February 2002

LM2792

Current Regulated Switched Capacitor LED Driver with Analog Brightness Control

General Description

The LM2792 is a CMOS charge-pump voltage doubler and regulator that provides two regulated current sources. They are designed to drive two white (or blue) LEDs with matched currents (within \pm 0.3%) to produce balanced light sources for display backlights. The LM2792 accepts an input voltage range from 3.0V to 5.8V and maintain a constant current determined by an external set resistor.

The LM2792 delivers up to 34mA of load current to accommodate two high forward voltage (typically white) LEDs. The switching frequency is 900kHz (min.) to keep the conducted noise spectrum away from sensitive frequencies within portable RF devices.

The LM2792 offers full off to maximum current control through the BRGT pin. The output current linearly tracks the BRGT pin voltage. The LM2792 is available in active high or low shutdown versions. The shutdown pin reduces the operating current to $1\mu A$ (max.).

The LM2792 is available in a 10 pin leadless leadframe (LLP) CSP package.

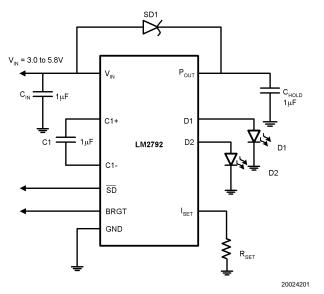
Features

- Output matching of ± 0.3% (typ.)
- Drives up to two LED's
- 3.0V to 5.8V Input Voltage
- Up to 34mA output current
- Soft start limits inrush current
- Analog brightness control
- Separate shutdown input
- Very small solution size and no inductor
- 1.4mA typical operating current
- 1µA (max.) shutdown current
- 900kHz switching frequency (min.)
- Linear regulation generates predictable noise spectrum
- LLP-10 package: 3mm X 3mm X 0.8mm

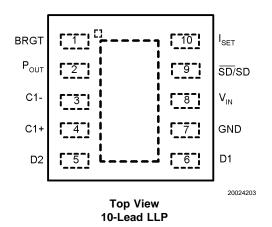
Applications

- White LED Display Backlights
- White LED Keypad Backlights
- 1-Cell Li-Ion battery-operated equipment including PDAs, hand-held PCs, cellular phones
- Flat Panel Dispalys

Basic Application Circuit



Connection Diagram



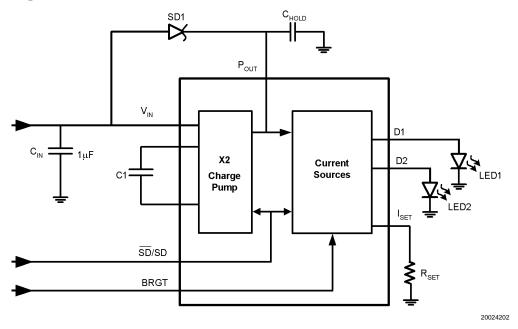
Ordering Information

Order Number	Shutdown Polarity	NSC Package Drawing	Package Marking	Supplied As
LM2792LD-L	Active Low	LLP-10	SRB	1000 Units, Tape and Reel
LM2792LDX-L	Active Low	LLP-10	SRB	4500 Units, Tape and Reel
LM2792LD-H	Active High	LLP-10	SPB	1000 Units, Tape and Reel
LM2792LDX-H	Active High	LLP-10	SPB	4500 Units, Tape and Reel

Pin Description

Pin	Name	Function	
1	BRGT	Variable voltage input controls output current.	
2	P _{OUT}	Charge pump output.	
3	C1-	Connect this pin to the negative terminal of C1.	
4	C1+	Connect this pin to the positive terminal of C1.	
5	D2	Current source outputs. Connect directly to LED.	
6	D1	Current source outputs. Connect directly to LED.	
7	GND	Power supply ground input.	
8	V _{IN}	Power supply voltage input.	
9	SD/SD	Shutdown input. Device operation is inhibited when pin is asserted.	
10	I _{SET}	Current Sense Input. Connect resistor to ground to set constant current through LED.	

Block Diagram



Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

 $\begin{array}{cccc} V_{IN} & & -0.3 \text{ to } 6.0 \text{V} \\ \text{BRGT, SD} & & -0.3 \text{ to } (\text{V}_{IN} + 0.2 \text{V}) \\ \text{Power Dissipation (Note 2)} & & 400 \text{ mW} \\ T_{JMAX} \text{ (Note 2)} & & 150 ^{\circ}\text{C} \\ \theta_{JA} \text{ (Note 7)} & & 55 ^{\circ}\text{C/W} \\ \text{Storge Temperature} & & -65 ^{\circ}\text{C to } +150 ^{\circ}\text{C} \\ \text{Lead Temp. (Soldering, 5 sec.)} & & 260 ^{\circ}\text{C} \end{array}$

ESD Rating
Human Body Model 2KV
Machine Model 200V

Operating Conditions

Temperature

Electrical Characteristics

Limits in standard typeface are for $T_J = 25^{\circ}C$ and limits in **boldface type** apply over the full **Operating Temperature Range**. Unless otherwise specified, $C_1 = C_{IN} = C_{HOLD} = 1 \ \mu F$, $V_{IN} = 3.6 V$, BRGT pin = 2.75 V.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
I _{DX}	Available Current at Output Dx	$3.0V \le V_{IN} \le 5.8V$	14.5	16.8		mA
		$V_{Dx} \le 3.6V$				
		$R_{SET} = 1650\Omega V_{BRGT} = 3V$				
I _{DX}	Line Regulation of Dx Output	$3.0V \le V_{IN} \le 4.4V$	13.5	16	17.8	mA
	Current	$V_{Dx} \le 3.6V$				
		$R_{SET} = 1800\Omega V_{BRGT} = 2.75V$				
V_{DX}	Load Regulation of Dx Output	V _{IN} = 3.6V				mA
	Current	$V_{Dx} = 3.0V$		16.1		
		$V_{Dx} = 4.0V$		15.4		
		$R_{SET} = 1800\Omega V_{BRGT} = 2.75V$				
I _{D-MATCH}	Current Matching Between Any	$V_{D1} = V_{D2} = 3.6V,$		0.3		%
	Two Outputs	BRGT = $2.75V$, $V_{IN} = 3.6V$,				
		$R_{SET} = 1800\Omega$				
IQ	Quiescent Supply Current	$3.0V \le V_{IN} \le 4.4V$, Active, No		1.4	2.7	mA
		Load Current				
I_{SD}	Shutdown Supply Current	$3.0V \le V_{IN} \le 5.5V$, Shutdown		0.1	1	μΑ
		At 85°C		0.3		
V _{IH}	SD Input Logic High	$3.0V \le V_{IN} \le 5.5V$, Note5	0.8* V _{IN}			V
V _{IL}	SD Input Logic Low	3.0V ≤ V _{IN} ≤ 5.5V, Note5			0.2* V _{IN}	V
I _{LEAK-SD}	SD Input Leakage Current	$0V \le V_{SD} \le V_{IN}$		0.01		μA
R _{BRGT}	BRGT Input Resistance			250		kΩ
BRGT	Brightness Voltage Range		0		3.0	V
I _{SET}	I _{SET} Pin Output Current			I _{Dx} /25		mA
f _{SW}	Switching Frequency (Note 4)	$3.0V \le V_{IN} \le 4.4V$	900	1100	1800	kHz
		$I_{D1} = I_{D2} \le 16 \text{mA}$				
t _{START}	Startup Time (Note 6)	I _{Dx} = 90% steady state		10		μs

Note 1: Absolute maximum ratings indicate limits beyond which damage to the device may occur. Electrical specifications do not apply when operating the device beyond its rated operating conditions.

Note 2: D1 and D2 may be shorted to GND without damage. POUT may be shorted to GND for 1sec without damage.

Note 3: In the test circuit, all capacitors are 1.0μF, 0.3Ω maximum ESR capacitors. Capacitors with higher ESR will increase output resistance, reduce output voltage and efficiency.

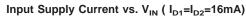
Note 4: The output switches operate at one half of the oscillator frequency, $f_{OSC} = 2f_{SW}$.

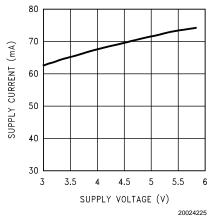
Note 5: The internal thresholds of the Shutdown bar are set at about 40% of $V_{\mbox{\scriptsize IN}}$

Note 6: This electrical specification is quarantee by design

Note 7: For more information regarding the LLP package, please refer to National Semiconductor Application note AN1187

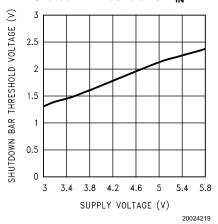
Typical Performance Characteristics Unless otherwise specified, $C_1 = C_{IN} = C_{HOLD} = 1 uF$, $V_{IN} = 3.6 V$, BRGT pin =2.75 V





I_{DIODE} vs. V_{IN} 20 18 16 14 ·I_D @ Rset = 1.8k I_{DIODE} (mA) 12 10 8 0 Rset = 4k 2 0 4.5 5 SUPPLY VOLTAGE (V) 20024224

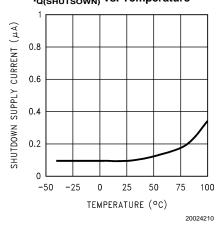
Shutdown Threshold vs. $V_{\rm IN}$

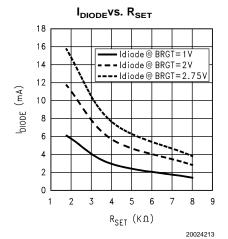


16.50 16.25 16.00 15.75 15.50 -50 -75 0 25 50 75 100 TEMPERATURE (°C)

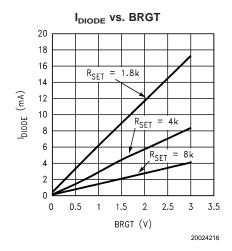
20024207

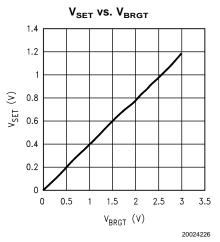
I_{Q(SHUTSOWN)} vs. Temperature

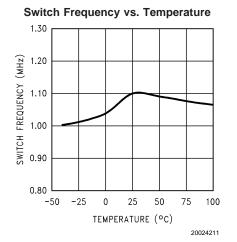


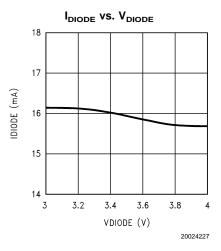


Typical Performance Characteristics Unless otherwise specified, $C_1 = C_{IN} = C_{HOLD} = 1 uF$, $V_{IN} = 3.6 V$, BRGT pin =2.75V (Continued)









Circuit Description

The LM2792 provides two matched current sources for driving high forward voltage drop LEDs from Li-lon battery sources. The part has on-chip current regulators which are composed of current mirrors with a 25 to 1 ratio. The mirrors control the LED current without using current limiting resistors in the LED current path. The device can drive up to 34mA through the Leds.

The LED brightness can be controlled by both analog and/or digital methods. The digital technique uses a PWM (Pulse Width Modulation) signal applied to the shutdown input. The analog technique applies an analog voltage to the brightness (BRGT) pin (see Application Information section).

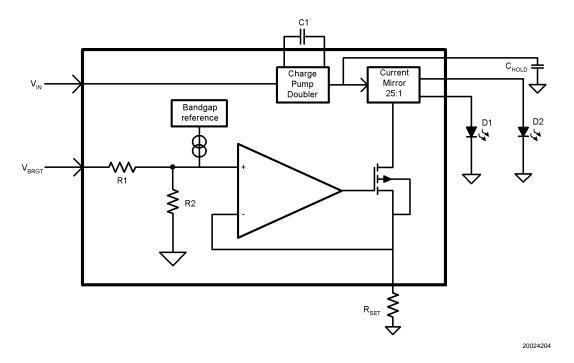


FIGURE 1.

Application Information

Soft Start

LM2792 includes a soft start function to reduce the inrush currents and high peak current during power up of the device. Soft start is implemented internally by ramping the bandgap more slowly than the applied voltage. This is done by holding the bandgap in shutdown for a short time. During soft start, the switch resistances limit the inrush current used to charge the flying and hold capacitors.

Shutdown Mode

A shutdown pin (SD or \overline{SD}) is available to disable the LM2792 and reduce the quiescent current to 1 μ A maximum. The LM2792 is available with both senses of shutdown polarity.

During normal operation mode of the '-L' options, an active high logic signal to the \overline{SD} pin or tying the \overline{SD} pin to $V_{\underline{IN}}$, will enable the device. Pulling \overline{SD} low or connecting \overline{SD} to ground will disable the device.

During normal operation mode of the '-H' options, an active low logic signal to the SD pin or tying the SD pin to GND, will enable the device. Pulling SD high or connecting SD to $V_{\rm IN}$ will disable the device.

Capacitor Selection

Low equivalent series resistance (ESR) capacitors such as X5R or X7R are recommended to be used for $C_{\rm IN}$, C1, and $C_{\rm HOLD}$ for best performance. Ceramic capacitors with less than or equal to 0.3 ohms ESR value are recommended for this application. *Table 1* below lists suggested capacitor suppliers for the typical application circuit.

TABLE 1. Low ESR Capacitor Manufactures

Manufacturer	Contact	website
TDK	(947) 803 6100	www.component.tdk.com
MuRata	(800) 831 9172	www.murata.com
Taiyo Yuden	(800) 348 2496	www.t-yuden.com

Schottky Diode Selection

A schottky diode (SD1) must be used between $V_{\rm IN}$ and $P_{\rm OUT}$ for proper operation. During start-up, the low voltage drop across this diode is used to charge $C_{\rm OUT}$ and start the oscillator. It is necessary to protect the device from turning-on its own parasitic diode and potentially latching-up. As a result, it is important to select a schottky diode that will carry at least 200mA or higher current to charge the output capacitor during start-up. A schottky diode like 1N5817 can

Application Information (Continued)

be used for most applications or a surface mount diode such as BAT54-series and MA2J704 can be used to reduce the circuit size. *Table 2* below lists suggested schottky diode manufactures.

TABLE 2. Diode Manufactures

Manufacturer	Contact	website
ON	(800) 344	www.onsemi.com
Semiconductor	3860	
Phillips	(800) 234	www.philipssemi-
Semiconductors	7381	conductor.com
Panasonic	(408) 945	www.panasonic.com
Semiconductors	5622	

LED selection

The LM2792 is designed to drive LEDs with a forward voltage of about 3.0V to 4.0V or higher. The typical and maximum V_F depends highly on the manufacturer and their technology. Table 3 lists two suggested manufactures and example part numbers. Each supplier makes many LEDs that work well with the LM2792. The LEDs suggested below are in a surface mount package and TOPLED or SIDEVIEW configuration with a maximum forward current of 20mA. These diodes also come in SIDELED or SIDEVIEW configuration and various chromaticity groups. For applications that demand color and brigthness matching, care must be taken to select LEDs from the same chromaticity group. Forward current matching is assured over the LED process variations due to the constant current output of the LM2792. For best fit selection for an application, consult the manufacturer for detailed information.

TABLE 3. White LED Selection:

Component	Manufacture	Contact	
LWT673/LWT67C	Osram	www.osram-os.com	
NSCW100/NSCW215	Nichia	www.nichia.com	

I_{SET} Pin

An external resistor, R_{SET} , sets the mirror current that is required to provide a constant current through the LEDs. The current through R_{SET} and the LED is set by the internal current mirror circuitry with a ratio of 25:1 The currents through each LED are matched within 0.3%. R_{SET} should be chosen not to exceed the maximum current delivery capability of the device. Table 3 shows a list of R_{SET} values when maximun BRGT=3V is applied. For other BRGT voltages, R_{SET} can be calculated using this formula:

$$R_{SET} = ((0.42*BRGT) / I_{SET})*25$$

TABLE 4. R_{SET}Selections (when BRGT pin = 3V maximum)

I _{LED} per LED	R _{SET}
15mA	2.1ΚΩ
10mA	3.15ΚΩ
5mA	6.3ΚΩ

BRGT Pin

The BRGT pin can be used to smoothly vary the brightness of the LEDs. In the LM2792, current on BRGT is connected to an internal resistor divider which gives a factor of 0.42 (see *Figure 1*) . This voltage is fed to the operational amplifier that controls the current through the mirror resistor $R_{\rm SET}.$ The nominal range on BRGT is 0V to 3V.

This means some current must be provided on the BRGT pin or no current will flow through the LEDs. The LM2792 can provide an infinite ratio, from fully off (essentially zero current) to the maximum current set by the R_{SET} resistor. Care must be taken to prevent voltages on BRGT that cause LED current to exceed a total of 34mA. Although this will not cause damage to the IC, it will not meet the guaranteed specifications listed in the Electrical Characteristics.

Calculation of LED Current When Using BRGT:

 $V_{IN} = 3.6V$

 $R_{SET} = 1800\Omega$

 $I_{SET} = ((V_{BRGT} * (0.42) / R_{SET}) * 25$

 $I_{SET} = ((2.75*(0.42)) / 1800)*25 = 16mA$

Note that making $V_{BRGT} = 0V$ results in $I_{SET} \sim = 0$ mA

Brigthness Control using PWM

Brigthness control can be implemented by pulsing a signal at the SD pin. The recommended signal should be between 100Hz to 1kHz. If the operating PWM frequency is much less than 100Hz, flicker may be seen in the LEDs. Likewise, if frequency is much higher, brightness in the LEDs will not be linear. When a PWM signal is used to drive the SD pin of the LM2792, connect BRGT pin to a maximun of 3V to ensure the widest range. Similarly, the voltage at the BRGT pin can be set higher than 3V without damage to the IC, it will not increase the brigthness of the LED significantly. R_{SET} value is selected using the above I _{SET} equation as if BRGT pin is used. The brigthness is controlled by increasing and decreasing the duty cycle of the PWM. Zero duty cycle will turn off the brigthness and a 50% duty cycle waveform produces an average current of 7.5mA if R_{SET} is set to produce a maximum LED current of 15mA. So the LED current varies linearly with the duty cycle.

Thermal Protection

The LM2792 has internal thermal protection circuitry to disable the charge pump if the junction temperature exceeds 150°C. This feature will protect the device from damage due to excessive power dissipation. The device will recover and operate normally when the junction temperature falls below the maximum operating junction temperature of 100°C. It is important to have good thermal conduction with a proper layout to reduce thermal resistance.

Power Efficiency

An ideal power efficiency for a voltage doubler switched capacitor converter is given as the output voltage of the doubler over twice the input voltage as follows:

Efficiency =
$$(V_{DIODE}^* I_{DIODE}) / (V_{IN}^* I_{DIODE}^* Gain) = V_{DIODE} / 2V_{IN}$$

In the case of the LM2792, a more accurate efficiency calculation can be applied as the given formula below.

Efficiency =
$$((V_{D1}^* I_{D1}) + (V_{D2}^* I_{D2})) / (I_{SUPPLY}^* V_{IN})$$

It is clear that the efficiency will depend on the supply voltage in the above equation. As such, the lower the supply voltage, the higher the efficiency.

Application Information (Continued)

Power Dissipation

The maximum allowable power dissipation that this package is capable of handling can be determined as follows:

$$\mathsf{P}_{\mathsf{DMax}} = (\mathsf{T}_{\mathsf{JMax}} - \mathsf{T}_{\mathsf{A}}) \: / \: \theta_{\mathsf{JA}}$$

where T_{JMax} is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction-to-ambient thermal resistance of the specified package.

The actual power dissipation of the device can be calculated using this equation:

$$P_{Dissipation} = (2V_{IN} - V_{DIODE})*I_{LOAD}$$

As an example, if $V_{\rm IN}$ in the target application is 4.2V, $V_{\rm DIODE}$ = 3.0V and worse case current consumption is 32mA (17mA for each diode).

$$P_{Dissipation} = ((2*4.2) -3.0)*0.032 = 173 \text{mW}$$

Power dissipation must be less than that allowed by the package. Please refer to the Absolute Maximum Rating of the LM2792.

PCB Layout Considerations

The LLP is a leadframe based Chip Scale Package (CSP) with very good thermal properties. This package has an exposed DAP (die attach pad) at the center of the package measuring 2.0mm x 1.2mm. The main advantage of this exposed DAP is to offer lower thermal resistance when it is soldered to the thermal land on the PCB. For PCB layout, National highly recommends a 1:1 ratio between the package and the PCB thermal land. To further enhance thermal conductivity, the PCB thermal land may include vias to a ground plane. For more detailed instructions on mounting LLP packages, please refer to National Semiconductor Application Note AN-1187.

Physical Dimensions inches (millimeters) unless otherwise noted (10X 0.5) DIMENSIONS ARE IN MILLIMETERS (10X 0.25) RECOMMENDED LAND PATTERN 1:1 RATION WITH PKG SOLDER PADS **C** 0.8 MAX PIN 1 INDEX AREA

LLP-10 Pin Package (LDA) For Ordering, Refer to Ordering Information Table **NS Package Number LDA10A**

♦ 0.1**%** C A**S** B**S**

LIFE SUPPORT POLICY

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2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



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