# INTEGRATED CIRCUITS

# DATA SHEET



# OM6208 65 x 96 pixels matrix grey-scale LCD driver

Product specification Supersedes data of 2003 Jan 30 2003 feb 10





# 65 x 96 pixels matrix grey-scale LCD driver

OM6208

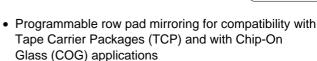
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### 1 FEATURES

- Single chip LCD Multiple Row Addressing (MRA) grey-scale/colour controller/driver
- · Four grey levels/colours
- 65 row outputs and 96 column outputs
- Display Data RAM (DDRAM)  $65 \times 96 \times 2$  bits
- · Selectable interface:
  - 6.5 MHz 3-line or 4-line Serial Peripheral Interface (SPI)
  - 6.5 MHz 3-line serial interface
  - High speed I<sup>2</sup>C-bus interface.
- · On-chip:
  - Configurable voltage multiplier generating  $V_{\text{LCD}}$ ; external  $V_{\text{LCD}}$  also possible
  - Four-segment V<sub>I CD</sub> temperature compensation
  - Generation of intermediate LCD bias voltage
  - Oscillator requires no external components; external clock input also possible
  - Integrated charge pump capacitors (reducing total system cost).
- External reset input
- · Temperature read-back
- Selectable N-line inversion and frame inversion
- CMOS compatible inputs
- Logic supply voltage range 1.7 to 3.3 V
- High-voltage generator supply voltage range 2.4 to 4.5 V
- Display supply voltage range 5 to 9 V
- Low power consumption; suitable for battery operated systems



- Status read which allows chip recognition
- Start address line; for example, for scrolling the displayed image
- Slim chip layout; suitable for COG, COF and TCP applications
- Operating temperature range -40 to +85 °C.

### 2 APPLICATIONS

- Telecom equipment
- Portable instruments
- · Point of sale terminals.

### 3 GENERAL DESCRIPTION

The OM6208 is a low power CMOS LCD controller driver, designed to drive a graphic display of 65 rows and 96 columns. All necessary functions for the display are provided in a single chip, including on-chip generation of LCD supply and bias voltages, resulting in a minimum of external components and low power consumption. The OM6208 can be interfaced to microcontrollers via a serial bus and I<sup>2</sup>C-bus.

The OM6208 is manufactured in n-well CMOS technology.

Operation is with the substrate at V<sub>SS</sub> potential.

### 4 ORDERING INFORMATION

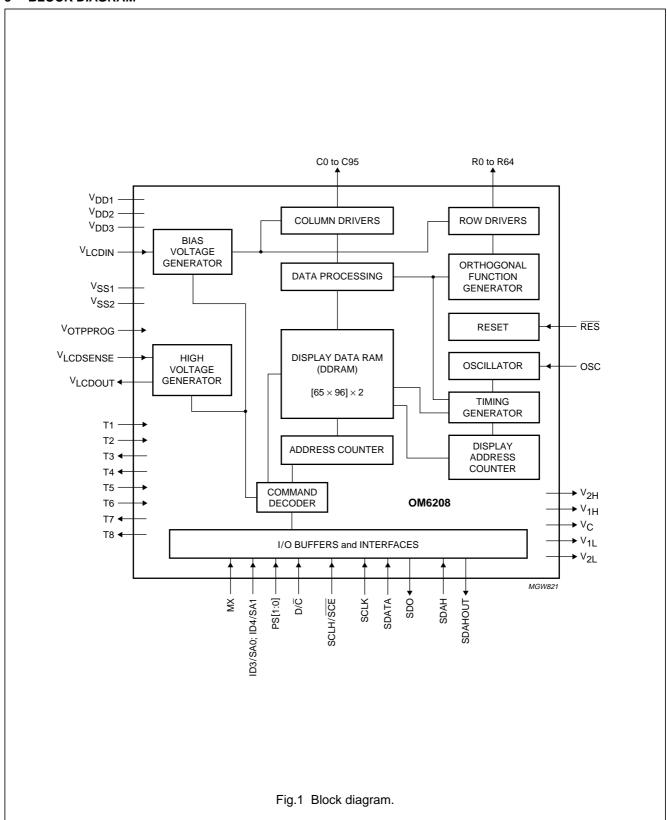
TYPE NUMBER		PACKAGE						
TIPE NUMBER	NAME	NAME DESCRIPTION						
OM6208MU/2DA/1	_	chip with bumps in tray	_					



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### **5 BLOCK DIAGRAM**



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### 6 PINNING

SYMBOL	<b>PAD</b> <sup>(1)</sup>	DESCRIPTION
V <sub>LCDIN</sub>	5 to 8	LCD supply voltage input; note 2
V <sub>LCDOUT</sub>	9 to 15	LCD supply voltage output from high voltage generator; note 2
V <sub>LCDSENSE</sub>	16	regulation input to high voltage generator; note 2
$V_{DD2}$	17 to 26	supply voltage 2; note 3
$V_{DD3}$	27 to 29	supply voltage 3; note 3
OSC	30	oscillator input; note 4
D/C	31	data/command input/output; note 5
PS[1:0]	32 and 33	interface selection inputs
V <sub>DD1</sub>	34 to 39	supply voltage 1; note 3
SDAHOUT	44	I <sup>2</sup> C-bus data output; note 6
SCLK	46	serial data clock input; used in 3-line or 4-line SPI or 3-line serial interface mode
SDAH	52	I <sup>2</sup> C-bus data input; note 7
SDO	53	serial data output; note 8
SDATA	54	serial data input; note 9
V <sub>SS2</sub>	55 to 61	ground 2 (analog ground); note 10
V <sub>SS1</sub>	62 to 67	ground 1 (digital ground); note 10
MX	68	horizontal mirroring input
Т3	69	test outputs; note 11
T4	70	
T1	71	
T2	72	
T5	73	
T6	74	
ID3/SA0; ID4/SA1	75 and 76	manufacturer device identification/I <sup>2</sup> C-bus slave address input pads; note 12
$V_{DD1}$	77	supply voltage 1 (tie-off pad)
SCLH/SCE	83	I <sup>2</sup> C-bus clock input/serial chip enable input in 3-line or 4-line SPI mode; note 5
V <sub>OTPPROG</sub>	84 to 86	supply voltage input for OTP programming; note 13
RES	88	external reset input; active low; must be applied to initialize the chip properly
R32 to R64	105 to 137	LCD row driver outputs
V <sub>C</sub>	138	bias buffer output; note 14
Т8	141	test outputs; note 11
T7	142	
V <sub>1L</sub>	143	bias buffer output; note 14
C95 to C48	144 to 191	LCD column driver outputs
C47 to C0	194 to 241	
R31 to R0	247 to 278	LCD row driver outputs
V <sub>1H</sub>	244	bias buffer outputs; note 14
V <sub>2L</sub>	245	
V <sub>2H</sub>	246	

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### **Notes**

1. Dummy pads are located at positions 1, 2, 4, 40 to 43, 45, 47 to 51, 78 to 82, 87, 89 to 92, 95 to 104, 139, 140, 192, 193, 242, 243, 279 and 280; alignment marks are located at positions 3 and 93; an alignment bump is located at position 94.

- 2. Positive power supply for the liquid crystal display (see also Figs 38, 39 and 40):
  - a) If the internal voltage generator is used, pads V<sub>LCDIN</sub>, V<sub>LCDSENSE</sub> and V<sub>LCDOUT</sub> must be connected together.
  - b) An external LCD supply voltage can be incorporated using the V<sub>LCDIN</sub> pad; the internal voltage generator must then switched off, pad V<sub>LCDOUT</sub> must be open-circuit (not connected to pad V<sub>LCDIN</sub>) and pad V<sub>LCDSENSE</sub> connected to the V<sub>LCDIN</sub> input; V<sub>DD2,3</sub> should be applied according to the specified voltage range. In Power-down mode, the external LCD supply voltage must be switched off.
- 3. V<sub>DD2</sub> and V<sub>DD3</sub> supply the internal voltage generator, both have the same voltage and may be connected together outside of the chip; V<sub>DD1</sub> supplies the remainder of the chip. V<sub>DD1</sub>, V<sub>DD2</sub> and V<sub>DD3</sub> can be connected together but then care must be taken with respect to the supply voltage range.
- 4. When the on-chip oscillator is used, the OSC input must be connected to V<sub>DD1</sub>. If an external clock signal is used, then this is connected to the OSC input. If both the oscillator and external clock are inhibited by connecting pad OSC to V<sub>SS1</sub>, the display is not clocked and may be in a DC state. To avoid this, the chip should always be put into Power-down mode before stopping the clock.
- This input is not used with the 3-line serial interface and must be connected to V<sub>DD1</sub> or V<sub>SS1</sub> when this interface is in use.
- 6. SDAHOUT is the serial data acknowledge output from the I<sup>2</sup>C-bus interface. By connecting SDAHOUT to SDAH externally, the SDAH line becomes fully I<sup>2</sup>C-bus compatible. Having the acknowledge output separated from the serial data line is advantageous in COG applications because here the track resistance from the SDAHOUT pad to the system SDAH line can be significant and a potential divider can be generated by the bus pull-up resistor and the ITO track resistance. It is possible that during the acknowledge cycle the OM6208 will not be able to create a valid logic 0. By splitting the SDAH input from the SDAHOUT output the device could be used in a mode that ignores the acknowledge bit. Therefore in COG applications where the acknowledge cycle is required, it is necessary to minimize the track resistance from the SDAHOUT pad to the system SDAH line to guarantee a valid logic 0. When SDAHOUT is not used, it must be connected to V<sub>DD1</sub> or V<sub>SS1</sub>.
- 7. When I<sup>2</sup>C-bus is not used, this pad must be connected to V<sub>DD1</sub> or V<sub>SS1</sub>.
- 8. SDO is a push-pull output; when it is intended to use the readback function of the OM6208, this pad must be connected to the SDATA pad, or used separately; when I<sup>2</sup>C-bus interface is selected, this pad should be connected to V<sub>DD1</sub> or V<sub>SS1</sub>.
- 9. When  $I^2C$ -bus interface is selected this pin should be connected to  $V_{DD1}$  or  $V_{SS1}$ .
- 10. Supply rails  $V_{SS1}$  and  $V_{SS2}$  must be connected together.
- 11. Test pads T1 to T8 are not accessible to users: T1, T2, T5 and T6 must be connected to V<sub>SS</sub>; T3, T4, T7 and T8 must be open-circuit.
- 12. Module identification bits: these bits may be read back via the 'read back' instruction; when the I<sup>2</sup>C-bus interface is being used, these bits are the two LSBs of the slave address.
- 13. V<sub>OTPROG</sub> can be connected to SCLH/SCE pad to reduce the external connections. If not connected in this configuration, then V<sub>OTPROG</sub> should be open-circuit during normal operation.
- 14. These pads are not accessible to users and must be left open-circuit; an explanation of the bias buffer function is given in Section 11.9.

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### 7 FUNCTIONAL DESCRIPTION

### 7.1 I/O buffers and interfaces

One of four industrial standard interfaces can be selected using the interface configuration inputs PS1 and PS0.

Table 1 Serial/I<sup>2</sup>C-bus interface selection

PS1	PS0	SELECTED INTERFACE
0	0	3-line SPI
0	1	4-line SPI
1	0	I <sup>2</sup> C-bus interface
1	1	3-line serial interface

### 7.2 Oscillator

The on-chip oscillator provides the clock signal for the display system. No external components are required when the internal oscillator is used. An external clock signal, if used, is connected to this input.

### 7.3 Address counter

The Address Counter (AC) assigns addresses to the display data RAM for writing. The X address X[6:0] and the Y address Y[4:0] are set separately.

### 7.4 Display data RAM

The OM6208 contains a  $65 \times 96 \times 2$  bit static RAM which stores the display data. The display data RAM is divided into 17 banks of 96 bytes, although only two bits of the 17th bank are used. During RAM access, data is transferred to the RAM via the serial interface. There is a direct correspondence between X address and column output number.

### 7.5 Display address counter

The display is generated by simultaneously reading out the RAM content for two or four rows, depending on the current display size that is selected. This content will be processed with the corresponding set of two or four orthogonal functions and so generate the signals for switching the pixels of the display on or off according to the RAM content.

The display status (all dots on/all dots off and normal/inverse video) is set by the bits DON, DAL and E in the command Display control (see Table 8).

### 7.6 Timing generator

The timing generator produces the various signals required to drive the internal circuitry. Internal chip operation is not affected by operations on the data bus.

### 7.7 Data processing

The data processing block receives data from the RAM and the orthogonal function from the logic circuits, then selects the correct voltage level to be provided to the columns.

### 7.8 High voltage generator

The high voltage generator provides the programmed  $V_{LCD}$  to the bias voltage generator block.

### 7.9 Bias voltage generator

The bias voltage generator generates all the voltage levels required for the MRA driving system.

### 7.10 Command decoder

The command decoder identifies command words arriving at the interface and routes the data bytes that follow to their destination.

### 7.11 Orthogonal function generator

The orthogonal function generator generates a set of orthogonal functions suitable for the selected value of p (number of active rows).

### **7.12** Reset

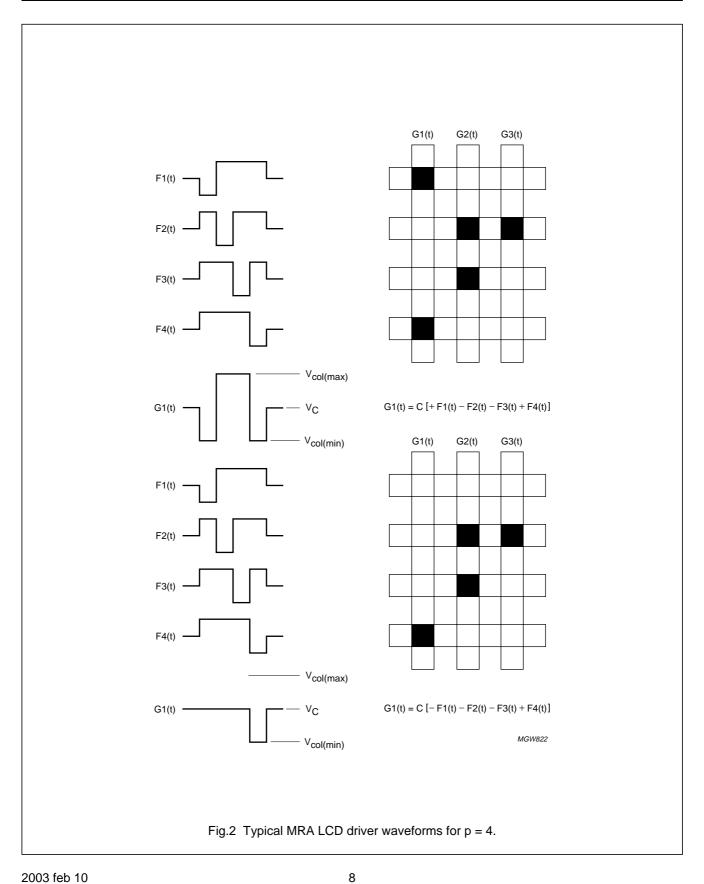
The reset block handles the hardware reset input ( $\overline{RES}$ ) and software reset and provides all internal blocks with the required reset signal.

### 7.13 Row drivers and column drivers

The OM6208 contains 65 row and 96 column drivers which connect the appropriate LCD bias voltages in sequence to the display in accordance with the data to be displayed. A typical MRA driving scheme with waveforms for p=4 is shown in Fig.2. The value of p represents the number of simultaneously selected rows.

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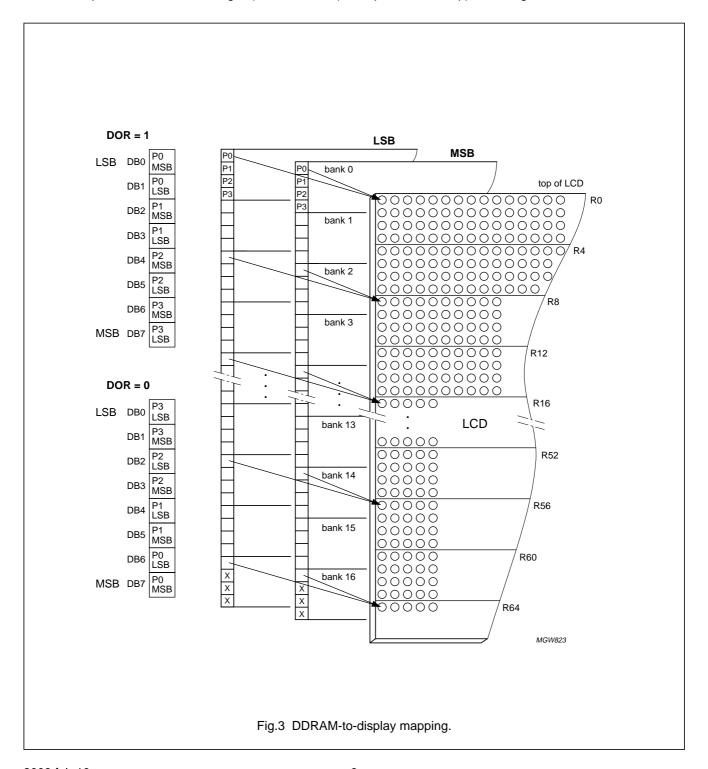
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### 8 RAM ADDRESSING

Data is downloaded in bytes into the RAM matrix of the OM6208 as indicated in Fig.3. The display RAM has a matrix of  $65 \times 96 \times 2$  bits. The columns are addressed by the address pointer. The address ranges (decimal values)

are X = 0 to 95 and Y = 0 to 16. The Y address represents the bank number. Addresses outside these ranges are not allowed.

The Data Order Bit (DOR) defines the bit order (LSB on top or MSB on top) for writing into the RAM.



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### 8.1 Display data RAM structure

The mode for storing data in the display data RAM is dependent on:

- Horizontal/vertical addressing mode set by bit V in the 'RAM addressing mode' instruction
- Data order set by bit DOR in the 'data order' instruction
- · Mirror the X-axis set by input MX.

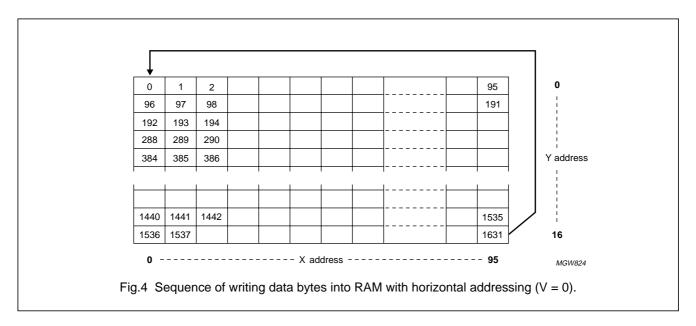
### 8.1.1 HORIZONTAL/VERTICAL ADDRESSING

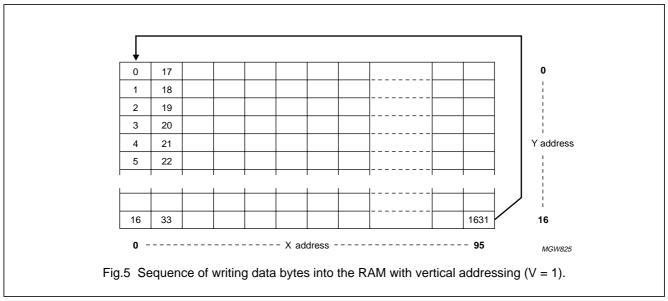
Two different addressing modes are possible; horizontal addressing mode and vertical addressing mode.

In the horizontal addressing mode (V = 0) the X address increments after each byte. After the last X address (X = 95), X wraps around to 0 and Y increments to address the next row (see Fig.4).

In the vertical addressing mode (V = 1), the Y address increments after each byte. After the last Y address (Y = 16), Y wraps around to 0 and X increments to address the next column (see Fig.5).

After the very last address, the address pointers wrap around to address X = 0 and Y = 0 in both horizontal and vertical addressing modes.





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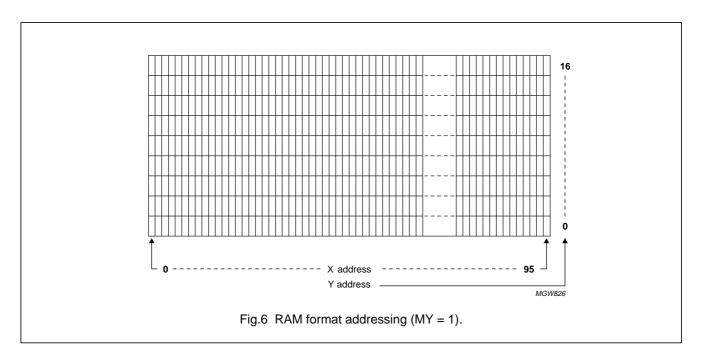
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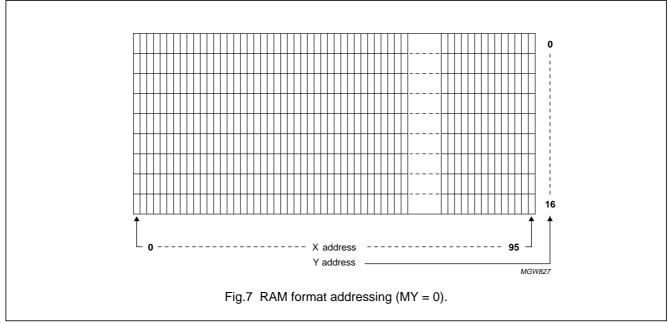
### 8.1.2 MIRROR Y

The Mirror Y (MY) bit allows vertical mirroring:

- When MY = 1, the Y address space is mirrored; the address Y = 0 is then located at the bottom of the display (see Fig.6)
- When MY = 0, the mirroring is disabled and the address Y = 0 is located at top of the display (see Fig.7).

Refer also to Section 11.6.





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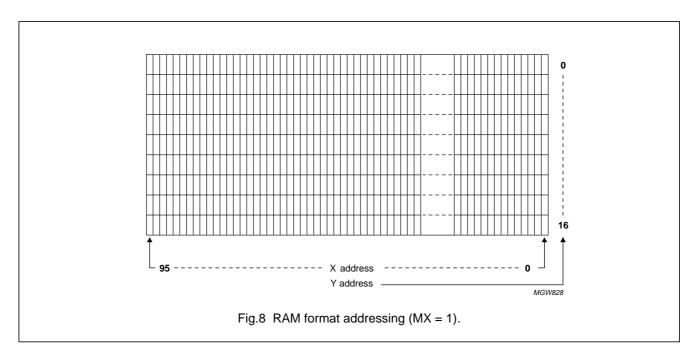
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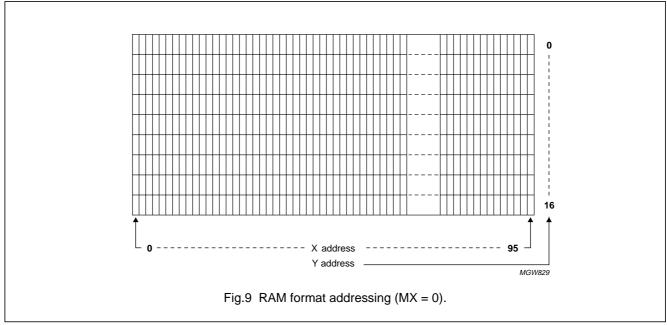
### 8.1.3 MIRROR X

The Mirror X (MX) input allows a horizontal mirroring:

- When MX = 1, the X address space is mirrored; the address X = 0 is then located at the right side (X<sub>max</sub>) of the display (see Fig.8)
- When MX = 0, the mirroring is disabled and the address X = 0 is located at the left side (column 0) of the display (see Fig.9).

Refer also to Section 11.6.





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### 9 SERIAL INTERFACING

Communication with the microcontroller can occur via a clock-synchronized serial peripheral interface. It is possible to select two different 3-line (SPI and serial interface) or a 4-line SPI interface. Selection is done via the PS[1:0] inputs.

### 9.1 Serial peripheral interface

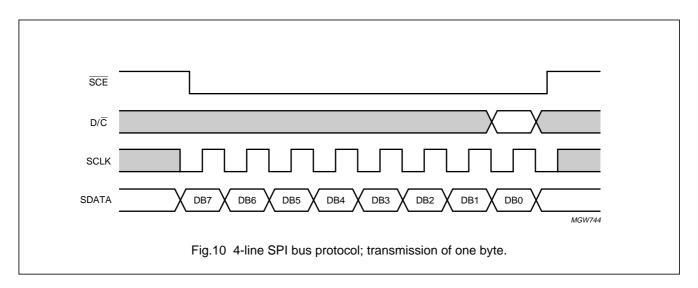
The Serial Peripheral Interface (SPI) is a 3-line or 4-line interface for communication between the microcontroller and the LCD driver chip. Three lines are common to both 3-line and 4-line SPI, these are  $\overline{SCE}$  (chip enable), SCLK (serial clock) and SDATA (serial data). For the 4-line SPI a separate D/ $\overline{C}$  line is added. The OM6208 is connected to the serial data I/O of the microcontroller by pads SDATA (data input) and SDO (data output) connected together.

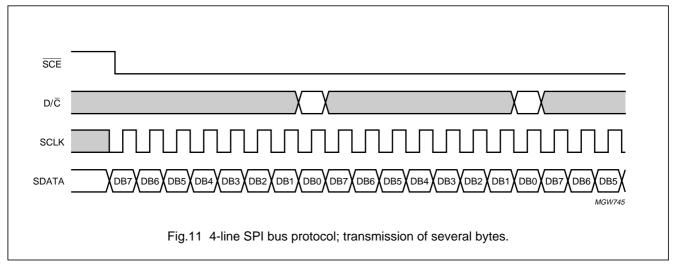
### 9.1.1 WRITE MODE

The display data/command indication may be controlled by software or by the  $D/\overline{C}$  select pin. When the  $D/\overline{C}$  pad is used, display data is transmitted when  $D/\overline{C}$  is HIGH, and command data is transmitted when  $D/\overline{C}$  is LOW (see Figs 10 and 11). When  $D/\overline{C}$  is not used, the 'display data length' instruction is used to indicate that a specific number of display data bytes (1 to 255) are to be transmitted (see Fig.11). The next byte after the display data string is handled as an instruction command.

When the 3-line SPI interface is used the display data/command is controlled by software (see Fig.12).

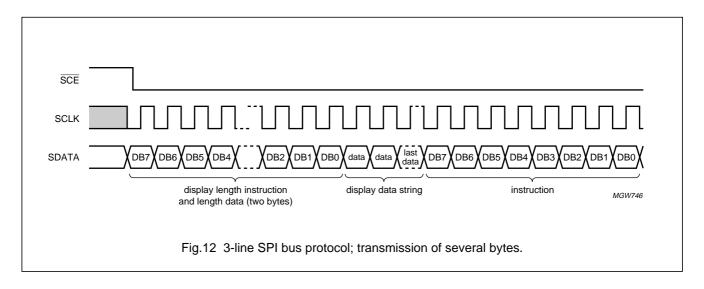
If SCE is pulled high during a serial display data stream, the interrupted byte is invalid data but all previously transmitted data is valid. The next byte received will be handled as an instruction command (see Fig.13).

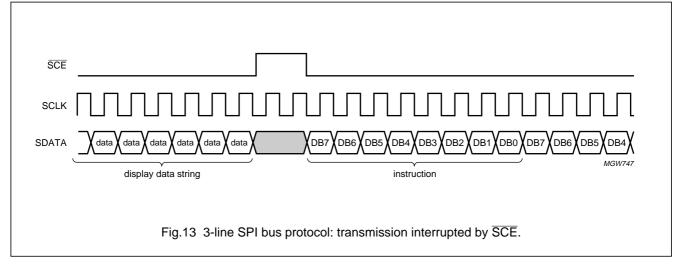




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### 9.1.2 READ MODE

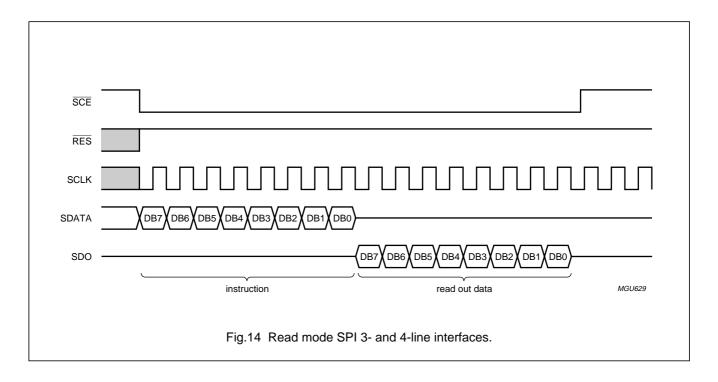
The read mode of the interface means that the microcontroller reads data from the OM6208. To do so the microcontroller first has to send a command, the read status command, and then OM6208 will respond by transmitting data on the SDO line. After that,  $\overline{SCE}$  is required to go HIGH (see Fig.14).

The OM6208 samples the SDATA data at rising SCLK edges, but shifts SDO data at falling SCLK edges. Thus

the SDO data is available to be read by the microcontroller at rising SCLK edges.

After the read status command has been sent, the SDATA line must be set to 3-state (high-impedance) not later than at the falling SCLK edge of the last bit (see Fig.14).

For the read data format, see Section 9.2.3; the serial interface timing diagram is given in Chapter 15.



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### 9.2 Serial interface (3-line)

The serial interface is also a 3-line bidirectional interface for communication between the microcontroller and the LCD driver chip. The three lines are  $\overline{SCE}$  (chip enable), SCLK (serial clock) and SDATA (serial data). The OM6208 is connected to the SDA of the microcontroller by the SDATA (data input) and SDO (data output) pads which are connected together.

### 9.2.1 WRITE MODE

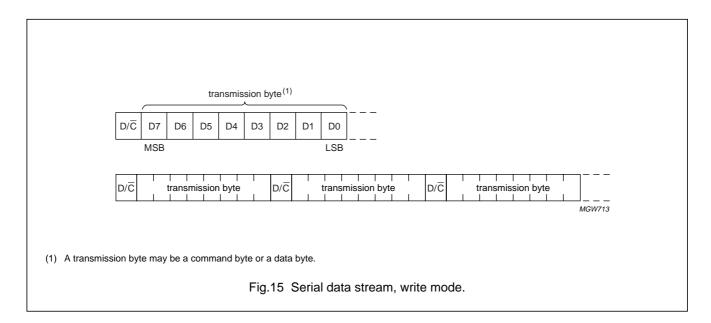
The write mode of the interface means that the microcontroller writes instructions and data to the OM6208. Each data packet contains a control bit  $D/\overline{C}$  and a transmission byte. If  $D/\overline{C}$  is LOW, the following byte is interpreted as command byte. The instruction set is given in Table 7. If  $D/\overline{C}$  is HIGH, the following byte is stored in the display data RAM. After every data byte the address counter is incremented automatically. The general format of the write mode and the definition of the transmission byte is shown in Fig.15.

Any instruction can be sent in any order to the OM6208. The MSB is transmitted first. The serial interface is initialized when  $\overline{SCE}$  is HIGH. In this state, SCLK clock

pulses have no effect and no power is consumed by the serial interface. A falling edge on  $\overline{SCE}$  enables the serial interface and indicates the start of data transmission.

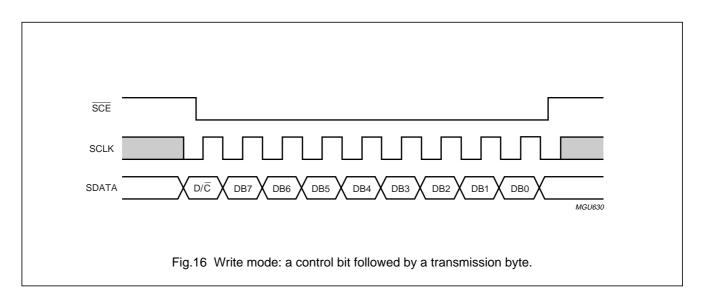
Figures 16, 17 and 18 show the protocol of the write mode:

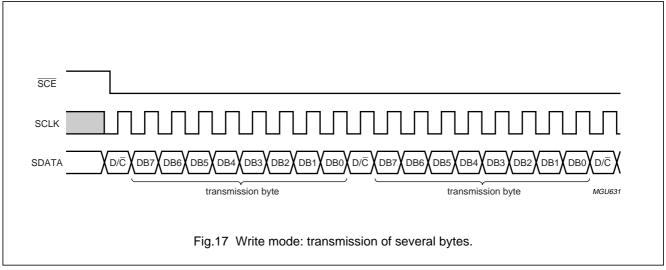
- When SCE is HIGH, SCLK clocks are ignored. During the HIGH time of SCE the serial interface is initialized (see Fig.16)
- At the falling SCE edge, SCLK must be LOW (see Fig.32)
- SDATA is sampled at the rising edge of SCLK
- D/C indicates whether the byte is a command (D/C = 0) or RAM data (D/C = 1) byte; it is sampled with the first rising SCLK edge
- If SCE stays LOW after the last bit of a command/data byte, the serial interface is ready for the D/C bit of the next byte at the next rising edge of SCLK (see Fig.17)
- A reset pulse with RES interrupts the transmission. The data being written into the RAM may be corrupted. The registers are cleared. If SCE is LOW after the rising edge of RES, the serial interface is ready to receive the D/C bit of a command/data byte (see Fig.18).

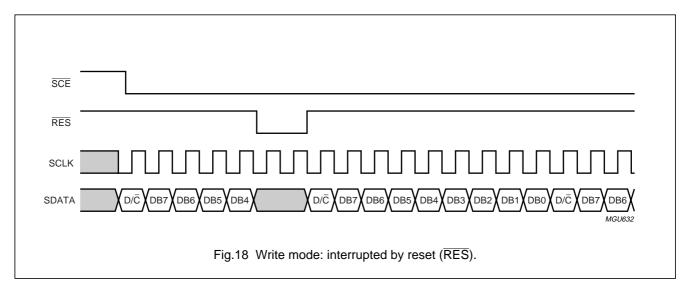


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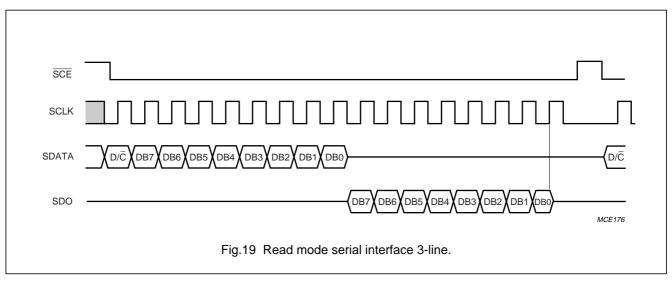




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### 9.2.2 READ MODE



The read mode of the interface means that the microcontroller reads data from the OM6208. To do so the microcontroller first has to send a command, the read status command, and then the following byte is transmitted in the opposite direction using SDO (see Fig.19). After that,  $\overline{SCE}$  is required to go HIGH before a new command is sent.

The OM6208 samples the SDATA data at rising SCLK edges and shifts SDO data at falling SCLK edges. Thus the SDO data is available for the microcontroller to read at rising SCLK edges.

After the read status command has been sent, the SDATA line must be set to 3-state not later then at the falling SCLK edge of the last bit (see Fig.19).

The 8th read bit is shorter than the others because it is terminated by the rising SCLK edge (see Fig.35). The last rising SCLK edge sets SDO to 3-state after the delay time  $t_4$ .

### 9.2.3 READ DATA FORMAT

Regardless of which serial interface is used there are five bits that can be read (ID1 to ID4 and VM) and one temperature register. For the bits, one bit is transmitted per byte read and is selected by issuing the appropriate read instruction from the instruction set. Bits ID1 and ID2 are hard-wired so that ID1 always returns a logic 0 and ID2 always returns a logic 1. Bits ID3 and ID4 are the identification bits and are set via ID3/SA0 and ID4/SA1 pads. The format for the read bit, B, is shown in Table 2.

Table 2 Read data format

D7 (MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)
x <sup>(1)</sup>	В	В	В	B	B	B	B

### Note

1. x = undefined.

Table 3 Read temperature sensor

Sending the instruction to read back the temperature sensor data will select the following status byte.

D7 (MSB)	SB) D6 D5		D4	D3	D2	D1	D0 (LSB)	
x <sup>(1)</sup>	TD[6]	TD[5]	TD[4]	TD[3]	TD[2]	TD[1]	TD[0]	

### Note

1. x = undefined.

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### 10 I2C-BUS INTERFACE

### 10.1 Characteristics of the I<sup>2</sup>C-bus (Hs-mode)

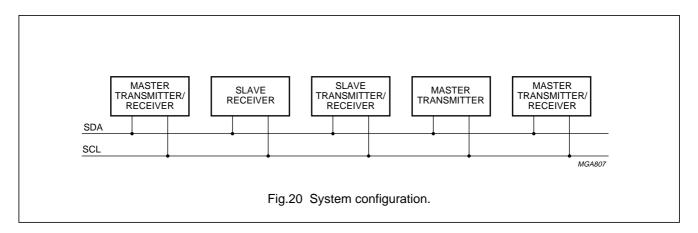
The I<sup>2</sup>C-bus Hs-mode is for bidirectional, two-line communication between different ICs or modules with speeds up to 3.4 MHz. The only difference between Hs-mode slave devices and F/S-mode slave devices is the speed at which they operate, therefore the buffers on the SDAH output have an open drain. This is the same for I<sup>2</sup>C-bus master devices which have an open-drain SDAH output and a combination of an open-drain pull-down and current source pull-up circuits on the SCLH output. Only the current source of one master is enabled at any one time and only during Hs-mode. Both lines must be connected to a positive supply via a pull-up resistor.

Data transfer may be initiated only when the bus is not busy.

### 10.1.1 System configuration

Definition (see Fig.20):

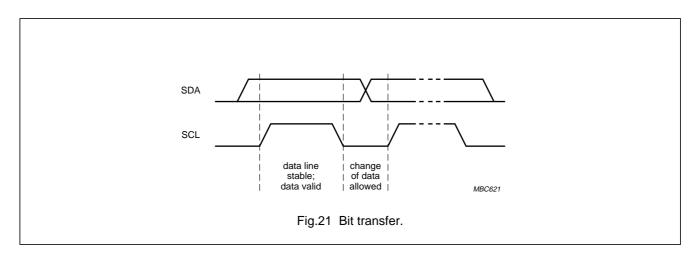
- Transmitter: the device that sends the data to the bus
- Receiver: the device that receives the data from the bus
- Master: the device that initiates a transfer, generates clock signals and terminates a transfer
- Slave: the device addressed by a master
- Multi-master: more than one master can attempt to control the bus at the same time without corrupting the message
- Arbitration: procedure to ensure that, if more than one master simultaneously tries to control the bus, only one is allowed to do so and the message is not corrupted
- Synchronisation: procedure to synchronize the clock signals of two or more devices.



### 10.1.2 BIT TRANSFER

One data bit is transferred during each clock pulse (see Fig.21). The data on the SDAH line must remain stable

during the HIGH period of the clock pulse as changes in the data line at this time will be interpreted as a control signal.



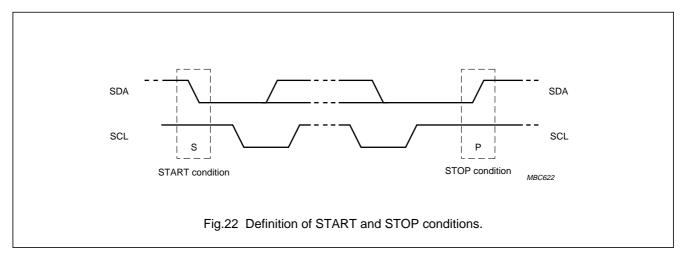
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### 10.1.3 START AND STOP CONDITIONS

Both data and clock lines remain HIGH when the bus is not busy (see Fig.22). A HIGH-to-LOW transition of the data

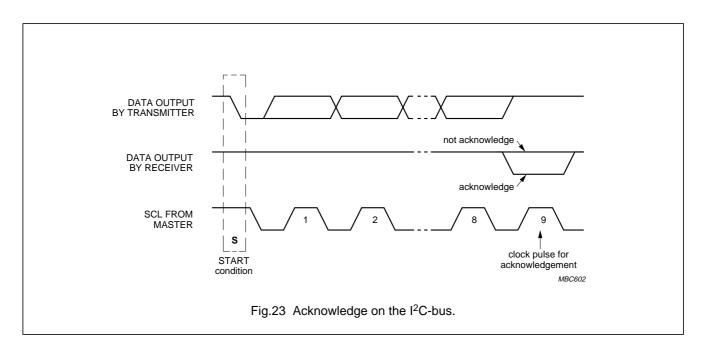
line, while the clock is HIGH is defined as the START condition (S). A LOW-to-HIGH transition of the data line while the clock is HIGH is defined as the STOP condition (P).



### 10.1.4 ACKNOWLEDGE

Each byte of 8 bits is followed by an acknowledge bit (see Fig.23). The acknowledge bit is a HIGH signal put on the bus by the transmitter during which time the master generates an extra acknowledge related clock pulse. A slave receiver which is addressed must generate an acknowledge after the reception of each byte. Also a master receiver must generate an acknowledge after the reception of each byte that has been clocked out of the

slave transmitter. The device that acknowledges must pull-down the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during the HIGH period of the acknowledge related clock pulse (set-up and hold times must be taken into consideration). A master receiver must signal an end of data to the transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this event the transmitter must leave the data line HIGH to enable the master to generate a stop condition.



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### 10.2 I<sup>2</sup>C-bus Hs-mode protocol

The OM6208 is a slave receiver/transmitter. If data is to be read from the device the SDAHOUT and SDAH pads must be connected for acknowledge to be used (see Table 1, note 6).

Hs-mode can only commence after the following conditions.

- START condition (S)
- 8-bit master code (00001XXX)
- not-acknowledge bit (A).

The master code has two functions as shown in Figs 24 and 25, it allows arbitration and synchronization between competing masters at F/S-mode speeds, resulting in one winner. Also the master code indicates the beginning of an Hs-mode transfer.

As no device is allowed to acknowledge the master code, then a master code transmission must be followed by a not-acknowledge  $(\overline{A})$ . After this  $\overline{A}$  bit, and the SCLH line has been pulled up to a HIGH level, the active master switches to Hs-mode and enables at  $t_H$  the current-source pull-up circuit for the SCLH signal (see Fig.25).

The active master will then send a repeated START condition (Sr) followed by a 7-bit slave address with a

 $R/\overline{W}$  bit, and receives an acknowledge bit (A) from the selected slave. After each acknowledge bit (A) or not-acknowledge bit  $(\overline{A})$  the active master disables its current-source pull-up circuit. The active master re-enables its current source again when all devices have released and the SCLH signal reaches a HIGH level. The rising of the SCLH is done by a resistor pull-up and so is slower, the last part of the SCLH rise time is speeded up because the current source is enabled. Data transfer only switches back to F/S-mode after a STOP (P) condition.

The write sequence that occurs after the Hs-mode is selected is shown in Fig.26. The sequence is initiated with a START (S) condition from the I<sup>2</sup>C-bus master which is followed by the slave address. All slaves with the corresponding address acknowledge in parallel, all the others will ignore the I<sup>2</sup>C-bus transfer.

After an acknowledgement cycle of a write  $(\overline{W})$ , one or more command words follow which define the status of the addressed slaves. A command word consists of a control byte, which defines Co and  $D/\overline{C}$ , plus a data byte (see Fig.26 and Table 4).

The last control byte is tagged with a cleared most significant bit, the continuation bit Co. The control and data bytes are also acknowledged by all addressed slaves on the bus.

Table 4 Co and Sr definition

Со	D/C	R/W	ACTION
0	_	_	last control byte to be sent; only a stream of data bytes are allowed to follow; this stream may only be terminated by a STOP or repeated START condition
1	_	_	another control byte will follow the data byte unless a STOP or repeated START condition is received
	0	0	data byte will be decoded and used to set up the device
_		1	data byte will return the status byte
	1	0	data byte will be stored in the display RAM
_	-   1		RAM read back is not supported

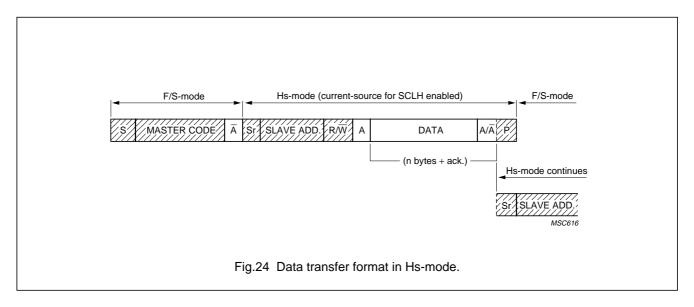
After the last control byte, depending on the  $D/\overline{C}$  bit setting, a series of display data bytes or command data bytes may follow. If the Sr bit was set to logic 1, these display bytes are stored in the display RAM at the address specified by the data pointer. The data pointer is automatically updated and the data is directed to the intended OM6208 device. If the Sr bit of the last control byte was set to logic 0, these command bytes will be decoded and the setting of the device will be changed according to the received commands. The acknowledgement after each byte is made only by the addressed OM6208. At the end of the

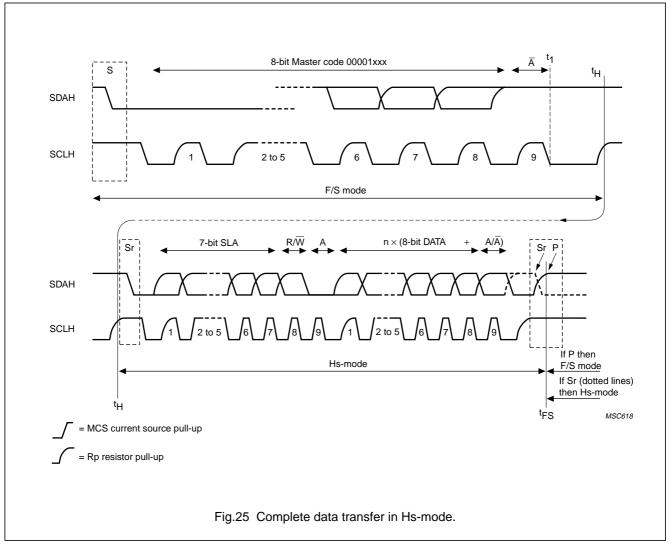
transmission the I<sup>2</sup>C-bus master issues a STOP condition (P) and switches back to F/S-mode, however, to reduce the overhead of the master code, it is possible that a master links a number of Hs-mode transfers, separated by repeated START conditions (Sr).

A read sequence (see Fig.27) follows after the Hs-mode is selected. The OM6208 will immediately start to output the requested data until a not acknowledge is transmitted by the master. The write access should be terminated by a repeated START condition so that the Hs-mode is not disabled.

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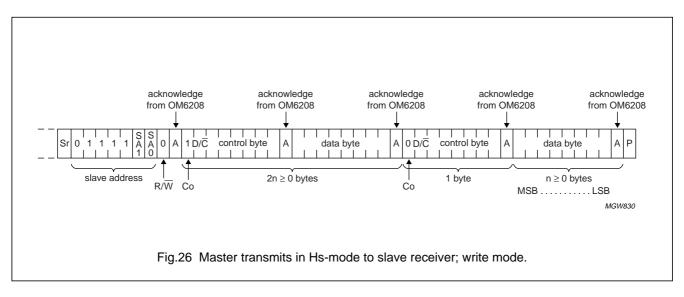
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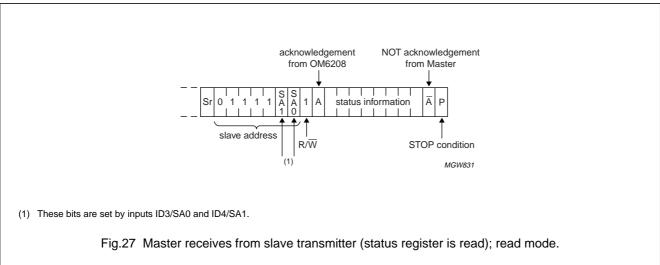




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### 10.3 Command decoder

The command decoder identifies command words that arrive on the I<sup>2</sup>C-bus:

- · Pairs of bytes
  - first byte determines whether information is display or instruction data
  - 2nd byte contains information.
- Stream of information bytes after Co = 0; display or instruction data depending on last D/C.

The most-significant bit of a control byte is the continuation bit Co. If this bit is logic 1, it indicates that only one byte, either command or RAM-data, will follow. If this bit is logic 0, it indicates that a series of bytes, either command or RAM-data, may follow. The DB6 bit of a control byte is the RAM-data/command bit  $\overline{D/C}$ . When this bit is logic 1, it indicates that a RAM-data byte will be transferred next. If the bit is logic 0, it indicates that a command byte will be transferred next.

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### 10.4 Read mode

 $I^2C$ -bus read mode operates differently from the other interfaces. Two different status bytes can be read back and are selected by first sending a 'read' instruction. A repeated START or STOP and START must then be generated followed by the slave address with the R/W bit set to read in order to read the status register.

Sending the instruction to read ID1, ID2 and VM will select the status byte shown in Table 5.

Sending the instruction to read back the temperature sensor will select the status byte shown in Table 6.

Table 5 Read status byte ID1, ID2 and VM

D7 (MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)
x <sup>(1)</sup>	VM	ID2 <sup>(2)</sup>	ID1 <sup>(2)</sup>				

### **Notes**

- 1. x = undefined.
- 2. Bits ID3 and ID4 are not available for I<sup>2</sup>C-bus because they are used to make up the two LSBs of the slave address.

Table 6 Read temperature sensor

D7 (MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)
x <sup>(1)</sup>	TD[6]	TD[5]	TD[4]	TD[3]	TD[2]	TD[1]	TD[0]

### Note

1. x = undefined.

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### 11 INSTRUCTIONS

The OM6208 may be interfaced via 3-line or 4-line Serial Peripheral Interface (SPI), 3-line serial interface or I<sup>2</sup>C-bus interface. In all cases, processing of instructions is asynchronous and does not require the internal/external oscillator to be running.

Data transmission to OM6208 may be of two types, those that define the operating mode of the device (commands) and those that fill display RAM (data). Table 7 lists all commands that are recognised by OM6208.

The Most Significant Bit (MSB) is sent first. The mode in which the  $\overline{D/C}$  bit is defined varies with the type of serial interface that is used.

D/C bit definitions:

- With 4-line SPI interface selected, the D/C bit is implemented as hard-wired input at pad D/C
- With 3-line SPI interface selected, the D/C bit is not implemented and all transmission are commands by default unless preceded by the Display data length command
- With 3-line serial and I<sup>2</sup>C-bus interface selected, the D/C
  bit is implemented through the interface protocol.

Commands can consist of one byte (single-byte) and two bytes (double-byte). Unless otherwise specified, commands may be executed in any order.

 Table 7
 Instruction set

 Instructions not expressly defined in this table and reserved instructions must not be used.

				C	OMMAI	ND BY	ΓΕ			
COMMAND NAME	D/C	(MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)	FUNCTION DESCRIPTION
Write data	1	D <sub>7</sub>	D <sub>6</sub>	D <sub>5</sub>	D <sub>4</sub>	D <sub>3</sub>	D <sub>2</sub>	D <sub>1</sub>	D <sub>0</sub>	RAM data
Horizontal addressing	0	0	0	0	0	X <sub>3</sub>	X <sub>2</sub>	X <sub>1</sub>	X <sub>0</sub>	set X address; lower 4 bits
Horizontal addressing	0	0	0	0	1	δ(1)	X <sub>6</sub>	X <sub>5</sub>	X <sub>4</sub>	set X address; upper 3 bits
Power control	0	0	0	1	0	1	PC	δ <sup>(1)</sup>	δ <sup>(1)</sup>	charge pump on/off
Charge pump control	0	0	0	1	1	1	1	0	1	set multiplication factor
	0	δ <sup>(1)</sup>	δ(1)	δ(1)	δ(1)	δ(1)	δ(1)	S <sub>1</sub>	S <sub>0</sub>	
Set V <sub>PR</sub>	0	0	0	1	0	0	V <sub>pr7</sub>	V <sub>pr6</sub>	V <sub>pr5</sub>	write V <sub>pr</sub> register
Set V <sub>PR</sub>	0	1	0	0	V <sub>pr4</sub>	V <sub>pr3</sub>	V <sub>pr2</sub>	V <sub>pr1</sub>	V <sub>pr0</sub>	write V <sub>pr</sub> register
Set bias	0	0	0	1	1	0	BS <sub>2</sub>	BS <sub>1</sub>	BS <sub>0</sub>	set bias
Display mode	0	1	0	1	0	0	1	0	DAL	all on/normal display
Display mode	0	1	0	1	0	0	1	1	Е	normal/inverse display
Display mode	0	1	0	1	0	1	1	1	DON	display ON/OFF
Data order	0	1	0	1	0	1	0	0	DOR	swap RAM MSB/LSB order
RAM addressing	0	1	0	1	0	1	0	1	V	vertical or horizontal mode
Vertical addressing	0	1	0	1	1	Y <sub>3</sub>	Y <sub>2</sub>	Y <sub>1</sub>	Y <sub>0</sub>	set Yaddress
Vertical addressing	0	0	0	1	1	1	1	1	Y <sub>4</sub>	set Yaddress
Vertical mirroring	0	1	1	0	0	MY	δ(1)	δ(1)	δ <sup>(1)</sup>	mirror Y
ID read	0	1	1	0	1	1	0	1	0	identification: ID1 <sup>(2)(3)</sup>
ID read	0	1	1	0	1	1	0	1	1	identification: ID2 <sup>(2)(3)</sup>
ID read	0	1	1	0	1	1	1	0	0	identification: ID3 <sup>(2)</sup>
ID read	0	1	1	0	1	1	1	0	1	identification: ID4 <sup>(2)</sup>
Temperature sense	0	1	1	0	1	1	1	1	0	temperature read back
VM read	0	1	1	0	1	1	1	1	1	voltage monitor(3)(4)
Row control	0	1	1	1	0	0	0	0	BRS	swap the bottom rows

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			COMMAND BYTE							
COMMAND NAME	D/C	(MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)	FUNCTION DESCRIPTION
Software reset	0	1	1	1	0	0	0	1	0	internal reset
NOP	0	1	1	1	0	0	0	1	1	no operation
Display data length	0	1	1	1	0	1	0	0	0	display data length for 3-line
	0	D <sub>7</sub>	D <sub>6</sub>	D <sub>5</sub>	D <sub>4</sub>	D <sub>3</sub>	D <sub>2</sub>	D <sub>1</sub>	D <sub>0</sub>	SPI
Temperature	0	0	0	1	1	1	0	0	0	set TC slopes A and B (SLA
compensation	0	δ(1)	SLB <sub>2</sub>	SLB <sub>1</sub>	SLB <sub>0</sub>	δ(1)	SLA <sub>2</sub>	SLA <sub>1</sub>	SLA <sub>0</sub>	and SLB)
Temperature	0	0	0	1	1	1	0	0	1	set TC slopes C and D
compensation	0	δ <sup>(1)</sup>	SLD <sub>2</sub>	SLD <sub>1</sub>	SLD <sub>0</sub>	δ(1)	SLC <sub>2</sub>	SLC <sub>1</sub>	SLC <sub>0</sub>	(SLC and SLD)
Frame frequency	0	0	0	1	1	1	1	0	0	frame frequency range and
range, oscillator tune and mode	0	MOD	T <sub>2</sub>	T <sub>1</sub>	T <sub>0</sub>	δ <sup>(1)</sup>	FR <sub>2</sub>	FR <sub>1</sub>	FR <sub>0</sub>	oscillator tune and working mode
Temperature compensation enable	0	1	1	1	0	1	0	1	TCE	enable/disable temperature compensation
Oscillator selection	0	0	0	1	1	1	0	1	EC	external oscillator
OTP programming	0	1	1	1	1	0	0	OSE	CAL MM	enter calibration mode and control programming
LOAD 0	0	1	1	0	1	1	0	0	0	write 0 to shift register
LOAD 1	0	1	1	0	1	1	0	0	1	write 1 to shift register
Select factory defaults	0	1	1	1	0	1	1	0	SFD	enable/disable defaults
N-line inversion and	0	1	0	1	0	1	1	0	1	N-line inversion and
super-frame inversion	0	FI	NL <sub>6</sub>	NL <sub>5</sub>	NL <sub>4</sub>	NL <sub>3</sub>	NL <sub>2</sub>	NL <sub>1</sub>	NL <sub>0</sub>	super-frame Inversion
	0	0	1	δ <sup>(1)</sup>	reserved					
	0	1	0	1	0	0	0	δ(1)	δ(1)	reserved
	0	1	0	1	0	1	1	0	0	reserved
	0	1	1	0	1	0	δ <sup>(1)</sup>	δ <sup>(1)</sup>	δ <sup>(1)</sup>	reserved
	0	1	1	1	0	1	0	0	1	reserved
	0	1	1	1	0	0	1	δ <sup>(1)</sup>	δ <sup>(1)</sup>	reserved
	0	1	1	1	0	1	1	1	δ <sup>(1)</sup>	reserved
	0	1	1	1	1	0	1	δ <sup>(1)</sup>	δ(1)	reserved for testing
	0	1	1	1	1	1	δ <sup>(1)</sup>	δ <sup>(1)</sup>	δ <sup>(1)</sup>	reserved for testing

### **Notes**

- 1.  $\delta$  = don't care.
- 2. ID1, ID2, ID3, ID4 and VM are read back via interface as described in Section 9.2.3. Reading back with I<sup>2</sup>C-bus interface is possible for temperature, ID1, ID2 and VM, as described in Section 10.4.
- 3. ID1 will always return to logic 0; ID2 will always return to logic 1. The VM bit is set to logic 1 when the charge pump is running and logic 0 when the charge pump is not running.
- 4. If the Factory Defaults bit (MMFD) has been programmed to 1, then the SFD instruction is ignored and the device will always use the OTP default data.

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### 11.1 Description of command bits

Table 8 Bit descriptions

BIT	0	1	RESET STATE				
DON	display off	display on	0				
E	normal display	splay inverse video mode					
DAL	normal display	all pixel on	1				
MY	no Y mirroring	Y mirroring	0				
PC	charge pump off	charge pump on	0				
DOR	normal data order	MSB/LSB transposed for RAM data	0				
V	horizontal addressing	vertical addressing	0				
BRS	bottom rows are not mirrored	bottom rows are mirrored	0				
EC	internal oscillator is selected	external clock to be used	0				
CALMM	exit OTP calibration mode	enter OTP calibration mode <sup>(1)</sup>	0				
TCE	disable temperature compensation	enable temperature compensation	1				
OSE	disable OTP programmed voltage	le OTP programmed voltage enable OTP programmed voltage <sup>(1)</sup>					
SFD	use interface programmed data	use OTP programmed data <sup>(2)</sup>	0				
MOD	grey-scale mode is selected	black-and-white mode is selected	0				
SLA[2:0]	select slope	for segment A	000 <sup>(2)</sup>				
SLB[2:0]	select slope	for segment B	000(2)				
SLC[2:0]	select slope	for segment C	000 <sup>(2)</sup>				
SLD[2:0]	select slope	for segment D	000(2)				
X[6:0]	sets X address (colum	n) for writing in the RAM	0000000				
Y[4:0]	sets Yaddress (bank	) for writing in the RAM	00000				
S[1:0]	charge pump multiplica	ation factor (see Table 10)	0000(2)				
NL[6:0]	sets N-line inver	sion (see Table 18)	0001101 <sup>(2)</sup>				
FR[2:0]	sets frame frequenc	sets frame frequency range (see Table 11)					
T[2:0]	oscillator tune; sets frame frequ	oscillator tune; sets frame frequency within a range (see Table 11)					
D[7:0]	display data length	for 3-line SPI interface	00000000				
VPR[7:0]	V <sub>PR</sub>	V <sub>PR</sub> register					
BS[2:0]	bias setting lev	vel (see Table 13)	000(2)				
FI	super-fran	me inversion	0 <sup>(2)</sup>				

### **Notes**

- Calibration mode may not be entered if the SEAL bit has been set. Programming is only possible when in calibration mode.
- 2. These values can be set by the module maker. If the factory defaults OTP bit (MMFD) has been set then these values cannot be changed via the interface. Otherwise, the OTP data will only be used if bit SFD is set to 1.
- 3. FR[2:0] = 001 and T[2:0] = 110 gives 150 Hz as default frame frequency.

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Table 9 Display and power mode bits DON, DAL and E

DON	DAL <sup>(1)</sup>	<b>E</b> <sup>(2)</sup>	DESCRIPTION
0	0	X <sup>(3)</sup>	display off; all row and column outputs at V <sub>SS</sub> ; oscillator on; HV generator enabled
0	1	X <sup>(3)</sup>	Power-down mode; display off; all row and column outputs at V <sub>SS</sub> ; oscillator off; HV generator disabled
1	0	0	normal display mode
1	0	1	inverse display mode
1	1	X <sup>(3)</sup>	all pixels on

### **Notes**

- 1. The DAL bit has priority over the  $\overline{E}$  bit.
- 2. Refer also to Table 17.
- X = don't care.

Table 10 Multiplication settings for charge pump

S1	S0	VOLTAGE MULTIPLIER
0	0	4 ×
0	1	5×
1	0	6×
1	1	7×

### 11.2 Frame frequency setting and oscillator tuning

Grey-scale mode and black-and-white mode require different frame frequencies. The appropriate frame frequency ( $f_{frame}$ ) is derived from the oscillator frequency ( $f_{osc}$ ) using a presettable divider as shown in the equation

$$f_{frame} = \frac{f_{OSC}}{division ratio}$$

There are eight possible divider settings and these are selected by the parameter FR[2:0], see Table 11.

**Table 11** Frame frequencies for  $f_{osc} = 400 \text{ kHz}$ 

FR2	FR1	FR0	DIVISION RATIO	f <sub>frame</sub> (Hz)
0	0	0	2448	163.4
0	0	1	3265	122.5
0	1	0	4082	98.0
0	1	1	4896	81.7
1	0	0	5714	70.0
1	0	1	7340	54.5
1	1	0	8968	44.6
1	1	1	11428	35.0

Oscillator tuning is controlled by the parameter T[2:0]. As a result of oscillator tuning,  $f_{OSC}$  is increased by approximately 4% per step according to the equation

$$f_{OSC} = 400 \text{ kHz } \times (1 + 0.04 \times T)$$
 (1)

where T is the decimal value of T[2:0].

**Example.** For the default values given in Table 8 (i.e. FR[2:0] = 001 and T[2:0] = 110) the selected frame frequency is  $122.5 \text{ Hz} \times (1 + 6 \times 0.04) = 151.9 \text{ Hz}$ .

Equation (1) shows the typical value of the oscillator frequency. The accuracy of this parameter is defined in Chapter 15. The frame frequency accuracy results directly from the oscillator accuracy.

### 11.3 Initialization

Immediately following power-on, all internal registers and the RAM content are undefined. A reset pulse must be applied to the  $\overline{\text{RES}}$  pad.

Reset is accomplished by applying an external reset pulse (active LOW) to the  $\overline{RES}$  input. When reset occurs within the specified time, all internal registers are reset, however the RAM remains undefined. The state after reset is described in Section 11.4.

At power-on, the  $\overline{RES}$  input must be  $\leq 0.3 V_{DD1}$  when  $V_{DD1}$  reaches  $V_{DD(min)}$  (or higher) within the maximum time  $t_{VHRL}$  after  $V_{DD1}$  going HIGH (see Fig.37). Alternatively a reset pulse can be applied when  $V_{DD1}$  is stable.

A reset can also be made by sending a reset command. This command can be used during normal operation but not to initialize the chip after power-on.

After power-off, the  $\overline{\text{RES}}$  input must not be HIGH when  $V_{DD1}$  is not HIGH.

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### 11.4 Reset function

After reset, the LCD driver is in Power-down mode, the RAM is undefined and the internal registers have the status shown in Table 8.

### 11.5 Power-down mode

In the Power-down mode:

- All LCD outputs (row and column outputs) are at V<sub>SS</sub> (display off)
- Bias generator and V<sub>LCD</sub> generator are switched off; external V<sub>LCD</sub> supply can be applied or disconnected
- · Oscillator is off (an external clock is possible)
- RAM contents are unchanged; RAM data can be written
- V<sub>LCD</sub> is discharged to V<sub>SS</sub>.

Power-down mode is active when the display is off (DON = 0) and all the pixels are on (DAL = 1).

### 11.6 Display Control

The bits DON, E and DAL select the display mode (see Table 9).

### 11.6.1 HORIZONTAL MIRRORING

When the MX input is at logic 0, the display RAM is written from left to right (X = 0) is on the left side).

When the MX input is set to 1, the display RAM is written from right to left (X = 0) is on the right side).

The MX input value has an impact on the way the RAM is written: if a horizontal mirroring of the display is desired, the RAM must be rewritten after changing the MX pad value.

### 11.6.2 VERTICAL MIRRORING

When the MY bit is set to logic 1, the display is mirrored vertically.

A change of this bit has an immediate effect on the display, it is not necessary to rewrite RAM for the effect to take place.

### 11.7 Set Y address of RAM

Y[4:0] defines the Y address of the display RAM.

Table 12 Yaddress range

Y4	Y3	Y2	Y1	Y0	DISPLAY RAM
0	0	0	0	0	bank 0
0	0	0	0	1	bank 1
0	0	0	1	0	bank 2
0	0	0	1	1	bank 3
0	0	1	0	0	bank 4
0	0	1	0	1	bank 5
0	0	1	1	0	bank 6
0	0	1	1	1	bank 7
0	1	0	0	0	bank 8
0	1	0	0	1	bank 9
0	1	0	1	0	bank 10
0	1	0	1	1	bank 11
0	1	1	0	0	bank 12
0	1	1	0	1	bank 13
0	1	1	1	0	bank 14
0	1	1	1	1	bank 15
1	0	0	0	1	bank 16

### 11.8 Set X address of RAM

The X address points to the columns. The range of X is 0 to 95.

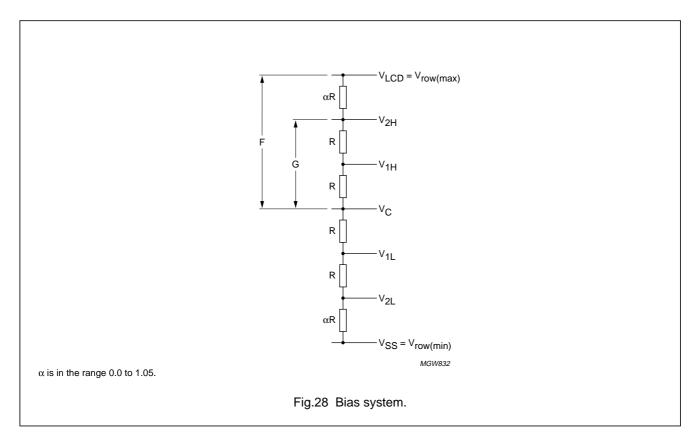
### 11.9 Bias levels

The OM6208 is a grey-scale driver able to provide different bias voltage levels for rows and columns. The row voltage values are  $V_{LCD}$ ,  $V_{SS}$  and  $V_{C}$ , generated using the resistor chain shown in Fig.28.

The five levels used to drive the columns are shown in Fig.28. These are  $V_{2L}$ ,  $V_{1L}$ ,  $V_{C}$ ,  $V_{1H}$  and  $V_{2H}$ , all of which depend on the value of alpha. Table 13 shows all possible combinations of alpha settable by programming the BS[2:0] bits.

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**Table 13** Bias setting levels for p = 4

BS2	BS1	BS0	F/G	α	а
0	0	0	1.000	0.00	4.00
0	0	1	1.075	0.15	4.30
0	1	0	1.150	0.30	4.60
0	1	1	1.225	0.45	4.90
1	0	0	1.300	0.60	5.20
1	0	1	1.375	0.75	5.50
1	1	0	1.450	0.90	5.80
1	1	1	1.525	1.05	6.10

Each of the eight possible values of alpha results in a different set of five values for the column voltages.

Bias level F (see Fig.28) is half of the maximum row voltage level as shown by the equation

$$F = \frac{V_{LCD}}{2}$$

Figure 28 also shows that G is used to define the maximum column voltage level related to the V<sub>C</sub> level.

Because the voltage level of the row depends on the programmed bias level, it can be seen that  $F \ge G$ 

The situation where F = G occurs only when BS[2:0] is zero and alpha is zero. In this case  $\alpha R = 0$  and G = F; therefore  $V_{2H} = V_{LCD}$  and  $V_{2L} = V_{SS}$ , also two of the internal buffers are no longer needed and therefore are switched off to reduce power consumption.

The relationship between F and G is defined by the parameter a (indicated in Table 13) and p as follows

$$\frac{F}{G} = \frac{a}{p}$$

It can be seen from Fig.28 that

$$\frac{F}{G} = \frac{a}{p} = \frac{(\alpha+2)}{2} = 1 + \frac{\alpha}{2}$$

or

$$a \ = \ \left(\frac{\alpha}{2} + 1\right) \cdot p \qquad \quad \alpha \ = \ \left(\frac{a}{p} - 1\right) \cdot 2$$

The BS[2:0] bias bits can be selected by a command and also can be programmed by OTP.

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The relationship between the parameters F, p, a and N (number of rows of the display) and the  $V_{on(rms)}$  and  $V_{off(rms)}$  voltage values according to the typical LCD properties of the pixel are shown in equations (2) and (3).

$$V_{on(rms)} = \frac{F}{a} \cdot \sqrt{\frac{p \cdot (a^2 + N + 2a)}{N}}$$
 (2)

$$V_{off(rms)} = \frac{F}{a} \cdot \sqrt{\frac{p \cdot (a^2 + N - 2a)}{N}}$$
 (3)

### 11.10 LCD drive voltage

### 11.10.1 LCD DRIVE VOLTAGE GENERATION

V<sub>LCD</sub> may be supplied externally or generated internally by the on-chip capacitive charge pump. OM6208 features on-chip capacitors resulting in a minimum of external components required for operation (see Chapter 16).

The 'power control' instruction may be used to switch  $V_{LCD}$  generation on or off. The charge pump control instruction may be used to select the required voltage multiplication factor. The 'set  $V_{PR}$ ' instruction is used for programming the LCD drive voltage  $V_{LCD}$ .

The generation of  $V_{LCD}$  in OM6208 is illustrated in Fig.29. This shows all factors that effect  $V_{LCD}$  generation, including the 6 bits of MMVOPCAL (from OTP) and the 7 bits resulting from the temperature compensation mechanism. Equations summarizing all factors are

$$V_{OP} = V_{PR} + MMVOPCAL + V_{T}$$
 (4)

and

$$V_{ICD} = V_{OP} \cdot b + a \tag{5}$$

Where:

 $V_{PR}[7:0]$  is set in the instruction decoder and is the programmed  $V_{PR}$  register value as an unsigned number MMVOPCAL[5:0] is the value of the offset stored in the OTP cells in twos complement format

 $V_T[7:0]$  in twos complement format comes from the temperature compensation block (see Table 16) a and b are fixed constant values (see Table 14).

Table 14 Parameters of V<sub>LCD</sub>

SYMBOL	VALUE	UNIT
р	0.03	V
а	3	V

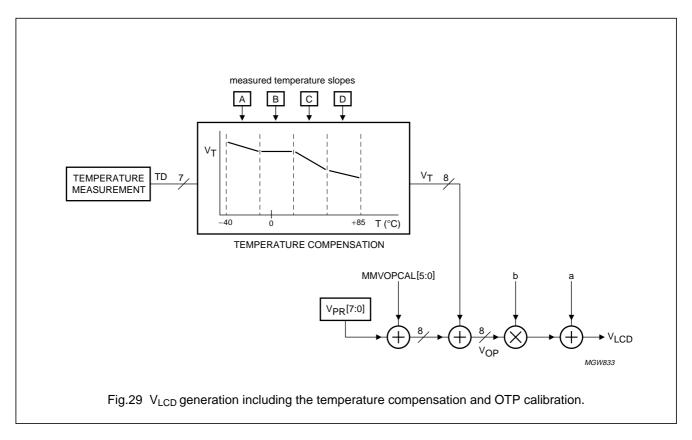
### **CAUTION**

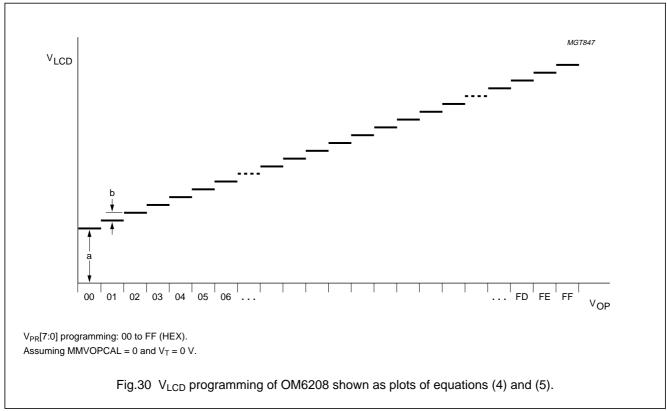
As the programming range for the internally generated  $V_{LCD}$  allows values above the maximum allowed  $V_{LCD}$  (9 V), the user has to ensure, while setting the  $V_{PR}$  register and selecting the temperature compensation, that under all conditions and including all tolerances  $V_{LCD}$  remains below 9.0 V.

Also, because the programming range for the internally generated  $V_{LCD}$  allows values below the minimum allowed  $V_{LCD}$  (5 V), the user has to ensure, while setting the  $V_{PR}$  register and selecting the temperature compensation, that under all conditions and including all tolerances  $V_{LCD}$  remains above 5.0 V.

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### 11.10.2 TEMPERATURE MEASUREMENT

The temperature measurement is repeated every 10 seconds. The measured value is provided as a 7-bit digital value TD[6:0] which can be read back via the interface. The temperature can be determined from TD[6:0] using the equation

$$T = (1.875 \times TD - 40)^{\circ}C$$
 (6)

### 11.10.3 TEMPERATURE COMPENSATION

Due to the temperature dependency of the liquid crystal's viscosity, the LCD controlling voltage V<sub>LCD</sub> may have to be adjusted at different temperatures to maintain optimal contrast.

Internal temperature compensation may be enabled via the 'temperature compensation enable' instruction. When the internal temperature compensation is applied (TCE bit is set to 1) then according to Equation (4) the  $V_{LCD}$  depends also on  $V_T$  (the temperature compensation component defined in Table 16), otherwise  $V_T$  is considered to be 0 V.

After the reset, the  $V_{LCD}$  is fixed because the  $V_{PR}$  is a register that is reset to zero. The MMVOPCAL is also set to zero because this comes from the registers of OTP that are not refreshed yet, also  $V_T$  is evaluated after the reset because the temperature measurement block supplies a TD value that is the default value stored in the register after the reset.

The four temperature coefficients MA, MB, MC and MD correspond to four equally spaced temperature regions. Each coefficient can be selected from a choice of eight different slopes, or multiplication factors. Each one of these coefficients may be independently selected by the user via the 'temperature compensation enable' instruction. The default for each slope register can be stored in OTP.

**Table 15** Temperature coefficients Slopes of  $V_{LCD}$  are calculated from equations (4), (5), (6) and Table 16.

SLA, SLB, SLC and SLD	MA, MB, MC and MD	SLOPE (mV/K)
111	3.00	-48
110	2.00	-32
101	1.25	-20
100	1.00	-16
011	0.75	-12
010	0.50	-8
001	0.25	-4
000	0.00	0

Temperature compensation is implemented by adding an offset  $V_T$  to the  $V_{PR}$  value (additionally to the OTP calibration offset MMVOPCAL).

The final result for  $V_{LCD}$  calculation is an 8-bit positive number as shown in equations (4) and (5). Care must be taken by the user to ensure that the ranges of  $V_{PR}$ , MMVOPCAL and  $V_{T}$  do not cause clipping and hence undesired results. The adder stages will not permit overflow or underflow and will clamp results to either end of the range.

The temperature read-out generates a 7-bit result, TD[6:0]. For temperatures below  $-40~^{\circ}$ C, the value of TD is zero. For temperatures above 79  $^{\circ}$ C, the value of TD is higher than 63, but for V<sub>T</sub> calibration the value TD = 63 is used.

The offset value  $V_T$  may be calculated from Table 16. The effect on  $V_{LCD}$  can be calculated by multiplying the offset value with the value of b (from Table 14).

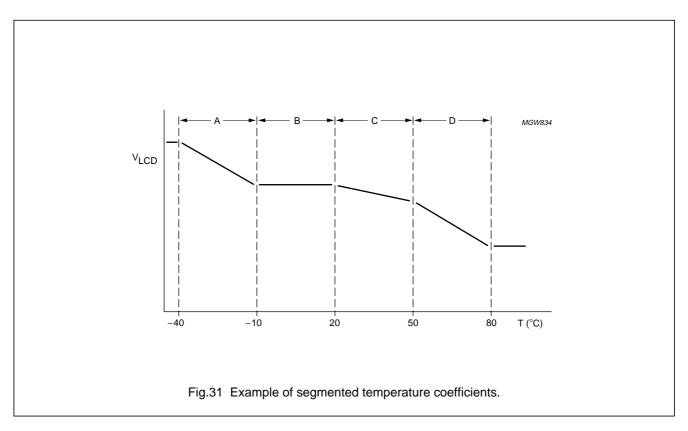
For example, if T = -10 °C, TD = 16 and MB = 1.25 then  $V_{LCDoffset}$  = 30 mV × (32 - 16) × 1.25 = 600 mV.

Table 16 Temperature compensation equations

TEMPERATURE RANGE (°C)	TD RANGE	EQUATION
−40 to −11	0 to 15	$V_T = (16 \times MB) + MA \times (16 - TD)$
−10 to +19	16 to 31	$V_T = (32 - TD) \times MB$
+20 to +49	32 to 47	$V_T = -(TD - 32) \times MC$
+50 to +79	48 to 63	$V_T = -(16 \times MC) + MD \times (TD - 48)$

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### 11.11 Grey-scale mode and black-and-white mode

It is possible to set via command the working mode of the OM6208. This is by setting the MOD bit of the 'frame frequency' instruction, oscillator tune and mode. By default, the MOD bit is set to logic 0 and grey-scale mode is selected. In that mode, grey-scales are generated using Frame Rate Control (FRC). Three frames together form a super-frame. The frame frequency is adjustable but all three frames have the same duration. A grey-scale is generated by selecting either 0, 1, 2 or all 3 frames (see Table 17).

If the MOD bit is set to logic 1 black-and-white mode is selected, meaning that only black-and-white levels are generated and only one frame type is sent to the display. Thus only the MSBs stored in the RAM are used for all three frames. The LSBs are ignored. Thus the way the data is stored in the RAM is the same as for grey-scale. As all frames are identical the frame frequency may be reduced (see Table 11).

Table 17 Grey-scale levels with FRC

GS[	[1:0]	SUPER-FRAME(1)	GREY-SCALE LEVELS						
Normal	mode (								
0	0	000	white						
0	1	001	light grey						
1	0	110	dark grey						
1	1	111	black						
Inverse	mode (	E = 1)							
0	0	111	black						
0	1	110	dark grey						
1	0	001	light grey						
1	1	000	white						

### Note

 The first and second frames in each super-frame are related to the MSB of GS[1:0] (GS = 11); the third frame is related to the LSB (GS = 00).

### 11.12 N-line inversion and frame inversion

N-line inversion can be set from 0 to 127 as shown in Table 18.

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Table 18 N-line inversion

INVERSION AFTER	NL <sub>6</sub>	NL <sub>5</sub>	NL <sub>4</sub>	NL <sub>3</sub>	NL <sub>2</sub>	NL <sub>1</sub>	NL <sub>0</sub>	FI
0-line inversion	0	0	0	0	0	0	0	0
1-line inversion	0	0	0	0	0	0	1	0
2-line inversion	0	0	0	0	0	1	0	0
3-line inversion	0	0	0	0	0	1	1	0
4-line inversion	0	0	0	0	1	0	0	0
5-line inversion	0	0	0	0	1	0	1	0
6-line inversion	0	0	0	0	1	1	0	0
7-line inversion	0	0	0	0	1	1	1	0
8-line inversion	0	0	0	1	0	0	0	0
:	:	:	:	:	:	:	:	:
66-line inversion	1	0	0	0	0	1	0	0
67-line inversion	1	0	0	0	0	1	1	0
68-line inversion	1	0	0	0	1	0	0	0
:	:	:	:	:	:	:	:	:
127-line inversion	0	1	0	0	0	0	0	0
0 only super-frame inversion	0	0	0	0	0	0	0	1
1-line inversion and super-frame inversion	0	0	0	0	0	0	1	1
2-line inversion and super-frame inversion	0	0	0	0	0	1	0	1
3-line inversion and super-frame inversion	0	0	0	0	0	1	1	1
4-line inversion and super-frame inversion	0	0	0	0	1	0	0	1
5-line inversion and super-frame inversion	0	0	0	0	1	0	1	1
6-line inversion and super-frame inversion	0	0	0	0	1	1	0	1
7-line inversion and super-frame inversion	0	0	0	0	1	1	1	1
8-line inversion and super-frame inversion	0	0	0	1	0	0	0	1
:	:	:	:	:	:	:	:	:
66-line inversion and super-frame inversion	1	0	0	0	0	1	0	1
67-line inversion and super-frame inversion	1	0	0	0	0	1	1	1
68-line inversion and super-frame inversion	1	0	0	0	1	0	0	1
:	:	:	:	:	:	:	:	:
127-line inversion and super-frame inversion	1	1	1	1	1	1	1	1

### **Notes**

- In grey-scale mode the super-frame inversion is performed if bit FI in the 'N-line inversion and super-frame inversion'
  instruction is set to logic 1. In black-and-white mode, the super-frame inversion continues in groups of three frames.
- 2. NL[6:0] may be set in the range 0 to 127. If NL = 0, then no line inversion is performed; if NL = MUX rate = 68 then N-line inversion is equal to frame inversion.
- 3. With N-line inversion the output signal polarity changes every N row pulse periods (with p = 4 this means inversion occurs after every  $4 \times NL$  rows of the display).
- 4. If after a super-frame FI = 1 and there is an inversion due to an N-line inversion, this inversion occurs only once.

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The example in Table 19 shows the first super-frame with the settings NL = 3, MUX = 20 and p = 4 applied; the super-frame contains three frames. The next super-frame will be a repeat of the first super-frame if bit FI is set to logic 0 (no super-frame inversion), or will start with the first frame having the opposite sign if bit FI is at logic 1 (super-frame inversion) and the N-line inversion counter will also restart.

Super-frame inversion requires that the state of the previous super-frame is remembered, i.e., if the previous super-frame started '+', then the next super-frame must start '-'. This has priority over inversions triggered by the counter, so that if the counter triggers an inversion at a super-frame boundary and super-frame inversion is active, then the two do not cancel each other out but the super-frame inversion has priority.

**Table 19** Example showing line inversions in one super-frame: NL = 3, MUX = 20 and p = 4

																						S	UF	PEF	₹-F	RA	ME	1																				
						FI	RA	ME	<u> </u>									FRAME 2														FRAME 3																
		UB ME	0	FI	SU		1	F	SI RAI			SUB FRAME 3			3	SUB FRAME				F	SUB FRAME 1			1	SUB FRAME 2			2	SUB FRAME 3				SUB FRAME 0				SUB FRAME 1				SUB FRAME 2					SUI	3 IE 3	
+				-				-				-					+				+	-				+				-	_		11-									+			1	+		
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T	+				+				-				-				-	-				+				4	۲				+				-				-			-	-			+	•	
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T		1-	1			+				4	F				+				1-	1				_				1-				1	+			+		H		+				-				_
T		1-	+			+				4	+				+				1-	+				_				-				-	+			+		H		+				-				_
	T	T-	+			+				1	+				+				1-	+				_				<u> </u>				-	+			+				+								_
1	t	†-	+	H		+	П			1	+		П		+	$\top$			†-	1	T	T						-		1	$\top$	-	+			+				+	1			-	1			
1		$\dagger$	†-			T				1	+		$\Box$			+				4	+				7		$\dagger$		-		$\top$	$\dagger$	1-				+				+				+			
T			†-	H		T				$\dagger$	+		H			+	$\top$			4	+	T			$\dashv$		T		-	$\dashv$	$\top$		1-			$\dagger$	+	$\Box$		$\forall$	+				+			-
T	1		†-	H						t	+		H			+		T		+	F	t			_		T	T	1-1	$\dashv$	T		1_		+	t	+	$\vdash$		H	+			$\Box$	+			-
$\top$			†_		$\top$						+		H			+		$\dagger$		4	+				$\exists$	+				$\dashv$	$\top$	$\dagger$	1_				+				+				+			-

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### 12 LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 60134); notes 1 and 2.

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
V <sub>DD1</sub>	supply voltage (logic circuits)	-0.5	+6.5	V
V <sub>DD2</sub> , V <sub>DD3</sub>	supply voltage (analog circuits)	-0.5	+5.0	V
V <sub>LCD</sub>	LCD supply voltage	-0.5	+10.0	V
VI	input voltage (any pad)	-0.5	$V_{DD1} + 0.5$	V
I <sub>SS</sub>	ground supply current	-50	+50	mA
$I_I$ , $I_O$	DC input or output current	-10	+10	mA
P <sub>tot</sub>	total power dissipation	1	300	mW
P/out	power dissipation per output	_	30	mW
T <sub>stg</sub>	storage temperature	-65	+150	°C

### **Notes**

- 1. Stresses above those listed under limiting values may cause permanent damage to the device.
- 2. Parameters are valid over operating temperature range unless otherwise specified. All voltages are with respect to V<sub>SS</sub> unless otherwise noted.

### 13 HANDLING

Inputs and outputs are protected against electrostatic discharge in normal handling. However, to be totally safe, it is desirable to take normal precautions appropriate to handling MOS devices (see "Handling MOS devices").

## 14 DC CHARACTERISTICS

 $V_{DD1}$  = 1.7 to 3.3 V;  $V_{SS}$  = 0 V;  $V_{LCD}$  = 5 to 9.0 V;  $T_{amb}$  = -40 to +85 °C; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>DD1</sub>	supply voltage (logic circuits)		1.7	_	3.3	V
V <sub>DD2</sub> , V <sub>DD3</sub>	supply voltage (analog circuits)		2.4	_	4.5	V
V <sub>LCDIN</sub>	LCD supply voltage input	LCD voltage supplied externally; high voltage generator disabled	_	_	9.0	V
V <sub>LCDOUT</sub>	LCD supply voltage output	LCD voltage generated internally; high voltage generator enabled; note 1	_	_	9.0	V
V <sub>LCD(tol)</sub>	tolerance of generated V <sub>LCD</sub>	with calibration; note 2	-70	_	+70	mV
I <sub>DD</sub>	supply current; pins $V_{DD1}$ , $V_{DD2}$ and $V_{DD3}$	Power-down mode (all static currents switched off); note 3	_	3	_	μΑ
I <sub>DD1</sub>	supply current; pin V <sub>DD1</sub>	notes 3 and 4	_	20	40	μΑ
$I_{DD2}, I_{DD3}$	supply current; pins V <sub>DD2</sub>	notes 3 and 4				
	and V <sub>DD3</sub>	DC load on V <sub>LCD</sub> = 300 μA	_	2400	-	μΑ
		DC load on V <sub>LCD</sub> = 170 μA	_	1200	2000	μΑ
I <sub>DD2</sub>	supply current; pin V <sub>DD2</sub>	notes 4 and 5				
		DC load on V <sub>LCD</sub> = 32 μA	_	360	_	μΑ

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT						
Logic circ	Logic circuits											
V <sub>OL</sub>	LOW-level output voltage	_	0.2V <sub>DD1</sub>	V								
V <sub>OH</sub>	HIGH-level output voltage	$I_{OH} = -0.5 \text{ mA}$	0.8V <sub>DD1</sub>	_	V <sub>DD1</sub>	V						
V <sub>IL</sub>	LOW-level input voltage		V <sub>SS</sub>	_	0.3V <sub>DD1</sub>	V						
V <sub>IH</sub>	HIGH-level input voltage		0.7V <sub>DD1</sub>	_	V <sub>DD1</sub>	V						
IL	leakage current	$V_I = V_{DD1}$ or $V_{SS}$	-1	_	+1	μΑ						
Column a	nd row outputs					•						
R <sub>col</sub>	column output resistance C0 to C95	V <sub>LCD</sub> = 7 V	_	3	7	kΩ						
R <sub>row</sub>	row output resistance R0 to R64	V <sub>LCD</sub> = 7 V; load 10 μA; outputs tested one at a time	_	2	5	kΩ						
V <sub>col</sub>	bias tolerance C0 to C95		-70	0	+70	mV						
V <sub>row</sub>	bias tolerance R0 to R64		-70	0	+70	mV						
Temperati	ure read-back			•								
$\Delta T_{RB}$	temperature read-back tolerance		-8	_	+8	°C						

### **Notes**

- $1. \quad \text{The maximum possible $V_{LCD}$ voltage that may be generated is dependent on voltage, temperature and (display) load.}$
- 2. Valid for the values of temperature,  $V_{PR}$  and temperature compensation used at calibration.
- 3.  $V_{DD1}$  = 1.8 V;  $V_{DD2}$  = 2.7 V; inputs at  $V_{DD1}$  or  $V_{SS}$ ; interface inactive; internal  $V_{LCD}$  generation.
- 4. Grey-scale or black-and-white mode; display mode ON; all outputs open-circuit; BS[2:0] = 000.
- 5. V<sub>LCD</sub> = 6.84 V; default frequency; data pattern in RAM is with all bytes = AA(HEX); multiplication factor = 5.

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## 15 AC CHARACTERISTICS

 $V_{DD1}$  = 1.7 to 3.3 V;  $V_{SS}$  = 0 V;  $V_{LCD}$  = max. 9.0 V;  $T_{amb}$  = -40 to +85 °C; all timings are between 20% and 80% of  $V_{DD1}$ ; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
f <sub>OSC</sub>	oscillator frequency selection	see Section 11.2, equation (1)	400 <sup>(1)</sup>	496 <sup>(2)</sup>	512 <sup>(3)</sup>	kHz
f <sub>EXT</sub>	external clock frequency		400	496	512	kHz
f <sub>frame</sub>	frame frequency selection	requency selection default frequency: $FR[2:0] = 001$ ; $T[2:0] = 110$				Hz
$\Delta f_{OSC}$ ,	accuracy of oscillator frequency	V <sub>DD1</sub> = 2.8 V;	-15	_	+15	%
$\Delta f_{frame}$	and frame frequency	$T_{amb} = -20 \text{ to } +70 ^{\circ}\text{C}$				
Serial timin	g characteristics: 3-line and 4-lin	e SPI and serial interface; $V_{DD1} =$	1.8 to 3.3	V; see F	igs 32 to	35
f <sub>SCLK</sub>	clock frequency		_	_	6.5	MHz
T <sub>CYC</sub>	SCLK clock cycle time		153	_	_	ns
t <sub>PWH1</sub>	SCLK pulse width high		70	_	-	ns
t <sub>PWL1</sub>	SCLK pulse width low		60	_	_	ns
t <sub>S2</sub>	SCE set-up time		60	_	_	ns
t <sub>H2</sub>	SCE hold time		60	_	_	ns
t <sub>PWH2</sub>	SCE minimum HIGH time		50	_	_	ns
t <sub>S4</sub>	SDATA set-up time		60	_	<b> </b> -	ns
t <sub>H4</sub>	SDATA hold time		60	_	_	ns
t <sub>S3</sub>	data/command set-up time		60	_	_	ns
t <sub>H3</sub>	data/command hold time		60	_	1-	ns
t <sub>S1</sub>	SDATA set-up time		50	_	-	ns
t <sub>H1</sub>	SDATA hold time		70	_	-	ns
t <sub>1</sub>	SDO access time		_	_	50	ns
t <sub>2</sub>	SDO disable time	3-line SPI or 4-line SPI interface	_	_	50	ns
t <sub>3</sub>	SCE hold time		50	_	_	ns
t <sub>4</sub>	SDO disable time	3-line serial interface	25	_	110	ns
C <sub>b</sub>	capacitive load for SDO	note 4	_	_	50	pF
R <sub>b</sub>	series resistance for SDO	note 4	_	_	500	Ω
I <sup>2</sup> C-bus inte	erface timing characteristics; V <sub>DD</sub>	1 = 1.8 to 3.3 V; see Fig.36				
f <sub>SCLH</sub>	SCLH clock frequency		0	_	3.4	MHz
t <sub>SU;STA</sub>	set-up time (repeated) START condition		160	_	_	ns
t <sub>HD;STA</sub>	hold time (repeated) START condition		160	_	_	ns
t <sub>LOW</sub>	LOW period of the SCLH clock		160	_	_	ns
t <sub>HIGH</sub>	HIGH period of the SCLH clock		60	_	_	ns
t <sub>SU;DAT</sub>	data set-up time		10	_	-	ns
t <sub>HD;DAT</sub>	data hold time		15	_	70	ns
t <sub>rDA</sub>	rise time of SDAH signal		20	_	80	ns
t <sub>fDA</sub>	fall time of SDAH signal		20	_	80	ns

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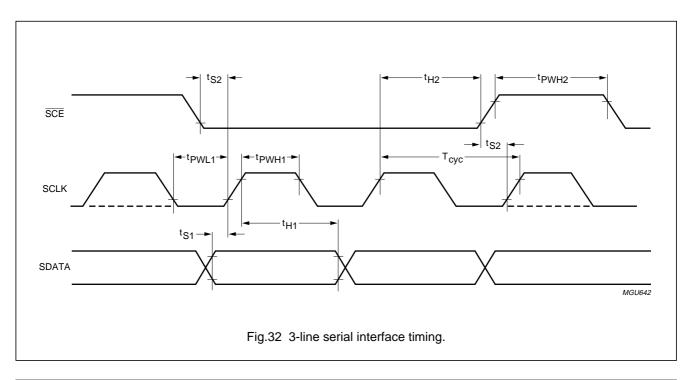
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
t <sub>SU;STO</sub>	set-up time for STOP condition		160	_	_	ns
C <sub>b</sub>	capacitive load for SDAH and SCLH lines	note 5	_	_	100	pF
	capacitive load for SDAH + SDA line and SCLH + SCL line		_	_	400	pF
t <sub>SW</sub>	tolerable spike width on bus		_	_	5	ns
V <sub>nL</sub>	noise margin at the LOW level for each connected device (including hysteresis)		0.1V <sub>DD</sub>	_	_	V
V <sub>nH</sub>	noise margin at the HIGH level for each connected device (including hysteresis)		0.2V <sub>DD</sub>	-	-	V
RESET timi	ng characteristics; see Fig.37					•
t <sub>VHRL</sub>	V <sub>DD</sub> to RES LOW		0(6)	_	1	μs
t <sub>RW</sub>	reset low pulse width		1000	_	_	ns
t <sub>RWS</sub>	reset pulse width spike suppression		_	_	100	ns

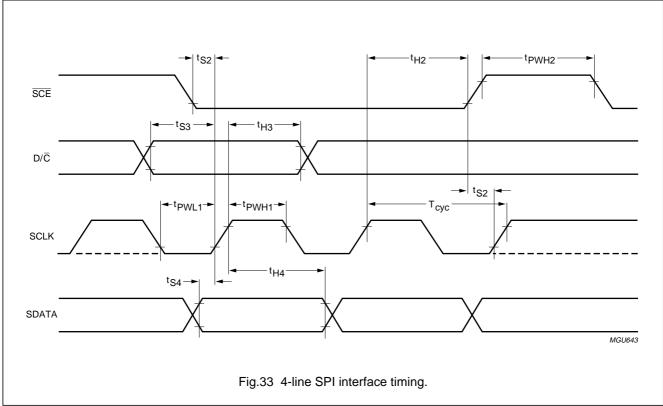
### **Notes**

- 1.  $f_{OSC}$  defined for T[2:0] = 000.
- 2.  $f_{OSC}$  defined for T[2:0] = 110 (default value).
- 3.  $f_{OSC}$  defined for T[2:0] = 111.
- 4. Maximum value is for  $f_{SCLK} = 6.5$  MHz; series resistance includes ITO track + connector resistance + PCB.
- 5.  $C_b = 100 \text{ pF}$  total capacitance of one bus line.
- 6.  $\overline{\text{RES}}$  may be LOW before  $V_{DD1}$  goes HIGH.

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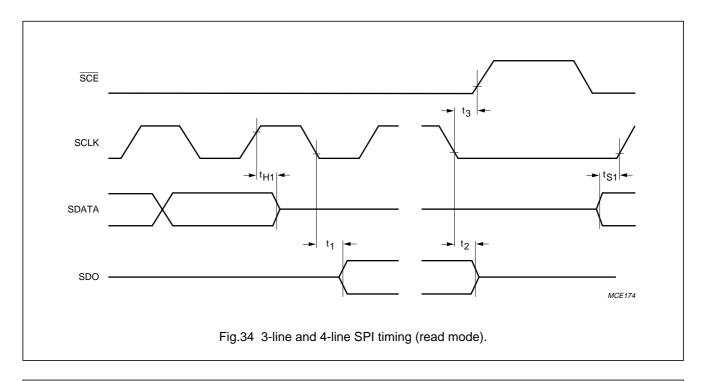
OM6208

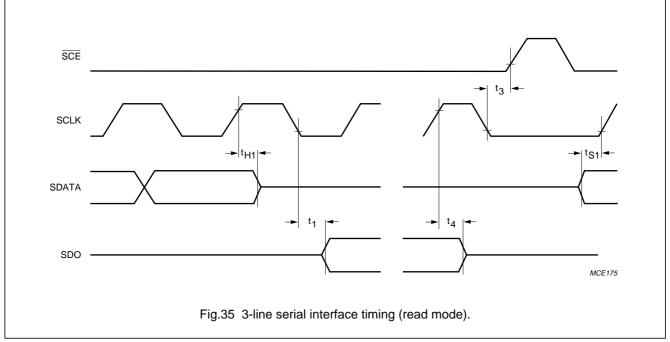




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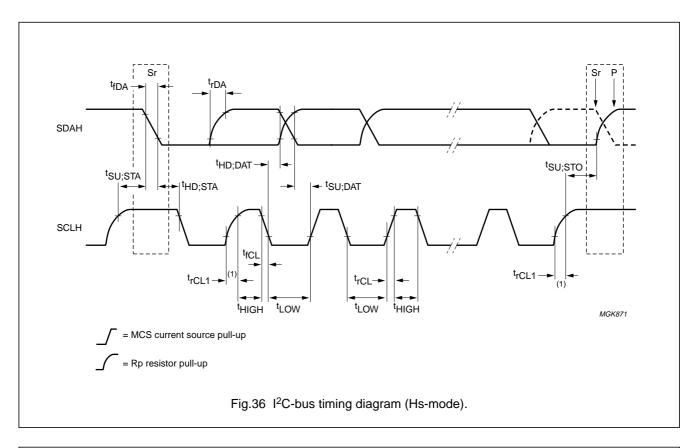
OM6208

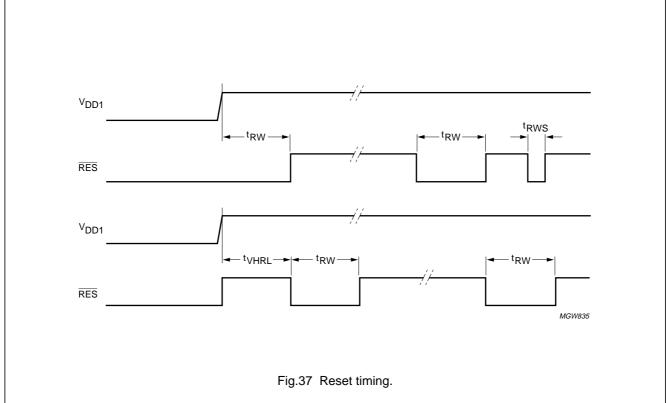




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#### 16 APPLICATION INFORMATION

### 16.1 Protection from light

Semiconductors are light sensitive. Exposure to light sources can cause malfunction of the IC. In the application it is therefore required to protect the IC from light. The protection has to be done on all sides of the IC, i.e. front, rear and all edges.

## 16.2 Chip-on-glass displays

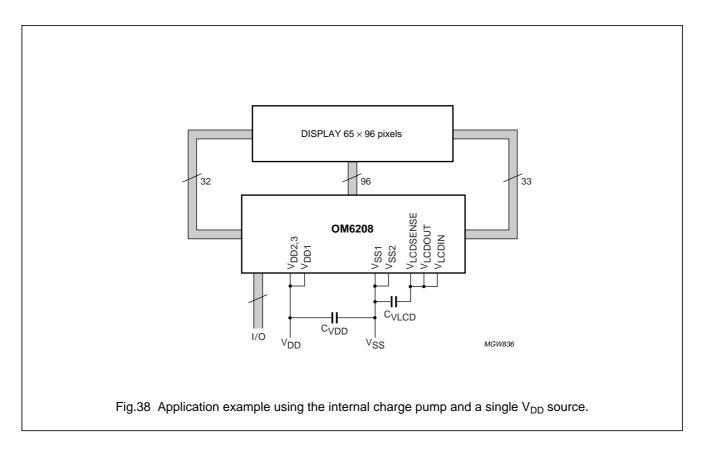
The pinning of the OM6208 has an optimal design for single plane wiring, e.g. for chip-on-glass display modules.

### 16.3 Application examples

In the following application examples, the required values of the external capacitors are:

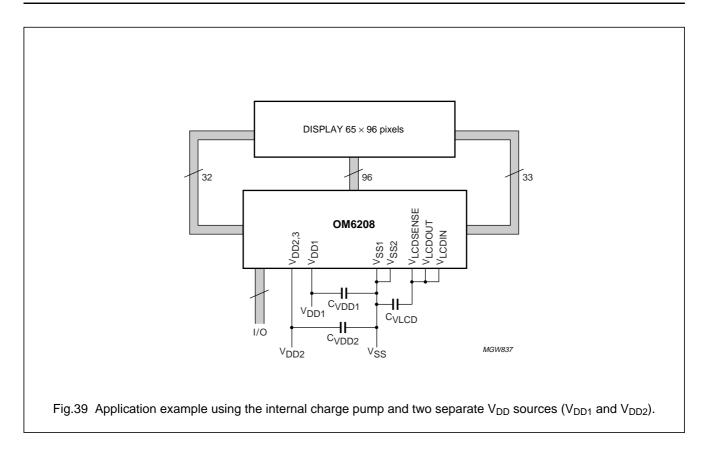
- $C_{VLCD} = 1 \mu F minimum$
- $C_{VDD}$ ,  $C_{VDD1}$  and  $C_{VDD2}$  = 1  $\mu F$  minimum
- Higher capacitor values can be used for the supply.

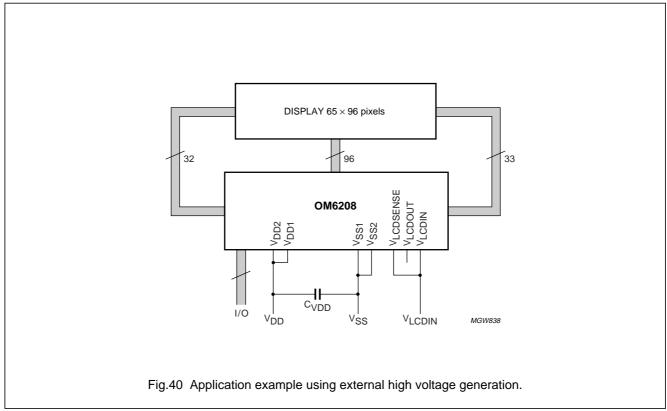
When the internal charge pump is used, the  $V_{LCD}$  lines must be short-circuited externally to ensure that the resistance between pads is zero. This is to allow the bias system to work correctly when BS[2:0] is not 000.



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#### 17 MODULE MAKER PROGRAMMING

One Time Programmable (OTP) technology has been implemented in the OM6208. This enables the module maker to program some extended features of the OM6208 after it has been assembled on an LCD module. Programming is made under the control of the interfaces and the use of one special pin. This pin must be made available on the module glass but does not need to be accessed by the set maker.

The OM6208 features the following module maker programmable parameters:

- V<sub>LCD</sub> calibration
- · Default temperature compensation slopes
- · Default charge pump multiplication factor
- Default V<sub>PR</sub> value
- Default bias levels BS[2:0]
- Default frame frequency range in grey-scale mode GFR[2:0]
- Default oscillator tuning in grey-scale mode GT[2:0]
- Default frame frequency in black-and-white mode SFR[2:0]
- Default oscillator tune in black-and-white mode ST[2:0]
- Default N-line inversion NL[6:0]
- Default frame inversion FI
- Enable factory default FD
- · Seal bit.

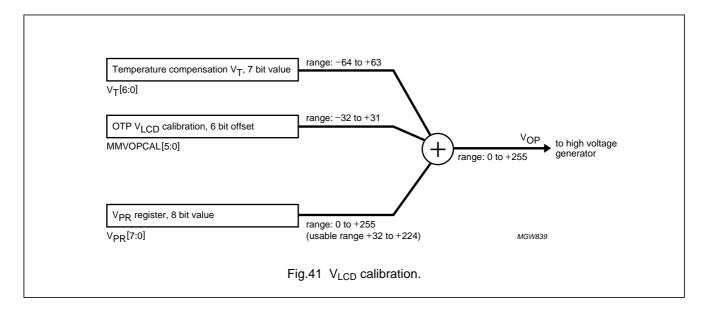
### 17.1 V<sub>LCD</sub> calibration

The first OTP feature included is the ability to tune the  $V_{LCD}$  voltage with a 6-bit code (MMVOPCAL). This code is implemented in twos complement notation giving rise to a positive or negative offset to the  $V_{PR}$  register. The adder in the circuit has underflow and overflow protection. In the event of an overflow, the output will be clamped to 255; with an underflow, the output will be clamped to logic 0.

The final control to the high voltage generator,  $V_{OP}$ , will be the sum of all the calibration registers according to Section 11.10, equation (4).

Table 20 V<sub>LCD</sub> calibration

	MI	MVC	PC	AL		DECIMAL	V <sub>LCD</sub> OFFSET
5	4	3	2	1	0	EQUIVALENT	(mV)
0	1	1	1	1	1	31	930
0	1	1	1	1	0	30	900
0	1	1	1	0	1	29	870
:	:	:	:	:	:	:	:
0	0	0	0	1	0	2	60
0	0	0	0	0	1	1	30
0	0	0	0	0	0	0	0
1	1	1	1	1	1	-1	-30
1	1	1	1	1	0	-2	-60
:	:	:	:	:	:	:	:
1	0	0	0	1	0	-30	-900
1	0	0	0	0	1	-31	-930
1	0	0	0	0	0	-32	-960



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### 17.2 Factory defaults

### 17.2.1 CONFIGURATION DERIVED FROM OTP CELLS

In some instances it is desirable that the configuration is derived from OTP cells and not from user-configurable registers. It is therefore possible to pre-define the following features using the OTP facility:

- · Default temperature compensation slopes
- · Default charge pump multiplication factor
- Default V<sub>PR</sub> value
- Default bias levels BS[2:0]
- Default frame frequency range in grey-scale mode GFR[2:0]
- Default oscillator tune in grey-scale mode GT[2:0]
- Default frame frequency in black-and-white mode SFR[2:0]
- Default oscillator tune in black-and-white mode ST[2:0]

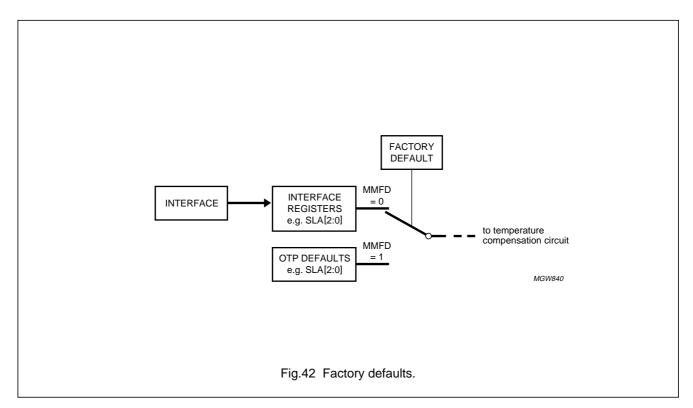
- Default N-line inversion NL[6:0]
- · Default frame inversion FI
- · Enable factory default FD
- · Seal bit.

The selection of the mode for factory defaults is made by setting the factory default OTP cell bit MMFD.

Table 21 Factory default bit MMFD

MMFD	ACTION
0	configuration data is taken from the interface
1	OTP values are used for configuration data

The operation can be shown as a switch that selects between two sources of data (see Fig.42). When the OTP defaults are selected, changing the default values via the interface is not possible.



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#### 17.2.2 DEFAULTS FROM INTERFACE REGISTERS

Factory defaults available from user-configurable registers are as follows:

- Temperature slope selection values are set by SLA[2:0], SLB[2:0], SLC[2:0] and SLD[2:0]
- Default charge pump multiplication factor value is set by S[1:0]
- Default V<sub>PR</sub> value is set by V<sub>PR</sub>[7:0]
- Default bias level values are set by BS[2:0]
- Grey-scale mode default frame frequency and tuning values are set by GFR[2:0] and GT[2:0]
- Black-and-white mode default frame frequency and tuning values are set by SFR[2:0] and ST[2:0].

### 17.3 Seal bit

The module maker programming is performed in a special mode: the calibration mode (CALMM). This mode is entered via a special interface command, CALMM. To prevent wrongful programming, a seal bit has been implemented which prevents the device from entering the

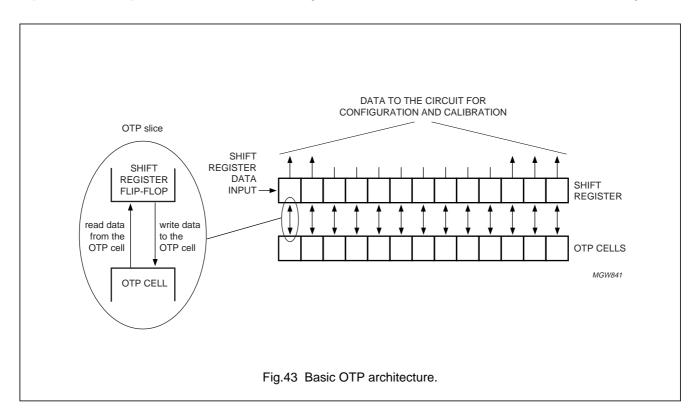
calibration mode. This seal bit, once programmed, cannot be reversed, thus further changes in programmed values are not possible. Applying the programming voltages when not in CALMM mode has no effect on the programmed values.

Table 22 Seal bit definition

SEAL BIT	ACTION
0	possible to enter calibration mode
1	calibration mode disabled

#### 17.4 OTP architecture

The OTP circuitry in the OM6208 contains many bits of data. The circuitry for one bit is called an OTP slice. Each OTP slice consists of two main parts: the OTP cell (a non-volatile memory cell) and the shift register cell (a flip-flop). The OTP cells are accessible only through their shift register cells: both reading from and writing to the OTP cells are performed with the shift register cells, but only the shift register cells are visible to the rest of the circuit. The basic OTP architecture is shown in Fig.43.



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#### 17.4.1 OTP OPERATIONAL EFFECTS

The OTP architecture allows the following operations:

- Reading data from the OTP cells. The content of the non-volatile OTP cells is transferred to the shift register where upon it may affect the OM6208 operation.
- Writing data to the OTP cells. First, all 9 bits of data are shifted into the shift register via the interface. Then the content of the shift register is transferred to the OTP cells (there are some limitations related to storing data in these cells, see Section 17.7).
- Checking calibration without writing to the OTP cells.
   Shifting data into the shift register allows the effects on the V<sub>LCD</sub> voltage to be observed.

The reading of data from the OTP cells is initiated by writing to the DON register. The OTP cells will not be updated until the device leaves power down and the oscillator starts. The reading operation needs up to 5 ms to complete.

The shifting of the data into the shift register is performed in the special mode CALMM. In the OM6208, the CALMM mode is entered through the CALMM command. Once in the CALMM mode the data is shifted into the shift register via the interface at the rate of 1-bit per command. After transmitting the last bit and exiting the CALMM mode the serial interface is again in the normal mode and all other commands can be sent. Care should be taken that always all bits of data (or a multiple of all bits) are transferred before exiting the CALMM mode, otherwise the bits will be in the wrong positions.

In the shift register the value of the seal bit is, like the other bits, always zero at reset. To make sure the security feature works correctly, the CALMM command is disabled until a Power-down mode has been left. Once a refresh is completed, the seal bit value in the shift register is valid and permission to enter CALMM mode can thus be determined.

The bits are shifted into the shift register in a predefined order as shown in Table 23.

**Table 23** OTP bit order (See Fig.44 for a graphical representation)

POSITION	OTP CELL	POSITION	OTP CELL	POSITION	OTP CELL
1	MMVPR[7]	19	MMSLA[1]	37	MMGT[1]
2	MMVPR[6]	20	MMSLA[0]	38	MMGT[0]
3	MMVPR[5]	21	MMS[1]	39	MMSFR[2]
4	MMVPR[4]	22	MMS[0]	40	MMSFR[1]
5	MMVPR[3]	23	MMBS[2]	41	MMSFR[0]
6	MMVPR[2]	24	MMBS[1]	42	MMST[2]
7	MMVPR[1]	25	MMBS[0]	43	MMST[1]
8	MMVPR[0]	26	MMFD	44	MMST[0]
9	MMSLD[2]	27	MMVOPCAL[5]	45	MMNL[6]
10	MMSLD[1]	28	MMVOPCAL[4]	46	MMNL[5]
11	MMSLD[0]	29	MMVOPCAL[3]	47	MMNL[4]
12	MMSLC[2]	30	MMVOPCAL[2]	48	MMNL[3]
13	MMSLC[1]	31	MMVOPCAL[1]	49	MMNL[2]
14	MMSLC[0]	32	MMVOPCAL[0]	50	MMNL[1]
15	MMSLB[2]	33	MMGFR[2]	51	MMNL[0]
16	MMSLB[1]	34	MMGFR[1]	52	MMFI
17	MMSLB[0]	35	MMGFR[0]	53	SEAL
18	MMSLA[2]	36	MMGT[2]	_	_

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#### 17.5 Interface commands

Table 24 OTP instructions

These instructions are in addition to those in the Instruction set, Table 7.

NAME	D/C		ACTION							
NAIVIE	NAME D/C		D6	D5	D4	D3	D2	D1	D0	ACTION
OTP programming	0	1	1	1	1	0	0	OSE	CALMM	enter calibration mode and control programming
DON (refresh)	0	1	0	1	0	1	1	1	DON	display ON/OFF
Load 0	0	1	1	0	1	1	0	0	0	write 0 to shift register
Load 1	0	1	1	0	1	1	0	0	1	write 1 to shift register

#### 17.5.1 CALMM INSTRUCTION

This instruction enters the device into the calibration mode. This mode enables the shift register for loading and allows programming of the non-volatile OTP cells to take place. If the seal bit is set, then this mode cannot be accessed and the instruction will be ignored. Once in calibration mode, data may be loaded into the shift register via the 'LOAD0' and 'LOAD1' instructions (on the falling edge of SCLK).

The CALMM mode may be left by setting the CALMM bit to logic 0. Reset will also clear this mode.

The programming can only take place when OTP Switch Enable (OSE) has been set to logic 1. This bit enables the  $V_{OTPPROG}$  input to be passed to the OTP cells. This allows  $V_{OTPGATE}$  to be tied to SCLH/SCE on the module for normal operation. Reset will also clear this mode.

#### 17.5.2 REFRESH INSTRUCTION

The action of the 'refresh' instruction is to force the OTP shift register to reload from the non-volatile OTP cells. This instruction takes up to 5 ms to complete. During this time all other instructions may be sent.

In the OM6208 the 'refresh' instruction is associated with the 'DON' instruction so that the shift register is automatically refreshed every time DON is enabled or disabled.

Note: If this instruction is sent while in power save mode, the DON bit will be updated but the refreshing is delayed until the device leaves power-down.

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### 17.6 Example of filling the shift register

An example sequence of commands and data is shown in Table 25. In this example the shift register is filled with the following data: MMVPR = 11010000, and the seal bit is logic 0.

It is assumed that the OM6208 has just been reset. After transmitting the last bit, the OM6208 can exit or remain in CALMM mode (see step 1). Note that while in CALMM

mode the interface does not recognize commands in the normal sense.

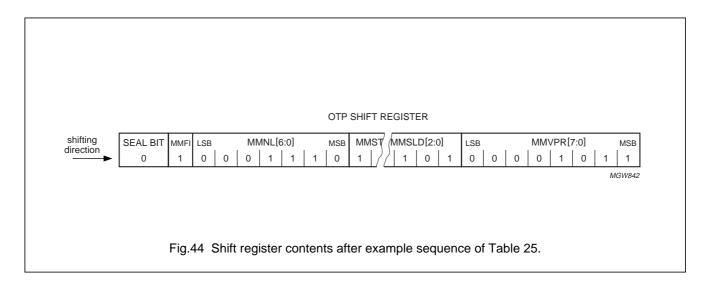
After this sequence has been applied it is possible to observe the impact of the data shifted in. The sequence described is not useful for OTP programming because the number of bits with the value logic 1' is greater than that allowed for programming (see Section 17.7). The shift register contents after this action are shown in Fig.44.

Table 25 Example sequence for filling the shift register

STEP	D/C	D7	D6	D5	D4	D3	D2	D1	D0	ACTION
1	0	1	0	1	0	1	1	1	1	exit power-down
2										wait 5 ms for refresh to take effect
3	0	1	1	1	1	0	0	0	1	enter CALMM mode
4	0	1	1	0	1	1	0	0	1	shift in data, first bit is MMVPR[7]; note 1
5	0	1	1	0	1	1	0	0	1	MMVPR[6]
6	0	1	1	0	1	1	0	0	0	MMVPR[5]
7	0	1	1	0	1	1	0	0	1	MMVPR[4]
8	0	1	1	0	1	1	0	0	0	MMVPR[3]
9	0	1	1	0	1	1	0	0	0	MMVPR[2]
10										
:	:	:	:	:	:	:	:	:	:	
57	0	1	1	0	1	1	0	0	1	MMFI
58	0	1	1	0	1	1	0	0	0	seal bit
59	0	1	1	1	1	0	0	0	0	exit CALMM mode

### Note

1. The data for the bits is not in the correct shift register position until all bits have been sent.



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### 17.7 Programming flow

Programming is achieved whilst in CALMM mode and with the application of the programming voltages. As the data for programming the OTP cell is contained in the corresponding shift register cell, the shift register cell must be loaded with a logic 1 in order to program the corresponding OTP cell. If the shift register cell contains a logic 0, then no action will take place when the programming voltages are applied.

Once an OTP cell is programmed it cannot be de-programmed. An already programmed cell (an OTP cell containing a logic 1) must not be reprogrammed.

A sequence of commands and data for OTP programming is shown as an example in Table 26.

Although the order for programming cells is not significant, it is recommended that the seal bit is programmed last. Once this bit has been programmed it will not be possible to re-enter the CALMM mode.

During programming, a substantial current flows in the  $V_{LCDIN}$  pin. For this reason it is recommended programming only one OTP cell at a time. This is achieved by filling all but one shift register cells with logic 0.

The programming specification refers to the voltages at the chip pads, therefore the contact resistance is significant and must be considered by the user.

Table 26 Sequence for OTP programming

This sequence assumes the OM6208 has just been reset.

						1				
STEP	D/C	D7	D6	D5	D4	D3	D2	D1	D0	ACTION
1	0	1	0	1	0	1	1	1	1	exit power-down (DON = 1)
2										wait 5 ms for refresh to take effect
3	0	1	1	1	1	0	0	1	1	enter CALMM mode and OSE
4	0	1	1	0	1	1	0	0	1	shift-in data, MMVPR[7] is first bit; note 1
5	0	1	1	0	1	1	0	0	0	MMVPR[6]
6	0	1	1	0	1	1	0	0	0	MMVPR[5]
7	0	1	1	0	1	1	0	0	0	MMVPR[4]
7	0	1	1	0	1	1	0	0	0	MMVPR[3]
9	0	1	1	0	1	1	0	0	0	MMVPR[2]
10										
:	:	:	:	:	:	:	:	:	:	
58	0	1	1	0	1	1	0	0	0	MMFI
59	0	1	1	0	1	1	0	1	0	seal bit
60										apply programming voltage at pins V <sub>OTPPROG</sub> and V <sub>LCDIN</sub> according to Section 17.8
Re	epeat st	eps 5 to	60 for	each bit	that sh	ould be	prograr	nmed to	1	
61										apply external reset

### Note

1. The data for the bits is not in the correct shift register position until all the bits have been sent.

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### 17.8 Programming specification

Table 27 Programming specification (refer to Fig.45)

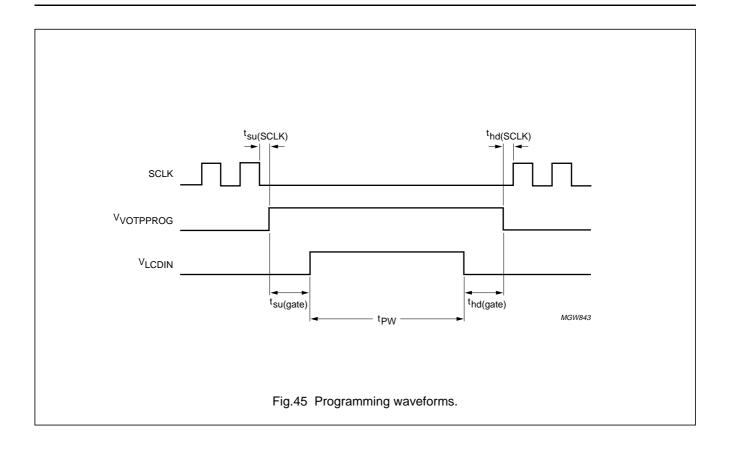
SYMBOL	PARAMETER	CONDITION	MIN.	TYP.	MAX.	UNIT
V <sub>OTPPROG</sub>	voltage applied to pad V <sub>OTPPROG</sub>	V <sub>OTPPROG</sub> relative to V <sub>SS1</sub> ; note 1				
		programming active	11.0	11.5	12.0	V
		programming inactive	$V_{SS} - 0.2$	0	$V_{DD1}$	V
V <sub>LCDIN</sub>	voltage applied to pad V <sub>LCDIN</sub>	V <sub>LCDIN</sub> relative to V <sub>SS1</sub> ; notes 1 and 2				
		programming active	9.0	9.5	10.0	V
		programming inactive	V <sub>SS2</sub>	$V_{DD2}$	4.5	V
I <sub>LCDIN</sub>	current drawn by pad V <sub>LCDIN</sub> during programming	when programming a single bit to logic 1	_	850	1000	μΑ
I <sub>OTPPROG</sub>	current drawn by pad V <sub>OTPPROG</sub> during programming		_	100	200	μΑ
T <sub>PROG</sub>	ambient temperature during programming		0	25	40	°C
t <sub>su(SCLK)</sub>	internal data set-up time after last clock		1	_	_	μs
t <sub>hd(SCLK)</sub>	internal data hold time before next clock		1	_	_	μs
t <sub>su(gate)</sub>	V <sub>OTPPROG</sub> gate set-up time prior to programming		1	_	10	μs
t <sub>hd(gate)</sub>	V <sub>OTPPROG</sub> gate hold time after programming		1	_	10	μs
t <sub>PW</sub>	pulse width of programming voltage		100	120	200	ms

### **Notes**

- 1. The voltage drop across the ITO track and zebra connector must be taken into account to guarantee sufficient voltage at the chip pads.
- 2. The Power-down mode (DON = 0 and DAL = 1) and CALMM mode must be active while the V<sub>LCDIN</sub> input is being driven.

# 65 x 96 pixels matrix grey-scale LCD driver

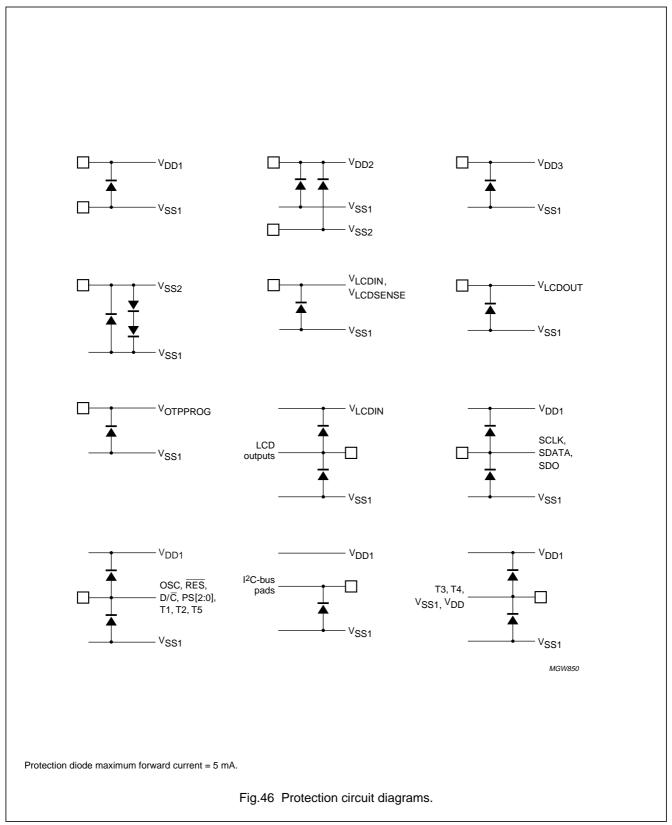
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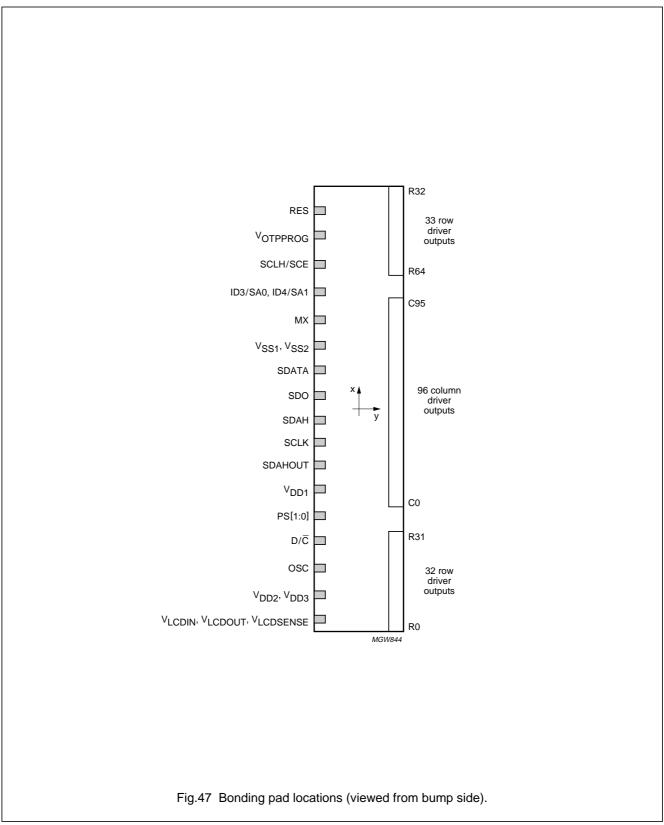
### 18 DEVICE PROTECTION DIAGRAM



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### 19 BONDING PAD INFORMATION



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**Table 28** Bonding pad locations All x and y coordinates indicate pad centres and are referenced to the centre of the chip; dimensions in  $\mu m$  (see Fig.47).

CVMDOL	PAD	COORE	COORDINATES		
SYMBOL		х	у		
dummy	1	-4919	+1279		
dummy	2	-4856	+1279		
alignment mark	3	-4762	+1220		
dummy	4	-4649	+1279		
V <sub>LCDIN</sub>	5	-4586	+1279		
V <sub>LCDIN</sub>	6	-4523	+1279		
V <sub>LCDIN</sub>	7	-4460	+1279		
V <sub>LCDIN</sub>	8	-4397	+1279		
V <sub>LCDOUT</sub>	9	-4323	+1279		
V <sub>LCDOUT</sub>	10	-4260	+1279		
V <sub>LCDOUT</sub>	11	-4197	+1279		
V <sub>LCDOUT</sub>	12	-4134	+1279		
V <sub>LCDOUT</sub>	13	-4071	+1279		
V <sub>LCDOUT</sub>	14	-4008	+1279		
V <sub>LCDOUT</sub>	15	-3945	+1279		
V <sub>LCDSEN</sub>	16	-3882	+1279		
$V_{DD2}$	17	-3566	+1279		
$V_{DD2}$	18	-3503	+1279		
$V_{DD2}$	19	-3440	+1279		
$V_{DD2}$	20	-3377	+1279		
$V_{DD2}$	21	-3314	+1279		
$V_{DD2}$	22	-3251	+1279		
$V_{DD2}$	23	-3188	+1279		
$V_{DD2}$	24	-3125	+1279		
$V_{DD2}$	25	-3062	+1279		
V <sub>DD2</sub>	26	-2999	+1279		
$V_{DD3}$	27	-2928	+1279		
V <sub>DD3</sub>	28	-2865	+1279		
$V_{DD3}$	29	-2802	+1279		
OSC	30	-2639	+1279		
DC	31	-2475	+1279		
PS1	32	-2345	+1279		
PS0	33	-2217	+1279		
V <sub>DD1</sub>	34	-2102	+1279		
V <sub>DD1</sub>	35	-2039	+1279		
V <sub>DD1</sub>	36	-1976	+1279		

SYMBOL	PAD	COORD	COORDINATES		
STWIBOL	PAD	х	у		
V <sub>DD1</sub>	37	-1913	+1279		
$V_{DD1}$	38	-1850	+1279		
V <sub>DD1</sub>	39	-1787	+1279		
dummy	40	-1669	+1279		
dummy	41	-1503	+1279		
dummy	42	-1337	+1279		
dummy	43	-1172	+1279		
SDAHOUT	44	-1006	+1279		
dummy	45	-829	+1279		
SCLK	46	-657	+1279		
dummy	47	-468	+1279		
dummy	48	-303	+1279		
dummy	49	-137	+1279		
dummy	50	29	1 279		
dummy	51	194	1 279		
SDAH	52	360	1 279		
SDO	53	605	1 279		
SDATA	54	696	1 279		
V <sub>SS2</sub>	55	795	1 279		
V <sub>SS2</sub>	56	858	1 279		
V <sub>SS2</sub>	57	921	1 279		
V <sub>SS2</sub>	58	984	1 279		
V <sub>SS2</sub>	59	1047	1 279		
V <sub>SS2</sub>	60	1110	1 279		
V <sub>SS2</sub>	61	1173	1 279		
V <sub>SS1</sub>	62	1246	1 279		
V <sub>SS1</sub>	63	1309	1 279		
V <sub>SS1</sub>	64	1372	1 279		
V <sub>SS1</sub>	65	1435	1 279		
V <sub>SS1</sub>	66	1498	1 279		
V <sub>SS1</sub>	67	1561	1 279		
MX	68	1693	1 279		
T3	69	1865	1 279		
T4	70	2028	1 279		
T1	71	2114	1 279		
T2	72	2232	1 279		
T5	73	2350	1 279		
T6	74	2468	1 279		
ID3_SA0	75	2586	1 279		

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CVMDOL	DAD	COORD	INATES	
SYMBOL	PAD	х	у	
ID4_SA1	76	2704	1 279	
VDD1	77	2849	1 279	
dummy	78	3005	1 279	
dummy	79	3171	1 279	
dummy	80	3336	1 279	
dummy	81	3502	1 279	
dummy	82	3667	1 279	
SCLH	83	3833	1 279	
V <sub>OTPPROG</sub>	84	3958	1 279	
V <sub>OTPPROG</sub>	85	4021	1 279	
V <sub>OTPPROG</sub>	86	4084	1 279	
dummy	87	4243	1 279	
RES	88	4387	1 279	
dummy	89	4489	1 279	
dummy	90	4552	1 279	
dummy	91	4615	1 279	
dummy	92	4678	1 279	
alignment mark	93	4799	1220	
bumps alignment mark	94	4902	1258	
dummy	95	+4900	-1275	
dummy	96	+4849	-1275	
dummy	97	+4797	-1275	
dummy	98	+4745	-1275	
dummy	99	+4693	-1275	
dummy	100	+4641	-1275	
dummy	101	+4589	-1275	
dummy	102	+4538	-1275	
dummy	103	+4486	-1275	
dummy	104	+4434	-1275	
R32	105	+4330	-1275	
R33	106	+4278	-1275	
R34	107	+4227	-1275	
R35	108	+4175	-1275	
R36	109	+4123	-1275	
R37	110	+4071	-1275	
R38	111	+4019	-1275	
R39	112	+3967	-1275	
R40	113	+3916	-1275	
R41	114	+3864	-1275	

SYMBOL	DAD	COORD	COORDINATES		
STWBUL	PAD	х	у		
R42	115	+3812	-1275		
R43	116	+3760	-1275		
R44	117	+3708	-1275		
R45	118	+3656	-1275		
R46	119	+3604	-1275		
R47	120	+3553	-1275		
R48	121	+3501	-1275		
R49	122	+3449	-1275		
R50	123	+3397	-1275		
R51	124	+3345	-1275		
R52	125	+3293	-1275		
R53	126	+3242	-1275		
R54	127	+3190	-1275		
R55	128	+3138	-1275		
R56	129	+3086	-1275		
R57	130	+3034.	-1275		
R58	131	+2982	-1275		
R59	132	+2931	-1275		
R60	133	+2879	-1275		
R61	134	+2829	-1275		
R62	135	+2775	-1275		
R63	136	+2723	-1275		
R64	137	+2671	-1275		
VC	138	+2620	-1275		
dummy	139	+2568	-1275		
dummy	140	+2516	-1275		
Т8	141	+2413	-1275		
T7	142	+2360	-1275		
V1_L	143	+2308	-1275		
C95	144	+2257	-1275		
C94	145	+2205	-1275		
C93	146	+2153	-1275		
C92	147	+2101	-1275		
C91	148	+2049	-1275		
C90	149	+1997	-1275		
C89	150	+1946	-1275		
C88	151	+1894	-1275		
C87	152	+1842	-1275		
C86	153	+1790	-1275		

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CVMDOL	PAD	COORDINATES		
SYMBOL	PAD	х	у	
C85	154	+1738	-1275	
C84	155	+1686	-1275	
C83	156	+1635	-1275	
C82	157	+1583	-1275	
C81	158	+1531	-1275	
C80	159	+1479	-1275	
C79	160	+1427	-1275	
C78	161	+1375	-1275	
C77	162	+1324	-1275	
C76	163	+1272	-1275	
C75	164	+1219.86	-1275	
C74	165	+1168	-1275	
C73	166	+1116	-1275	
C72	167	+1064	-1275	
C71	168	+961	-1275	
C70	169	+909	-1275	
C69	170	+857	-1275	
C68	171	+805	-1275	
C67	172	+753	-1275	
C66	173	+701	-1275	
C65	174	+650	-1275	
C64	175	+598	-1275	
C63	176	+546	-1275	
C62	177	+494	-1275	
C61	178	+442	-1275	
C60	179	+390	-1275	
C59	180	+339	-1275	
C58	181	+287	-1275	
C57	182	+235	-1275	
C56	183	+183	-1275	
C55	184	+131	-1275	
C54	185	+79	-1275	
C53	186	+28	-1275	
C52	187	-24	-1275	
C51	188	-76	-1275	
C50	189	-128	-1275	
C49	190	-180	-1275	
C48	191	-232	-1275	
dummy	192	-283	-1275	

CVMDOL	DAD	COORDINATES		
SYMBOL	PAD	х	у	
dummy	193	-335	-1275	
C47	194	-387	-1275	
C46	195	-439	-1275	
C45	196	-491	-1275	
C44	197	-543	-1275	
C43	198	-595	-1275	
C42	199	-646	-1275	
C41	200	-698	-1275	
C40	201	-750	-1275	
C39	202	-802	-1275	
C38	203	-854	-1275	
C37	204	-906	-1275	
C36	205	-957	-1275	
C35	206	-1 009	-1275	
C34	207	-1061	-1275	
C33	208	-1113	-1275	
C32	209	-1165	-1275	
C31	210	-1217	-1275	
C30	211	-1268	-1275	
C29	212	-1320	-1275	
C28	213	-1372	-1275	
C27	214	-1424	-1275	
C26	215	-1476	-1275	
C25	216	-1528	-1275	
C24	217	-1579	-1275	
C23	218	-1683	-1275	
C22	219	-1735	-1275	
C21	220	-1787	-1275	
C20	221	-1839	-1275	
C19	222	-1891	-1275	
C18	223	-1942	-1275	
C17	224	-1994	-1275	
C16	225	-2046	-1275	
C15	226	-2098	-1275	
C14	227	-2150	-1275	
C13	228	-2202	-1275	
C12	229	-2253	-1275	
C11	230	-2305	-1275	
C10	231	-2357	-1275	

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SYMBOL	PAD	COORDINATES	
STWIBOL	PAD	х	у
C9	232	-2409	-1275
C8	233	-2461	-1275
C7	234	-2513	-1275
C6	235	-2564	-1275
C5	236	-2616	-1275
C4	237	-2668	-1275
C3	238	-2720	-1275
C2	239	-2772	-1275
C1	240	-2824	-1275
C0	241	-2875	-1275
dummy	242	-2927	-1275
dummy	243	-2979	-1275
V1_H	244	-3031	-1275
V2_L	245	-3083	-1275
V2_H	246	-3135	-1275
R31	247	-3187	-1275
R30	248	-3238	-1275
R29	249	-3290	-1275
R28	250	-3342	-1275
R27	251	-3394	-1275
R26	252	-3446	-1275
R25	253	-3498	-1275
R24	254	-3549	-1275
R23	255	-3601	-1275
R22	256	-3653	-1275

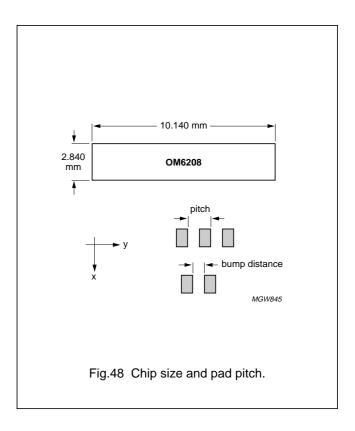
CYMPOL	DAD	COORD	ORDINATES	
SYMBOL	PAD	х	у	
R21	257	-3705	-1275	
R20	258	-3757	-1275	
R19	259	-3809	-1275	
R18	260	-3860	-1275	
R17	261	-3912	-1275	
R16	262	-3964	-1275	
R15	263	-4016	-1275	
R14	264	-4068	-1275	
R13	265	-4120	-1275	
R12	266	-4171	-1275	
R11	267	-4223	-1275	
R10	268	-4275	-1275	
R9	269	-4327	-1275	
R8	270	-4379	-1275	
R7	271	-4431	-1275	
R6	272	-4483	-1275	
R5	273	-4534	-1275	
R4	274	-4586	-1275	
R3	275	-4638	-1275	
R2	276	-4690	-1275	
R1	277	-4742	-1275	
R0	278	-4794	-1275	
dummy	279	-4845	-1275	
dummy	280	-4897	-1275	

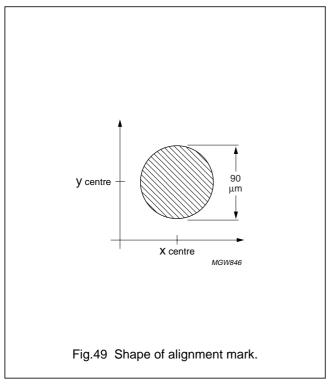
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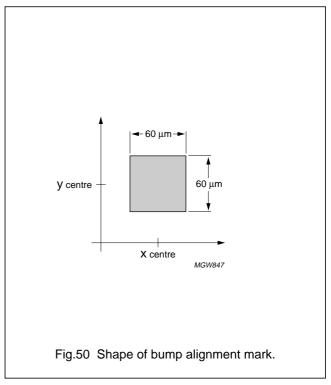
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Table 29 Chip information

ITEM	ROW/COL SIDE	UNIT		
Row/column side				
Pad pitch	51.84 (minimum)	μm		
CBB opening	15.3 × 5.4	μm		
Bump dimensions	30.0 × 99 (±3)	μm		
Bump height	15.0	μm		
Minimum bump distance	21.8	μm		
Wafer thickness (excl. bumps)	381 (±25)	μm		
Interface side				
Pad pitch	63 (minimum)	μm		
CBB opening	25.7 × 5.4	μm		
Bump dimensions	42 × 90 (±3)	μm		
Bump height	15.0	μm		
Minimum bump distance	shortened bumps: 21 normal bumps: 22	μm		
Wafer thickness (excl. bumps)	381 (±25)	μm		



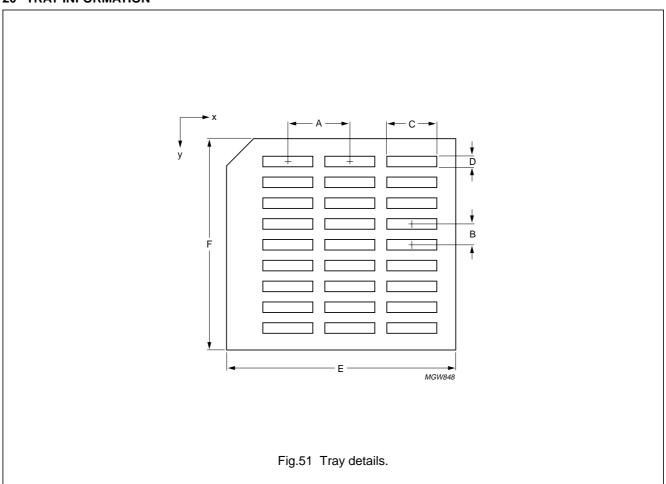


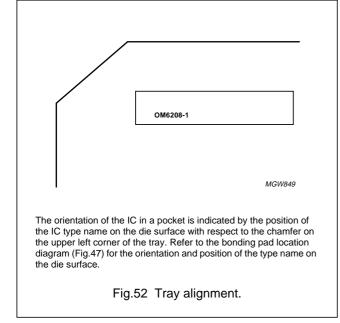


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### **20 TRAY INFORMATION**





## Table 30 Tray dimensions

DIM.	DESCRIPTION	VALUE
Α	pocket pitch, x direction	14.25 mm
В	pocket pitch, y direction	4.87 mm
С	pocket width, x direction	10.24 mm
D	pocket width, y direction	2.94mm
Е	tray width, x direction	50.80 mm
F	tray width, y direction	50.80 mm
х	number of pockets in X direction	3
У	number of pockets in Y direction	9

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#### 21 DATA SHEET STATUS

LEVEL	DATA SHEET STATUS <sup>(1)</sup>	PRODUCT STATUS(2)(3)	DEFINITION
I	Objective data	Development	This data sheet contains data from the objective specification for product development. Philips Semiconductors reserves the right to change the specification in any manner without notice.
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#### **Notes**

- 1. Please consult the most recently issued data sheet before initiating or completing a design.
- 2. The product status of the device(s) described in this data sheet may have changed since this data sheet was published. The latest information is available on the Internet at URL http://www.semiconductors.philips.com.
- 3. For data sheets describing multiple type numbers, the highest-level product status determines the data sheet status.

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**Short-form specification** — The data in a short-form specification is extracted from a full data sheet with the same type number and title. For detailed information see the relevant data sheet or data handbook.

Limiting values definition — Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 60134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

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Purchase of Philips I<sup>2</sup>C components conveys a license under the Philips' I<sup>2</sup>C patent to use the components in the I<sup>2</sup>C system provided the system conforms to the I<sup>2</sup>C specification defined by Philips. This specification can be ordered using the code 9398 393 40011.

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**NOTES** 

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**NOTES** 

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**NOTES** 

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