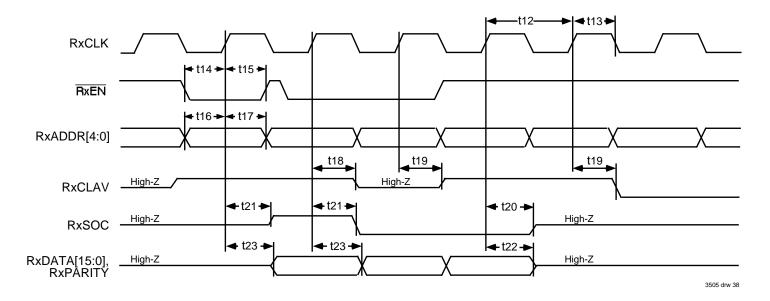


Figure 37. UTOPIA Level 2 Receive

## **UTOPIA Level 1 Bus Timing Parameters**

Symbol	Parameter	Min.	Max.	Units
t31	TxCLK Fequency	0.2	50	MHz
t32	TxCLK Duty Cycle (% of t31)	40	60	%
t33	TxDATA[7:0], TxPARITY Setup Time to TxCLK	7	_	ns
t34	TxDATA[7:0], TxPARITY Hold Time toTxCLK	2	_	ns
t35	TxSOC, TxEN[3:0] Setup Time to TxCLK	7	_	ns
t36	TxSOC, TxEN[3:0] Hold Time toTxCLK	2	_	ns
t37	TxCLK to TxCLAV[3:0] Invalid (min) and Valid (max)	2	14	ns
t39	RxCLK Fequency	0.2	50	MHz
t40	RxCLK Duty Cycle (% of t39)	40	60	%
t41	RxEN[3:0] Setup Time to RxCLK	7	_	ns
t42	RxEN[3:0] Hold Time to RxCLK	2	_	ns
t43	RxCLK to RxCLAV[3:0] Invalid (min) and Valid (max)	2	14	ns
t44	RxCLK to RxSOC High-Z	2	10	ns
t45	RxCLK to RxSOC Low-Z (min) and Valid (max)	2	14	ns
t46	RxCLK to RxDATA, RxPARITY High-Z	2	10	ns
t47	RxCLK to RxDATA, RxPARITY Low-Z (min) and Valid (max)	2	14	ns

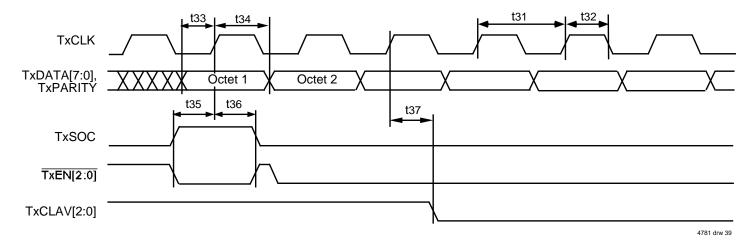


Figure 38. UTOPIA Level 1 Transmit

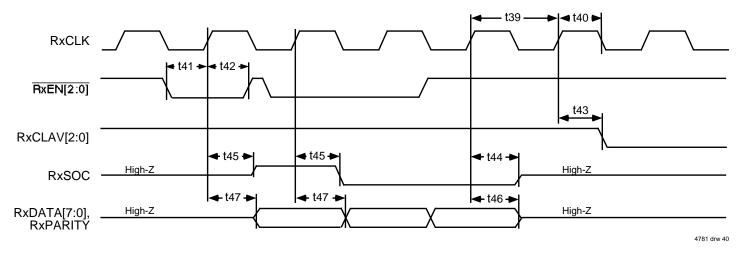


Figure 39. UTOPIA Level 1 Receive

## **DPI Bus Timing Parameters**

Symbol	Parameter	Min.	Max.	Units
t51	DPICLK Frequency	0.2	40	MHz
t52	DPICLK Duty Cycle (% of t51)	40	60	%
t53	DPICLK to Pn_TCLK Propagation Delay	2	14	ns
t54	Pn_TFRM Setup Time to Pn_TCLK	11	_	ns
t55	Pn_TFRM Hold Time to Pn_TCLK	1	_	ns
t56	Pn_TD[3:0] Setup Time to Pn_TCLK	11		ns
t57	Pn_TD[3:0] Hold Time to Pn_TCLK	1	_	ns
t61	Pn_RCLK Period	25	_	ns
t62	Pn_RCLK High Time	10		ns
t63	Pn_RCLK Low Time	10	_	ns
t64	Pn_RCLK to Pn_RFRM Invalid (min) and Valid (max)	2	12	ns
t65	Pn_RCLK to Pn_RD Invalid (min) and Valid (max)	2	12	ns

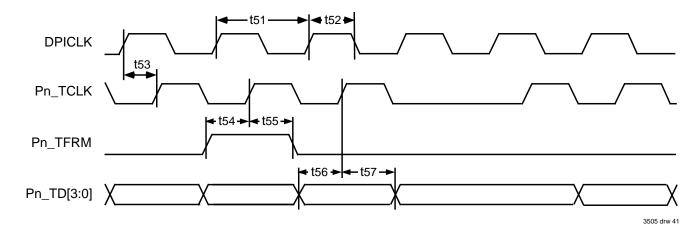


Figure 40. DPI Transmit

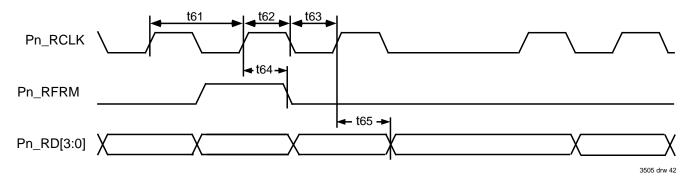


Figure 41. DPI Receive

## **Utility Bus Read Cycle**

Name	Min	Max	Unit	Description	
Tas	10		ns	Address setup to ALE	
Tcsrd	0		ns	Chip select to read enable	
Tah	5		ns	Address hold to ALE	
Tapw	10		ns	ALE min pulse width	
Ttria		0	ns	Address tri-state to RD assert	
Trdpw	20		ns	Min. RD pulse width	
Tdh	0	_	ns	Data Valid hold time	
Tch	0	_	ns	RD deassert to CS deassert	
Ttrid	_	10	ns	RD deassert to data tri-state	
Trd	5	18	ns	Read Data access	
Tar	5		ns	ALE low to start of read	
Trdd	0	_	ns	Start of read to Data low-Z	

## **Utility Bus Write Cycle**

Name	Min	Max	Unit	Description
Tapw	10	_	ns	ALE min pulse width
Tas	10	_	ns	Address set up to ALE
Tah	5	_	ns	Address hold time to ALE
Tcswr	0	_	ns	CS Assert to WR
Twrpw	20	_	ns	Min. WR pulse width
Tdws	20	_	ns	Write Data set up
Tdwh	10	_	ns	Write Data hold time
Tch	0	_	ns	WR deassert to CS deassert
Taw	20	_	ns	ALE low to end of write

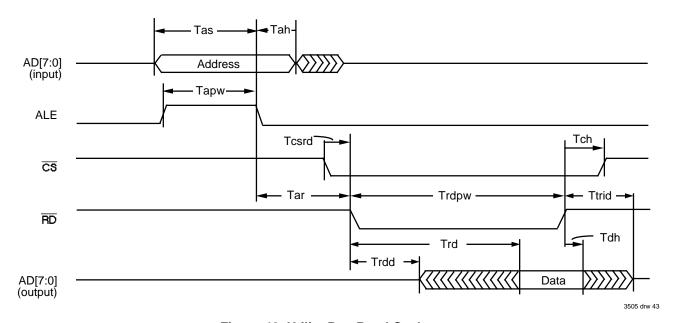


Figure 42. Utility Bus Read Cycle

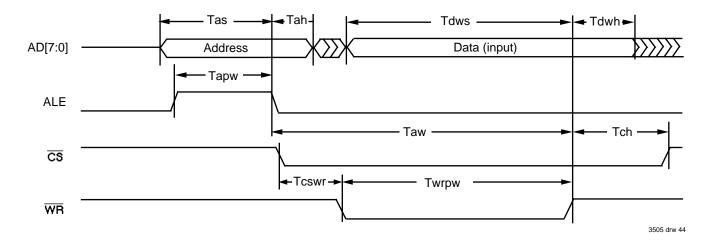


Figure 43. Utility Bus Write Cycle

## OSC, RXREF, TXREF and Reset Timing

Symbol	Parameter	Min.	Тур.	Max.	Unit
Тсус	OSC cycle period (25.6 Mbps) (51.2 Mbps)	30 15	31.25 15.625	33 16.5	ns ns
Tch	OSC high time	40	_	60	%
Tcl	OSC low time	40	_	60	%
Tcc	OSC cycle to cycle period variation	_	_	1	%
Trrpd <sup>(1)</sup>	OSC to RXREF Propagation Delay	1	_	30	ns
Ttrh	TXREF High Time	35	_	_	ns
Ttrl	TXREF Low Time	35	_	_	ns
Trspw <sup>(2)</sup>	Minimum RST Pulse Width	two TxCLK cycles two RxCLK cycles two DPICLK cycles two OSC cycles	_	_	_

#### NOTES:

- 1. The width of the RXREF pulse is programmable in the LED Driver and HEC Status/Control Registers.
- 2. The minimum RESET Pulse Width is either two RxCLK cycles, two TxCLK cycles, two DPICLK cycles or two OSC cycles, whichever is greater (and applicable).

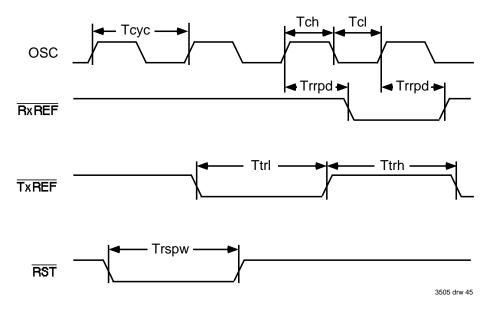


Figure 44. OSC, RXREF, TXREF and Reset Timing

#### **AC TEST CONDITIONS**

Input Pulse Levels	GND to 3.0V
Input Rise/Fall Times	3ns
Input Timing Reference Levels	1.5V
Output Reference Levels	1.5V
Output Load	See Figure 45

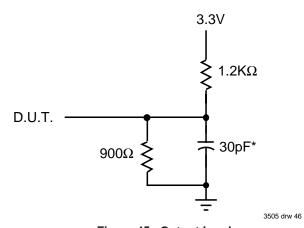
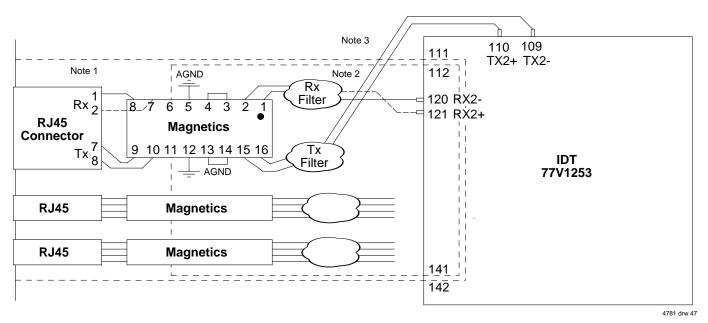


Figure 45. Output Load

\* Includes jig and scope capacitances.

A note about Figures 46 and 47: The ATM Forum and ITU-T standards for 25 Mbps ATM define "Network" and "User" interfaces. They are identical except that transmit and receive are switched between the two. A Network device can be connected directly to a User device with a straight-through cable. User-to-User or Network-to-Network connections require a cable with 1-to-7 and 2-to-8 crossovers.



#### Figure 46. PC Board Layout for ATM Network

- 1. No power or ground plane inside this area.
- 2. Analog power plane inside this area.

NOTES:

- 3. Digital power plane inside this area.
- 4. A single ground plane should extend over the area covered by the analog and digital power planes, without breaks.
- 5. All analog signal traces should avoid 90° corners.

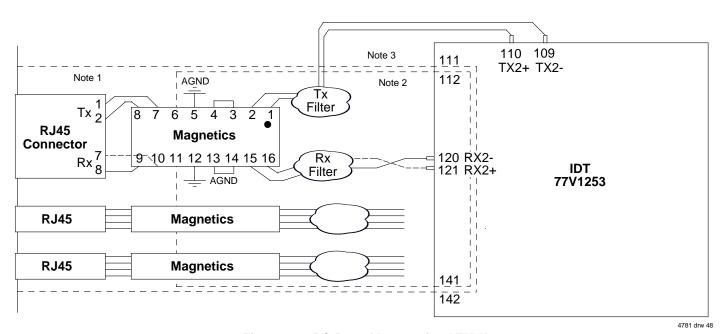
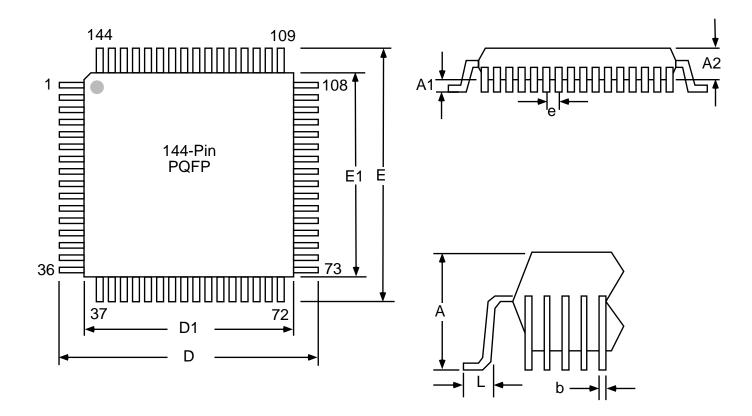


Figure 47. PC Board Layout for ATM User

#### NOTES:

- 1. No power or ground plane inside this area.
- 2. Analog power plane inside this area.
- 3. Digital power plane inside this area.
- 4. A single ground plane should extend over the area covered by the analog and digital power planes, without breaks.
- 5. All analog signal traces should avoid 90° corners.

## **PACKAGE DIMENSIONS**

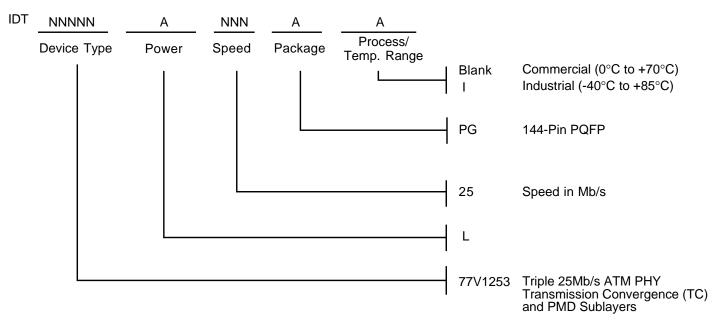


SYMBOL	MIN.	NOM.	MAX.
А	-	3.70	4.07
A1	0.25	0.33	-
A2	3.20	3.37	3.60
D	-	31.20	-
D1	-	28.00	-
Е	-	31.20	-
E1	-	28.00	-
L	0.73	0.88	1.03
е	-	0.65	-
b	0.22	-	0.38

Dimensions are in millimeters

3505 drw 49

#### ORDERING INFORMATION



4781 drw 50

#### PRELIMINARY DATASHEET: DEFINITION

"PRELIMINARY" datasheets contain descriptions for products soon to be, or recently, released to production, including features, pinouts and block diagrams. Timing data are based on simulation or initial characterization and are subject to change upon full characterization.

#### **Datasheet Document History**

11/30/98: PRELIMINARY. Initial draft.

Integrated Device Technology, Inc. reserves the right to make changes to the specifications in this data sheet in order to improve design or performance and to supply the best possible product.