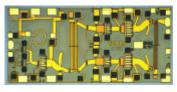


# Agilent HMMC-5033 17.7–32 GHz Power Amplifier

Data Sheet



Chip Size: Chip Size Tolerance: Chip Thickness: Pad Dimensions:  $\begin{array}{l} 2.74 \times 1.31 \text{ mm } (108 \times 51.6 \text{ mils}) \\ \pm 10 \ \mu\text{m} \ (\pm 0.4 \text{ mils}) \\ 127 \ \pm 15 \ \mu\text{m} \ (5.0 \ \pm \ 0.6 \text{ mils}) \\ \text{See Page } 6 \end{array}$ 

#### Features

•26 dBm Output P(-1dB)

- at 28 GHz
- •High Gain: 18 dB
- 50Ω Input/Output Matching
- Small Size
- RF Detector Network

#### Description

The HMMC-5033 is a MMIC power amplifier designed for use in wireless transmitters that operate within the 17.7 GHz to 32 GHz range. At 28 GHz it provides 26 dBm of output power  $(P_{-1dB})$  and 18 dB of small–signal gain from a small easy-to-use device. The HMMC-5033 was designed to be driven by the HMMC-5040 (20-40 GHz) or the HMMC-5618 (5.9-20 GHz) MMIC amplifier for linear transmit applications. This device has input and output matching circuitry for use in 50 ohm environments.

### Absolute Maximum Ratings<sup>[1]</sup>

Symbol	Parameters/Conditions	Min.	Max.	Units
V <sub>D1,2</sub>	Drain Supply Voltages		5.2	Volts
$V_{G1}, V_{GG}$	Gate Supply Voltages	-3.0	0.5	Volts
I <sub>D1</sub>	First Stage Drain Current		320	mA
I <sub>D2</sub>	Second Stage Drain Current		640	mA
P <sub>in</sub>	RF Input Power		23	dBm
Det.Bias	Applied Detector Bias (Optional)		5.2	Volts
T <sub>ch</sub>	Channel Temperature <sup>[2]</sup>		170	°C
T <sub>A</sub>	Backside Ambient Temperature	-55	+85	°C
T <sub>st</sub>	Storage Temperature	-65	+170	°C
T <sub>max</sub>	Maximum Assembly Temperature		300	°C

Notes:

1. Absolute maximum ratings for continuous operation unless otherwise noted.

2. Refer to *DC Specifications / Physical Properties* table for derating information.



Symbol	Parameters/Conditions	Min.	Тур.	Max.	Units
V <sub>D1</sub>	Drain Supply Operating Voltage		3.5	5	Volts
V <sub>D2</sub>	Drain Supply Operating Voltage		5	5	Volts
I <sub>D1</sub>	First Stage Drain Supply Current (V <sub>D1</sub> = 3.5 V, V <sub>G1</sub> = Open, V <sub>GG</sub> set for I <sub>D2</sub> typical)		240	320	mA
I <sub>D2</sub>	Second Stage Drain Supply Current ( $V_{D2} = 5 \text{ V}, V_{GG} \cong -0.8 \text{ V}$ )		460	640	mA
V <sub>G1</sub> ,V <sub>GG</sub>	Gate Supply Operating Voltages (I <sub>D1</sub> + I <sub>D2</sub> $\cong$ 700 mA)		-0.8		Volts
V <sub>P</sub>	Pinch-off Voltage [V <sub>DD</sub> = 2.5 V, (I <sub>D1</sub> + I <sub>D2</sub> ) $\leq$ 20 mA]	χ2.5	-1.2	-0.8	Volts
Det.Bias	Detector Bias Voltage (Optional)		V <sub>D2</sub>	5	Volts
$\theta_{1(ch-bs)}$	First Stage Thermal Resistance <sup>[2]</sup> (Channel–to–Backside at T <sub>ch</sub> = 160°C)		67		°C/Wat
$\theta_{2(ch-bs)}$	Second Stage Thermal Resistance <sup>[2]</sup> (Channel–to–Backside at T <sub>ch</sub> = 160°C)		37		°C/Wat
T <sub>ch</sub>	Second Stage Channel Temperature <sup>[3]</sup> (T <sub>A</sub> = 75°C, MTTF $\ge 10^6$ hrs, V <sub>D2</sub> = 5 V, I <sub>D2</sub> = 460 mA <sub>1</sub>		160		°C

# DC Specifications/Physical Properties<sup>[1]</sup>

1. Backside ambient operating temperature  $T_A = 25^{\circ}C$  unless otherwise noted. 2. Thermal resistance (°C/Watt) at a channel temperature T(°C) can be *estimated* using the equation:  $\theta(T) \cong q_{ch-bs} \notin [T(^{\circ}C)+273] / [160^{\circ}C+273].$ 3. Derate MTTF by a factor of two for every 8°C above  $T_{ch}$ .

## **RF Specifications**

 $(T_A = 25^{\circ}C, Z_0 = 50\Omega, V_{D1} = 3.5 \text{ V}, V_{D2} = 5 \text{ V}, I_{D2} = 460 \text{ mA} [I_{D1} \cong 240 \text{ mA}])$ 

Symbol	Parameters/Conditions	Lower Band Specifications		Mid Band Specifications		Upper Band Specifications		Units			
		Min.	Тур.	Max.	Min.	Тур.	Max.	Min.	Тур.	Max.	
BW	Operating Bandwidth	17.7		21	21		26.5	25		31.5	GHz
Gain	Small Signal Gain	17	22		17	20		15	18		dB
P <sub>-1dB</sub>	Output Power at 1dB Gain Compression	22	23		24	25		25	26		dBm
P <sub>SAT</sub>	Saturated Output Power <sup>[1]</sup>		25			27			28		dBm
(RL <sub>in</sub> ) <sub>MIN</sub>	Min. Input Return Loss	8	10		9	12		10	12		dB
(RL <sub>out</sub> ) <sub>MIN</sub>	Min. Output Return Loss	15	20		15	20		15	20		dB
Isolation	Min. Reverse Isolation		50			50			50		dB

Notes:

1. Devices operating continuously beyond 1 dB gain compression may experience power degradation.

#### Applications

The HMMC-5033 MMIC is a broadband power amplifier designed for use in transmitters that operate in various frequency bands between 17.7 GHz and 32 GHz. It can be attached to the output of the HMMC-5040 (20–40 GHz) or the HMMC-5618 (5.9–20 GHz) MMIC amplifier, increasing the power handling capability of transmitters requiring linear operation.

#### **Biasing and Operation**

The recommended DC bias condition for optimum efficiency, performance, and reliability is  $V_{D1} = 3.5$  volts and  $V_{D2} = 5$  volts with  $V_{GG}$  set for  $I_{D1} + I_{D2} = 700$  mA (no connection to  $V_{G1}$ ). This bias arrangement results in default drain currents  $I_{D1} = 240$  mA and  $I_{D2} = 460$  mA.

A single DC gate supply connected to  $V_{GG}$  will bias all gain stages.

If operation with both  $V_{D1}$  and  $V_{D2}$  at 5 volts is desired, an additional wire bond connection from the  $V_{G1}$  pad to the  $V_{GG}$  external bypass chip-capacitor (shorting  $V_{G1}$  to  $V_{GG}$ ) will balance the currents in each gain stage.  $V_{GG}$  (=  $V_{G1}$ ) can be adjust-

ed for  $I_{D1} + I_{D2} = 700$  mA.

Muting can be accomplished by setting  $V_{G1}$  and/or  $V_{GG}$  to the pinchoff voltage  $V_P$ .

An on chip RF output power detector network is provided. The differential voltage between the Det–Ref and Det–Out pads can be correlated with the RF power emerging from the RF Output port. Bias the diodes at ~200 mA.

The RF ports are AC–coupled at the RF input to the first stage and the RF output of the second stage.

If the output detector is biased using the on-chip optional Det-Bias network, an external ACblocking capacitor may be required at the RF Output port.

No ground wires are needed since ground connections are made with plated through-holes to the backside of the device.

#### Assembly Techniques

It is recommended that the electrical connections to the bonding pads be made using 0.7–1.0 mil diameter gold wire. The microwave/millimeter–wave connections should be kept as short as possible to minimize inductance. For assemblies requiring long bond wires, multiple wires can be attached to the RF bonding pads.

GaAs MMICs are ESD sensitive. ESD preventive measures must be employed in all aspects of storage, handling, and assembly.

MMIC ESD precautions, handling considerations, die attach and bonding methods are critical factors in successful GaAs MMIC performance and reliability.

Agilent application note #54, "GaAs MMIC ESD, Die Attach and Bonding Guidelines" provides basic information on these subjects.

#### Additional References:

AN# 52, "1 Watt 17.7 GHz– 32 GHz Linear Power Amplifier," and PN# 6, "HMMC-5033 Intermodulation Distortion."

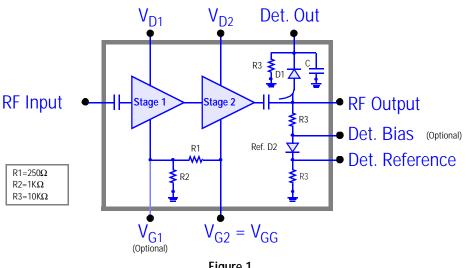
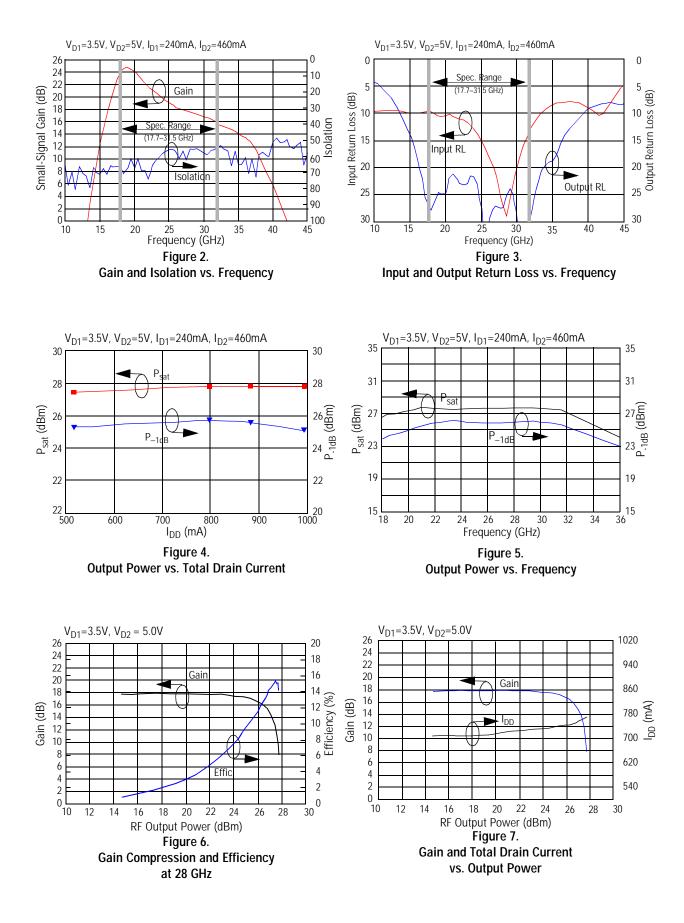


Figure 1. Simplified Schematic Diagram



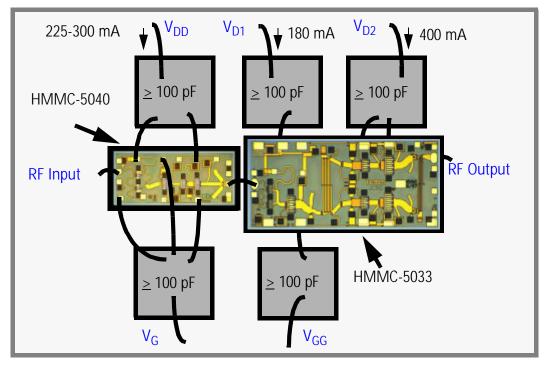


Figure 8. Assembly Diagram Illustrating the HMMC-5033 Cascaded with the HMMC-5040 for 20–32 GHz Applications

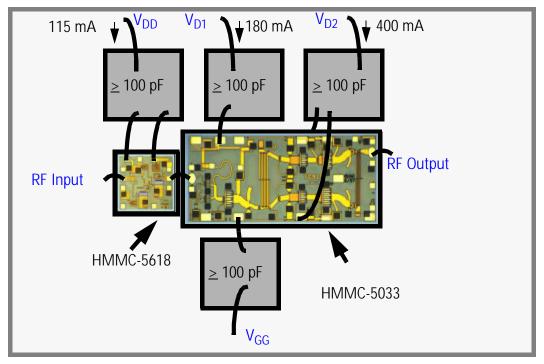


Figure 9. Assembly Diagram Illustrating the HMMC-5033 Cascaded with the HMMC-5618 for 17.7–20 GHz Applications

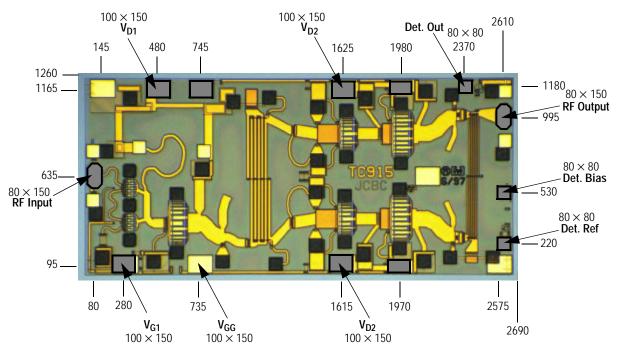


Figure 10. Bonding Pad Locations

This data sheet contains a variety of typical and guaranteed performance data. The information supplied should not be interpreted as a complete list of circuit specifications. In this data sheet the term *typical* refers to the 50th percentile performance. For additional information contact your local Agilent Technologies' sales representative.

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