# TEA1623P; TEA1623PH

# STARplug switched mode power supply controller IC Rev. 3 — 30 August 2010 Product

Product data sheet

#### **General description** 1.

The TEA1623 is a Switched Mode Power Supply (SMPS) controller IC that operates directly from rectified universal mains. It is implemented in the high voltage EZ-HV SOI process, combined with a low voltage BiCMOS process.

The device includes a high voltage power switch and a circuit for start-up directly from the rectified mains voltage. A dedicated circuit for valley switching is built in, which makes a very efficient slim-line electronic power-plug concept possible.

In its most basic version of application, the TEA1623 acts as a voltage source. Here, no additional secondary electronics are required. A combined voltage and current source can be realized with minimum costs for external components. Implementation of the TEA1623 renders an efficient and low cost power supply system.

#### Features and benefits 2.

- Designed for general purpose power supplies
- Integrated power switch: 6.5 Ω and 650 V
- Operates from universal AC mains supplies: 80 V to 276 V
- Adjustable frequency for flexible design
- RC oscillator for stable output regulation
- Valley switching for minimum switch-on loss
- Frequency reduction at low power output for low standby power: < 100 mW</p>
- Adjustable OverCurrent Protection (OCP)
- UnderVoltage Protection (UVP)
- Temperature protection
- Short-winding protection
- Safe restart mode for system fault conditions
- Simple application with both primary and secondary (opto) feedback
- Available in 8-pin and 16-pin DIP packages.

### **Applications**

- Adapters
- Set-Top Box (STB)
- DVD
- VCD
- CD(R)
- PC Silverbox standby SMPS



### 4. Quick reference data

Table 1. Quick reference data

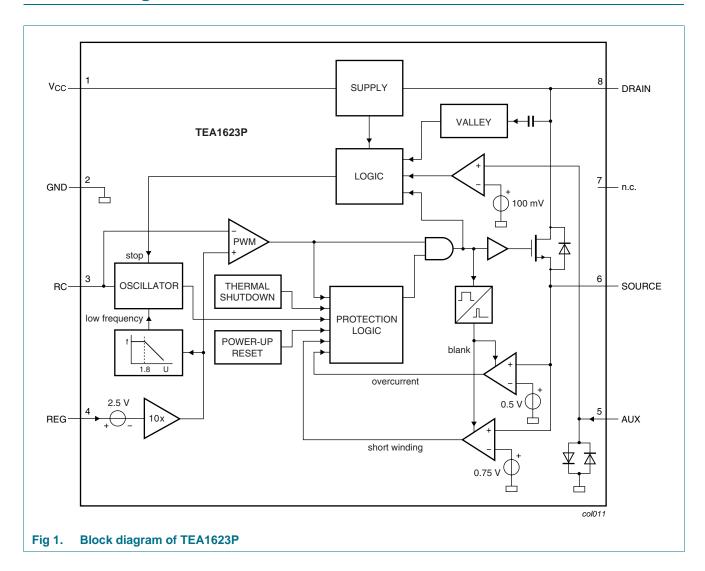
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{CC(max)}$	maximum supply voltage		-	-	40	V
V <sub>DRAIN(max)</sub>	maximum DMOS power transistor drain voltage	T <sub>j</sub> > 0 °C	-	-	650	V
I <sub>DRAIN</sub>	supply current drawn from pin DRAIN	no auxiliary supply	-	0.5	-	mA
R <sub>DSon</sub>	drain-source on-state resistance	$I_{SOURCE} = -0.5 A$				
		T <sub>j</sub> = 25 °C	-	6.5	7.5	Ω
		T <sub>j</sub> = 100 °C	-	9.0	10.0	Ω
f <sub>osc</sub>	oscillator frequency range		10	-	200	kHz

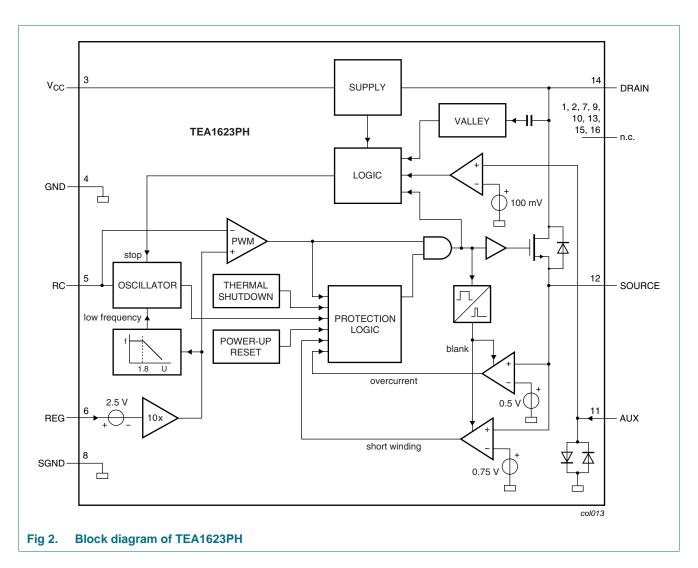
### 5. Ordering information

Table 2. Ordering information

Type number	Package		
	Name	Description	Version
TEA1623P	DIP8	plastic dual in-line package; 8 leads (300 mil)	SOT97-1
TEA1623PH	DIP16	plastic dual in-line package; 16 leads (300 mil); long body	SOT38-1

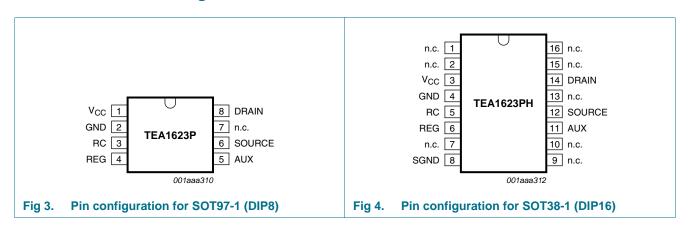
### 6. Block diagram





### 7. Pinning information

### 7.1 Pinning



TEA1623P\_TEA1623PH

### 7.2 Pin description

Table 3. Pin description

Symbol	Pin		Description		
	TEA1623P	TEA1623PH	1		
V <sub>CC</sub>	1	3	supply voltage		
GND	2	4	ground		
RC	3	5	frequency setting		
REG	4	6	regulation point		
SGND	-	8	signal ground; preferably connected to pin GND		
AUX	5	11	input for voltage from auxiliary winding for timing (demagnetization)		
SOURCE	6	12	source of internal MOS switch		
n.c.	7	1, 2, 7, 9, 10, 13, 15, 16	not connected		
DRAIN	8	14	drain of internal MOS switch; input for start-up current and valley sensing		

### 8. Functional description

The TEA1623 is the heart of a compact flyback converter, with the IC placed at the primary side. The auxiliary winding of the transformer can be used for indirect feedback to control the isolated output. This additional winding also powers the IC. A more accurate control of the output voltage and/or current can be implemented with an additional secondary sensing circuit and optocoupler feedback.

The TEA1623 uses voltage mode control. The frequency is determined by the maximum transformer demagnetizing time or the frequency of the oscillator. In the first case, the converter operates in the Self Oscillating Power Supply (SOPS) mode. In the latter case, it operates at a constant frequency, which can be adjusted with external components  $R_{RC}$  and  $C_{RC}.$  This mode is called Pulse Width Modulation (PWM). Furthermore, a primary stroke is started only in a valley of the secondary ringing. This valley switching principle minimizes capacitive switch-on losses.

### 8.1 Start-up and UnderVoltage LockOut (UVLO)

Initially, the IC is self supplying from the rectified mains voltage. The IC starts switching as soon as the voltage on pin  $V_{CC}$  passes the  $V_{CC(start)}$  level. The supply is taken over by the auxiliary winding of the transformer as soon as  $V_{CC}$  is high enough and the supply from the line is stopped for high efficiency operation.

As soon as the voltage on pin  $V_{CC}$  drops below the  $V_{CC(stop)}$  level, the IC stops switching and restarts from the rectified mains voltage.

#### 8.2 Oscillator

The frequency of the oscillator is set by the external resistor and capacitor on pin RC. The external capacitor is charged rapidly to the  $V_{RC(max)}$  level and, starting from a new primary stroke, it discharges to the  $V_{RC(min)}$  level. Because the discharge is exponential, the

relative sensitivity of the duty factor to the regulation voltage at low duty factor is almost equal to the sensitivity at high duty factors. This results in a more constant gain over the duty factor range compared to PWM systems with a linear sawtooth oscillator. Stable operation at low duty factors is easily realized. For high efficiency, the frequency is reduced as soon as the duty factor drops below a certain value. This is accomplished by increasing the oscillator charge time.

To ensure that the capacitor can be charged within the charge time, the value of the oscillator capacitor should be limited to approximately 1 nF.

### 8.3 Duty factor control

The duty factor is controlled by the internal regulation voltage and the oscillator signal on pin RC. The internal regulation voltage is equal to the external regulation voltage (–2.5 V) multiplied by the gain of the error amplifier (typical 20 dB or 10×).

The minimum duty factor of the switched mode power supply is 0 %. The maximum duty factor is set to 75 % (typical value at 100 kHz oscillation frequency).

#### 8.4 Valley switching

A new cycle is started at the primary stroke when the switch is switched on (see <u>Figure 5</u>). After a certain time (determined by the RC oscillator voltage and the internal regulation level), the switch is turned off and the secondary stroke starts. The internal regulation level is determined by the voltage on pin REG. After the secondary stroke, the drain

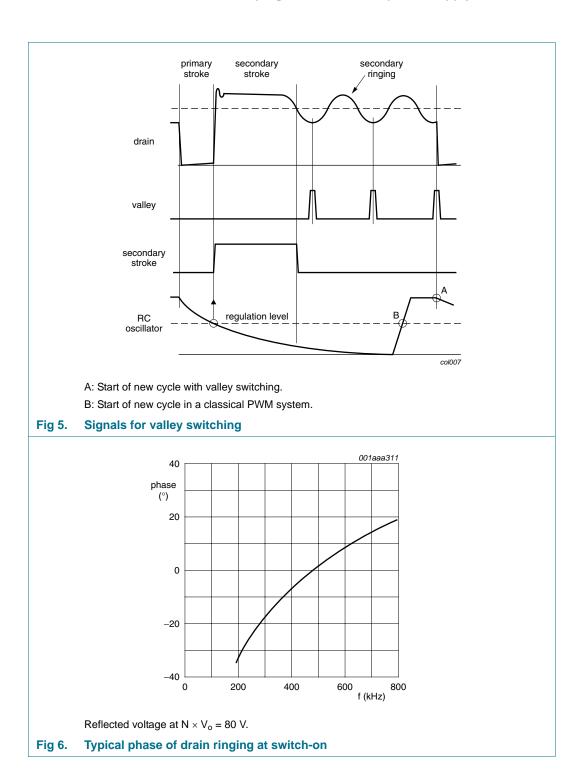
voltage shows an oscillation with a frequency of approximately  $\frac{1}{2\pi \times \sqrt{L_p C_p}}$  , where:

 $L_p$  is the primary self inductance on the drain node.

C<sub>p</sub> is the parasitic capacitance on the drain node.

As soon as the oscillator voltage is high again and the secondary stroke has ended, the circuit waits for a low drain voltage before starting a new primary stroke.

The primary stroke starts some time before the actual valley at low ringing frequencies, and some time after the actual valley at high ringing frequencies. Figure 6 shows a typical curve for a reflected voltage N  $\times$  V<sub>o</sub> of 80 V. This voltage is the output voltage V<sub>o</sub> (see Figure 7) transferred to the primary side of the transformer with the factor N (determined by the turns ratio of the transformer). Figure 6 shows that the system switches exactly at minimum drain voltage for ringing frequencies of 480 kHz, thus reducing the switch-on losses to a minimum. At 200 kHz, the next primary stroke is started at 33  $^{\circ}$  before the valley. The switch-on losses are still reduced significantly.



### 8.5 Demagnetization

The system operates in discontinuous conduction mode all the time. As long as the secondary stroke has not ended, the oscillator will not start a new primary stroke. During the suppression time  $t_{\text{suppr}}$ , demagnetization recognition is suppressed. This suppression may be necessary in applications where the transformer has a large leakage inductance and at low output voltages.

#### 8.6 Protection

#### 8.6.1 Overcurrent protection

The cycle-by-cycle peak drain current limit circuit uses the external source resistor  $R_l$  (see Figure 7) to measure the current. The circuit is activated after the leading edge blanking time  $t_{leb}$ . The protection circuit limits the source voltage to  $V_{source(max)}$ , and thus limits the primary peak current.

#### 8.6.2 Short-winding protection

The short-winding protection circuit is also activated after the leading edge blanking time. If the source voltage exceeds the short-winding protection voltage  $V_{\text{swp}}$ , the TEA1623 stops switching. Only a power-on reset will restart normal operation. The short-winding protection also protects in case of a secondary diode short circuit.

#### 8.6.3 OverTemperature Protection (OTP)

An accurate temperature protection is provided in the TEA1623. When the junction temperature exceeds the thermal shut-down temperature, the IC stops switching. During thermal protection, the IC current is lowered to the start-up current. The IC continues normal operation as soon as the overtemperature situation has disappeared.

#### 8.6.4 OverVoltage Protection (OVP)

Overvoltage protection can be achieved in the application by pulling pin REG above its normal operation level, or by keeping the level of pin AUX above  $V_{demag}$ . The current primary stroke is terminated immediately and no new primary stroke is started until the voltage on pin REG drops to its normal operation level. Pin REG has an internal clamp. The current feed into pin REG must be limited.

### 8.7 Characteristics of the complete power-plug

#### 8.7.1 Input

The input voltage range comprises the universal AC mains from 80 V to 276 V.

#### 8.7.2 Accuracy

The accuracy of the complete converter, functioning as a voltage source with primary sensing, is approximately 8 % (mainly dependent on the transformer coupling). The accuracy with secondary sensing is defined by the accuracy of the external components. For safety requirements in case of optocoupler feedback loss, the primary sensing remains active when an overvoltage circuit is connected.

#### 8.7.3 Efficiency

An efficiency of 75 % at maximum output power can be achieved for a complete converter designed for universal mains.

### **8.7.4** Ripple

A minimum ripple is obtained in a system designed for a maximum duty factor of 50 % under normal operating conditions and a minimized dead time. The magnitude of the ripple in the output voltage is determined by the frequency and duty factor of the converter, the output current level, and the value and ESR of the output capacitor.

### **Limiting values**

Table 4. **Limiting values** 

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
Voltage					
$V_{CC}$	supply voltage	continuous	<u>[1]</u> –0.4	+40	V
V <sub>RC</sub>	oscillator input voltage		[1] -0.4	+3	V
V <sub>SOURCE</sub>	DMOS power transistor source voltage		-0.4	+5	V
$V_{DRAIN}$	DMOS power transistor drain voltage	T <sub>j</sub> > 0 °C	-0.4	+650	V
Current					
I <sub>REG</sub>	regulation input current		[2] _	6	mA
I <sub>AUX</sub>	auxiliary winding input current		<u>[2]</u> –10	+5	mA
I <sub>RC</sub>	oscillator capacitor charge current		-3	-	mA
I <sub>SOURCE</sub>	source current		-2	+2	Α
I <sub>DRAIN</sub>	drain current		-2	+2	Α
General					
P <sub>tot</sub>	total power dissipation				
		TEA1623P; T <sub>amb</sub> < 45 °C	-	1.0	W
		TEA1623PH; T <sub>amb</sub> < 50 °C	-	1.7	W
T <sub>stg</sub>	storage temperature		<b>–55</b>	+150	°C
Tj	junction temperature		-20	+145	°C
V <sub>ESD</sub>	electrostatic discharge voltage	human body model[3]			
		pin DRAIN	-1500	+1500	V
		all other pins	-2000	+2000	V
		machine model[4]			
		all pins	-200	+200	V

<sup>[1]</sup> Pins  $V_{CC}$  and RC are not allowed to be current driven.

### 10. Thermal characteristics

Table 5. **Thermal characteristics** 

Symbol	Parameter	Conditions	Тур	Unit
$R_{th(j-a)}$	thermal resistance from junction to ambient	in free air	<u>[1]</u>	-
		TEA1623P	100	K/W
		TEA1623PH	55	K/W

Thermal resistance R<sub>th(j-a)</sub> can be lower when pin GND is connected to sufficient copper area on the printed-circuit board. See the TEA152x application note for details.

TEA1623P\_TEA1623PH

<sup>[2]</sup> Pins REG and AUX are not allowed to be voltage driven.

Human body model: equivalent to discharging a 100 pF capacitor through a 1.5 k $\Omega$  series resistor.

Machine model: equivalent to discharging a 200 pF capacitor through a 0.75  $\mu$ H coil and a 10  $\Omega$  series resistor.

### 11. Characteristics

#### Table 6. Characteristics

 $T_{amb}$  = 25 °C; no overtemperature; all voltages are measured with respect to ground; currents are positive when flowing into the IC, unless otherwise specified.

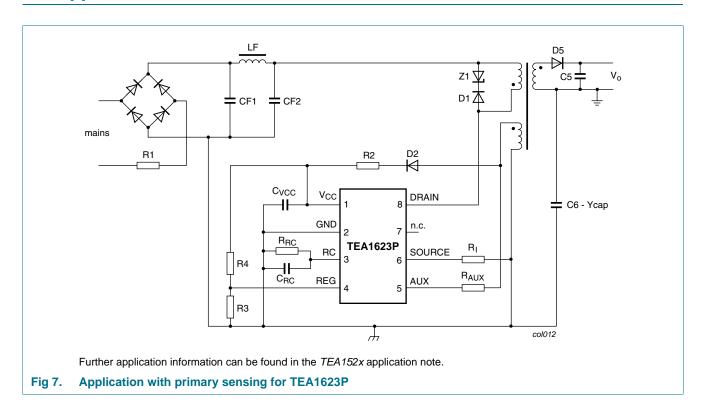
Symbol	Parameter	Conditions	Min	Тур	Max	Uni
Supplies						
Supply on p	in V <sub>CC</sub>					
V <sub>CC(start)</sub>	start voltage		9	9.5	10	V
V <sub>CC(stop)</sub>	stop voltage	undervoltage lockout	7.0	7.5	8.0	V
I <sub>CC(operate)</sub>	operating supply current	normal operation	-	1.3	1.9	mΑ
I <sub>CC(startup)</sub>	start-up supply current	start-up	-	180	400	μΑ
I <sub>CC(ch)</sub>	charging current	V <sub>DRAIN</sub> > 60 V				
		$V_{CC} = 0 V$	-650	-520	-390	μΑ
		V <sub>CC</sub> = 8.5 V	-375	-275	-175	μΑ
Supply on p	in DRAIN					
I <sub>DRAIN</sub>	supply current drawn from pin	no auxiliary supply	-	0.5	-	mΑ
	DRAIN	with auxiliary supply; $V_{DRAIN} > 60 \text{ V}$	-	30	125	μΑ
PWM mode						
$\delta_{min}$	minimum duty factor		-	0	-	%
$\delta_{\sf max}$	maximum duty factor	f <sub>osc</sub> = 100 kHz	-	75	-	%
Self-oscilla	ting power supply mode					
V <sub>demag</sub>	demagnetization recognition voltage level		50	100	150	mV
t <sub>suppr</sub>	time of suppression of transformer ringing at start of secondary stroke		1.0	1.5	2.0	μS
Oscillator:	pin RC					
$V_{RC(min)}$	minimum voltage of RC oscillator setting		60	75	90	mV
$V_{RC(max)}$	maximum voltage of RC oscillator setting		2.4	2.5	2.6	V
t <sub>RC(ch)</sub>	RC charging time		-	1	-	μS
f <sub>osc</sub>	oscillator frequency range		10	-	200	kHz
Duty factor	regulator: pin REG					
$V_{REG}$	input voltage on pin REG		2.4	2.5	2.6	V
G <sub>V(erroramp)</sub>	voltage gain of error amplifier		-	20	-	dB
V <sub>REG(clamp)</sub>	clamping voltage on pin REG	I <sub>REG</sub> = 6 mA	-	-	7.5	V
Valley swite	ching recognition					
dV/dt <sub>valley</sub>	valley recognition		-102	-	+102	V/µ
f <sub>valley</sub>	ringing frequency for valley switching	$N \times V_o = 100 \text{ V}$	200	550	800	kH
t <sub>d(valley-on)</sub>	delay from valley recognition to switch-on		-	150	-	ns
Output stag	ge (FET)					
I <sub>L(drain)</sub>	drain leakage current	$V_{DRAIN} = 650 \text{ V}$	-	-	125	μΑ
EA1623P_TEA1623F	PH All informati	ion provided in this document is subject to legal disclaimers.		© NXP	B.V. 2010. All ri	ghts rese

Table 6. Characteristics ... continued

 $T_{amb}$  = 25 °C; no overtemperature; all voltages are measured with respect to ground; currents are positive when flowing into the IC, unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>BR(drain)</sub>	drain breakdown voltage	T <sub>j</sub> > 0 °C	650	-	-	V
R <sub>DSon</sub>	drain-source on-state resistance	$I_{\text{SOURCE}} = -0.5 \text{ A}$				
		T <sub>j</sub> = 25 °C	-	6.5	7.5	Ω
		T <sub>j</sub> = 100 °C	-	9.0	10.0	Ω
t <sub>drain(f)</sub>	drain fall time	V <sub>DRAIN(switch_on)</sub> = 300 V; no external capacitor at pin DRAIN	-	75	-	ns
Temperatur	e protection					
T <sub>prot(max)</sub>	maximum threshold temperature		150	160	170	°C
T <sub>prot(hys)</sub>	threshold temperature hysteresis		-	2	-	°C
Overcurren	t and short winding protection: pi	1 SOURCE				
V <sub>source(max)</sub>	overcurrent protection voltage	$dV/dt = 0.1 V/\mu s$	0.47	0.50	0.53	V
V <sub>swp</sub>	short-winding protection voltage	$dV/dt = 0.5 V/\mu s$	0.7	0.75	8.0	V
t <sub>d(propagation)</sub>	delay from detecting $V_{\text{source}(\text{max})}$ to switch-off	$dV/dt = 0.5 V/\mu s$	-	160	185	ns
t <sub>leb</sub>	leading edge blanking time	overcurrent and short-winding protection	250	350	450	ns

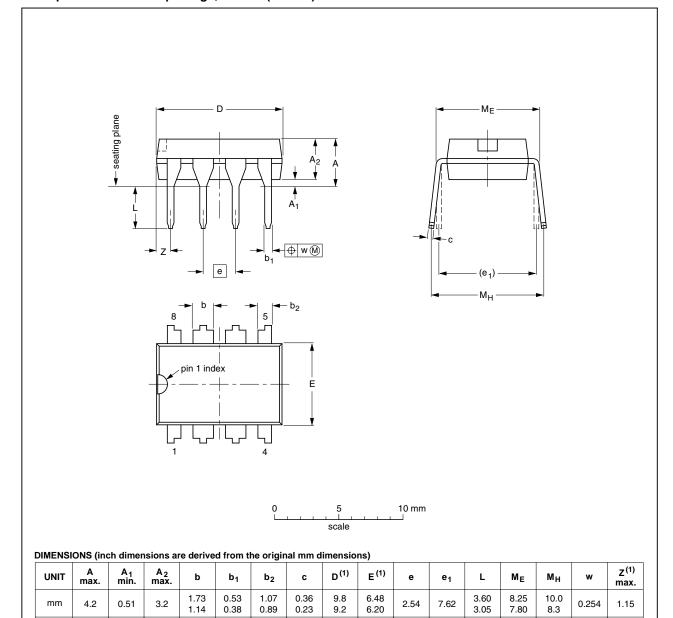
### 12. Application information



### 13. Package outline

### DIP8: plastic dual in-line package; 8 leads (300 mil)

SOT97-1



### Note

inches

0.17

1. Plastic or metal protrusions of 0.25 mm (0.01 inch) maximum per side are not included.

0.021

0.015

0.042

0.035

0.014

0.009

0.068

0.045

OUTLINE		REFERENCES			EUROPEAN ISSUE DATE		
VERSION	IEC	JEDEC JEITA			PROJECTION	ISSUE DATE	
SOT97-1	050G01	MO-001	SC-504-8			<del>99-12-27</del> 03-02-13	

#### Fig 8. Package outline SOT97-1 (DIP8)

0.02

0.13

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0.01

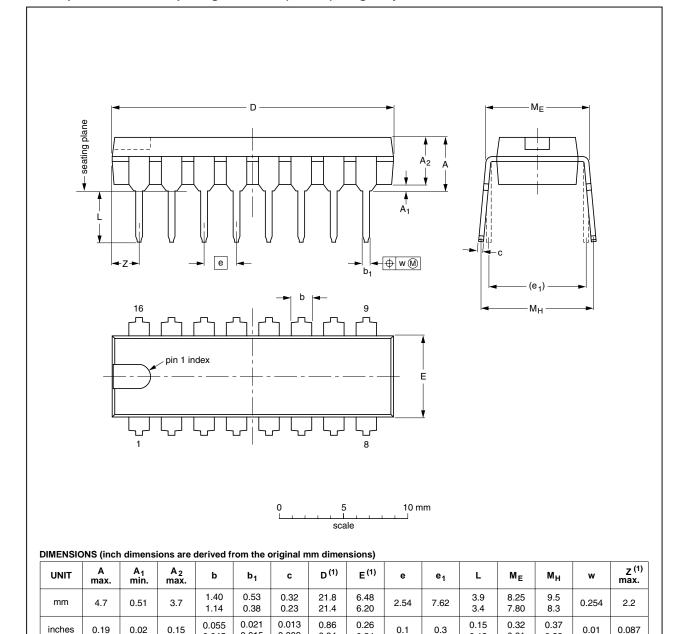
0.045

0.32

0.3

DIP16: plastic dual in-line package; 16 leads (300 mil); long body

SOT38-1



#### Note

1. Plastic or metal protrusions of 0.25 mm (0.01 inch) maximum per side are not included.

0.045

0.015

0.009

OUTLINE		REFERENCES			EUROPEAN	ISSUE DATE	
VERSION	IEC	JEDEC	JEITA		PROJECTION	ISSUE DATE	
SOT38-1	050G09	MO-001	SC-503-16			<del>99-12-27</del> 03-02-13	

0.84

Fig 9. Package outline SOT38-1 (DIP16)

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### 14. Soldering of through-hole mount packages

### 14.1 Introduction to soldering through-hole mount packages

This text gives a very brief insight into wave, dip and manual soldering.

Wave soldering is the preferred method for mounting of through-hole mount IC packages on a printed-circuit board.

### 14.2 Soldering by dipping or by solder wave

Driven by legislation and environmental forces the worldwide use of lead-free solder pastes is increasing. Typical dwell time of the leads in the wave ranges from 3 seconds to 4 seconds at 250 °C or 265 °C, depending on solder material applied, SnPb or Pb-free respectively.

The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature  $(T_{stg(max)})$ . If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

### 14.3 Manual soldering

Apply the soldering iron (24 V or less) to the lead(s) of the package, either below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300  $^{\circ}$ C it may remain in contact for up to 10 seconds. If the bit temperature is between 300  $^{\circ}$ C and 400  $^{\circ}$ C, contact may be up to 5 seconds.

### 14.4 Package related soldering information

Table 7. Suitability of through-hole mount IC packages for dipping and wave soldering

Package	Soldering method			
	Dipping	Wave		
CPGA, HCPGA	-	suitable		
DBS, DIP, HDIP, RDBS, SDIP, SIL	suitable	suitable <sup>[1]</sup>		
PMFP[2]	-	not suitable		

<sup>[1]</sup> For SDIP packages, the longitudinal axis must be parallel to the transport direction of the printed-circuit board.

<sup>[2]</sup> For PMFP packages hot bar soldering or manual soldering is suitable.

### 15. Abbreviations

#### Table 8. Abbreviations

Acronym	Description
BiCMOS	Bipolar CMOS
CMOS	Complementary Metal-Oxide Semiconductor
DMOS	Diffusion Metal-Oxide Semiconductor
ESR	Equivalent Series Resistance
OCP	OverCurrent Protection
OTP	OverTemperature Protection
OVP	OverVoltage Protection
PWM	Pulse Width Modulation
SMPS	Switched Mode Power Supply
SOI	Silicon On Insulator
SOPS	Self Oscillating Power Supply
STB	Set-Top Box
UVLO	UnderVoltage LockOut
UVP	UnderVoltage Protection
VCD	Video Compact Disc

### 16. Revision history

### Table 9. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes		
TEA1623P_TEA1623PH v.3	20100830	Product data sheet	-	TEA1623P_TEA1623PH_2		
Modifications:	<ul> <li><u>Table 1</u> "Qui</li> </ul>	ck reference data" update	ed.			
	• Table 4 "Lim	iting values" updated.				
TEA1623P_TEA1623PH_2	20091104	Product data sheet	-	TEA1623P_TEA1623PH_1		
Modifications:		of this data sheet has bee f NXP Semiconductors.	n redesigned to cor	mply with the new identity		
	<ul> <li>Legal texts h</li> </ul>	nave been adapted to the	new company nam	e where appropriate.		
	<ul> <li>Pin configuration drawing A and B split into <u>Figure 3 "Pin configuration for SOT97-1</u> (DIP8)" and Figure 4 "Pin configuration for SOT38-1 (DIP16)".</li> </ul>					
	<ul><li>Section 15 "</li></ul>	Abbreviations" added.				
TEA1623P_TEA1623PH_1	20040317	Product data sheet	-	-		

### 17. Legal information

#### 17.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
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