

14-Bit Parallel, Low Glitch Multiplying DAC with 4-Quadrant Resistors

November 1998

FEATURES

- DNL and INL: 1LSB Max
- On-Chip 4-Quadrant Resistors Allow Precise 0V to 10V, 0V to -10V or $\pm 10V$ Outputs
- Asynchronous Clear Pin
 - LTC1591: Reset to Zero Scale
 - LTC1591-1: Reset to Midscale
- Glitch Impulse < 2nV-s
- 28-Lead SSOP Package
- Low Power Consumption: 10 μ W Typ
- Power-On Reset

APPLICATIONS

- Process Control and Industrial Automation
- Direct Digital Waveform Generation
- Software-Controlled Gain Adjustment
- Automatic Test Equipment

DESCRIPTION

The LTC[®]1591 is a parallel input 14-Bit multiplying current output DAC that operates from a single 5V supply. INL and DNL are accurate to 1LSB over the industrial temperature range in both 2- and 4-quadrant multiplying modes. 14-bit 4-quadrant multiplication is achieved with on-chip 4-quadrant multiplication resistors.

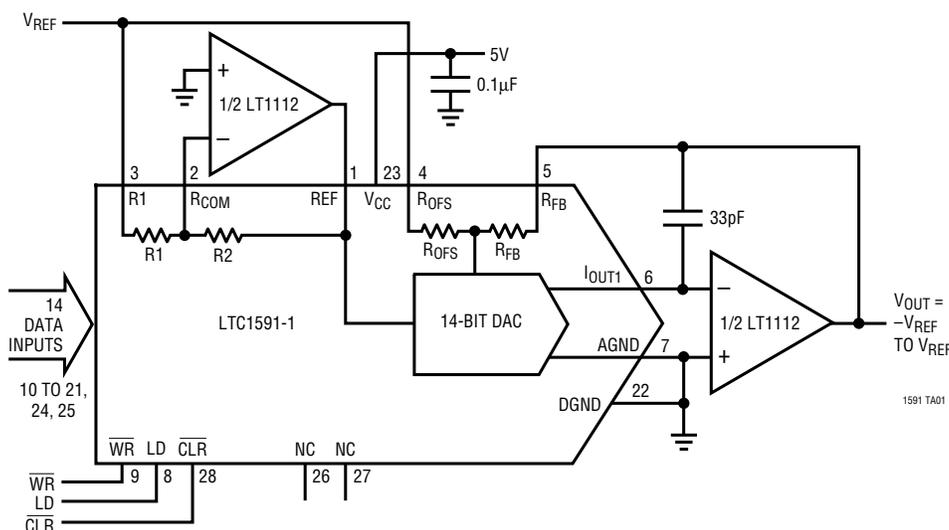
The LTC1591 is available in 28-pin SSOP package and is specified over the commercial and industrial temperature ranges. The device includes an internal deglitcher circuit that reduces the glitch impulse to less than 2nV-s (typ). The asynchronous $\overline{\text{CLR}}$ pin resets the LTC1591 to zero scale and the LTC1591-1 to midscale.

The LTC1591 is pin compatible with the LTC1597, 16-bit parallel current output DAC.

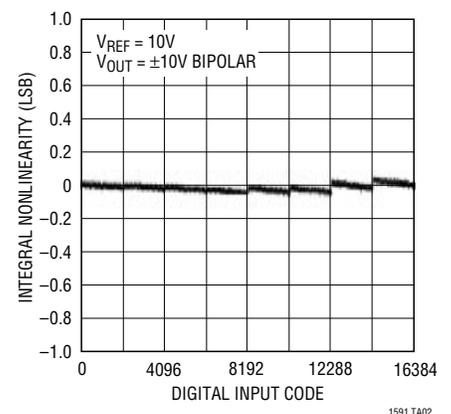
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TYPICAL APPLICATION

14-Bit, 4-Quadrant Multiplying DAC with a Minimum of External Components



LTC1591/LTC1591-1
 Integral Nonlinearity



ABSOLUTE MAXIMUM RATINGS

(Note 1)

V_{CC} to AGND	-0.5V to 7V
V_{CC} to DGND	-0.5V to 7V
AGND to DGND	$V_{CC} + 0.5V$
DGND to AGND	$V_{CC} + 0.5V$
REF, R_{OFS} , R_{FB} , R1, R_{COM} to AGND, DGND	$\pm 25V$
Digital Inputs to DGND	-0.5V to ($V_{CC} + 0.5V$)
I_{OUT1} to AGND	-0.5V to ($V_{CC} + 0.5V$)
Maximum Junction Temperature	125°C
Operating Temperature Range	
Commercial	0°C to 70°C
Industrial	-40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec).....	300°C

PACKAGE/ORDER INFORMATION

TOP VIEW		ORDER PART NUMBER
REF 1	28 CLR	LTC1591CG LTC1591CN LTC1591IG LTC1591IN LTC1591-1CG LTC1591-1CN LTC1591-1IG LTC1591-1IN
R_{COM} 2	27 NC	
R1 3	26 NC	
R_{OFS} 4	25 D0	
R_{FB} 5	24 D1	
I_{OUT1} 6	23 V_{CC}	
AGND 7	22 DGND	
LD 8	21 D2	
WR 9	20 D3	
D13 10	19 D4	
D12 11	18 D5	
D11 12	17 D6	
D10 13	16 D7	
D9 14	15 D8	
G PACKAGE 28-LEAD PLASTIC SSOP N PACKAGE 28-LEAD PDIP $T_{JMAX} = 125^{\circ}C, \theta_{JA} = 95^{\circ}C/W$ (G) $T_{JMAX} = 125^{\circ}C, \theta_{JA} = 70^{\circ}C/W$ (N)		

Consult factory for Military grade parts.

ELECTRICAL CHARACTERISTICS

$V_{DD} = 5V \pm 10\%$, $V_{REF} = 10V$, $I_{OUT1} = AGND = DGND = 0V$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	LTC1591/LTC1591-1			UNITS
			MIN	TYP	MAX	
Accuracy						
	Resolution		●	14		Bits
	Monotonicity		●	14		Bits
INL	Integral Nonlinearity	(Note 2) $T_A = 25^{\circ}C$ T_{MIN} to T_{MAX}	●		± 1	LSB
DNL	Differential Nonlinearity	$T_A = 25^{\circ}C$ T_{MIN} to T_{MAX}	●		± 1	LSB
GE	Gain Error	Unipolar Mode (Note 3) $T_A = 25^{\circ}C$ T_{MIN} to T_{MAX}	●		± 4	LSB
		Bipolar Mode (Note 3) $T_A = 25^{\circ}C$ T_{MIN} to T_{MAX}	●		± 6	LSB
	Gain Temperature Coefficient	(Note 4) $\Delta Gain/\Delta Temperature$	●		1	2 ppm/°C
	Bipolar Zero Error	$T_A = 25^{\circ}C$ T_{MIN} to T_{MAX}	●		± 3	LSB
			●		± 5	LSB
I_{LKG}	OUT1 Leakage Current	(Note 5) $T_A = 25^{\circ}C$ T_{MIN} to T_{MAX}	●		± 5	nA
			●		± 15	nA
PSRR	Power Supply Rejection	$V_{CC} = 5V \pm 10\%$	●		± 0.1	± 1 LSB/V

ELECTRICAL CHARACTERISTICS

$V_{CC} = 5V$, $V_{REF} = 10V$, $I_{OUT1} = AGND = DGND = 0V$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
Reference Input							
R_{REF}	DAC Input Resistance (Unipolar)	(Note 6)	●	4.5	6	10	k Ω
R1/R2	R1/R2 Resistance (Bipolar)	(Notes 6, 13)	●	9	12	20	k Ω
R_{OFS} , R_{FB}	Feedback and Offset Resistances	(Note 6)	●	9	12	20	k Ω
AC Performance (Note 4)							
	Output Current Settling Time	(Notes 7, 8)			1		μ s
	Midscale Glitch Impulse	(Note 12)			2		nV-s
	Digital-to-Analog Glitch Impulse	(Note 9)			1		nV-s
	Multiplying Feedthrough Error	$V_{REF} = \pm 10V$, 10kHz Sine Wave			1		mV _{P-P}
THD	Total Harmonic Distortion	(Note 10)			108		dB
	Output Noise Voltage Density	(Note 11)			10		nV/ \sqrt{Hz}
Analog Outputs (Note 4)							
C_{OUT}	Output Capacitance (Note 4)	DAC Register Loaded to All 1s: C_{OUT1}	●		115	130	pF
		DAC Register Loaded to All 0s: C_{OUT1}	●		70	80	pF
Digital Inputs							
V_{IH}	Digital Input High Voltage		●	2.4			V
V_{IL}	Digital Input Low Voltage		●			0.8	V
I_{IN}	Digital Input Current		●		0.001	± 1	μ A
C_{IN}	Digital Input Capacitance	(Note 4) $V_{IN} = 0V$	●			8	pF
Timing Characteristics							
t_{DS}	Data to \overline{WR} Setup Time		●	60	20		ns
t_{DH}	Data to \overline{WR} Hold Time		●	0	-12		ns
t_{WR}	\overline{WR} Pulse Width		●	60	25		ns
t_{LD}	LD Pulse Width		●	110	55		ns
t_{CLR}	Clear Pulse Width		●	60	40		ns
t_{LWD}	\overline{WR} to LD Delay Time		●	0			ns
Power Supply							
V_{DD}	Supply Voltage		●	4.5	5	5.5	V
I_{DD}	Supply Current	Digital Inputs = 0V or V_{CC}	●			10	μ A

The ● denotes specifications that apply over the full operating temperature range.

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: $\pm 1LSB = \pm 0.006\%$ of full scale = $\pm 61ppm$ of full scale.

Note 3: Using internal feedback resistor.

Note 4: Guaranteed by design, not subject to test.

Note 5: $I_{(OUT1)}$ with DAC register loaded to all 0s.

Note 6: Typical temperature coefficient is 100ppm/ $^{\circ}C$.

Note 7: I_{OUT1} load = 100 Ω in parallel with 13pF.

Note 8: To 0.006% for a full-scale change, measured from the rising edge of LD.

Note 8: $V_{REF} = 0V$. DAC register contents changed from all 0s to all 1s or all 1s to all 0s.

Note 10: $V_{REF} = 6V_{RMS}$ at 1kHz. DAC register loaded with all 1s.

Note 11: Calculation from $e_n = \sqrt{4kTRB}$ where: k = Boltzmann constant (J/ $^{\circ}K$), R = resistance (Ω), T = temperature ($^{\circ}K$), B = bandwidth (Hz).

Note 12: Midscale transition code 0111 1111 1111 11 to 1000 0000 0000 00.

Note 13: R1 and R2 are measured between R1 and R_{COM} , REF and R_{COM} .

PIN FUNCTIONS

REF (Pin 1): Reference Input and 4-Quadrant Resistor R2. Typically $\pm 10V$, accepts up to $\pm 25V$. In 2-quadrant mode, this pin is the reference input. In 4-quadrant mode, this pin is driven by an external inverting reference amplifier.

R_{COM} (Pin 2): Center Tap Point of the Two 4-Quadrant Resistors R1 and R2. Normally tied to the inverting input of an external amplifier in 4-quadrant operation, otherwise shorted to the REF pin. See Figures 1 and 3.

R1 (Pin 3): 4-Quadrant Resistor R1. In 2-quadrant operation short to the REF pin. In 4-quadrant mode tie to the reference input.

R_{OFS} (Pin 4): Bipolar Offset Resistor. Typically swings $\pm 10V$, accepts up to $\pm 25V$. In 2-quadrant operation tie to R_{FB}. In 4-quadrant operation tie to R1.

R_{FB} (Pin 5): Feedback Resistor. Normally tied to the output of the current to voltage converter op amp. Swings to $\pm V_{REF}$. Typically V_{REF} is $\pm 10V$.

I_{OUT1} (Pin 6): DAC Current Output. Tie to the inverting input of the current to voltage converter op amp.

AGND (Pin 7): Analog Ground. Tie to ground.

LD (Pin 8): DAC Digital Input Load Control Input. When LD is taken to a logic high, data is loaded from the input register into the DAC register, updating the DAC output.

WR (Pin 9): DAC Digital Write Control Input. When \overline{WR} is taken to a logic low, data is loaded from the digital input pins into the 14-bit wide input register.

D13 to D2 (Pins 10 to 21): Digital Input Data Bits.

DGND (Pin 22): Digital Ground. Tie to ground.

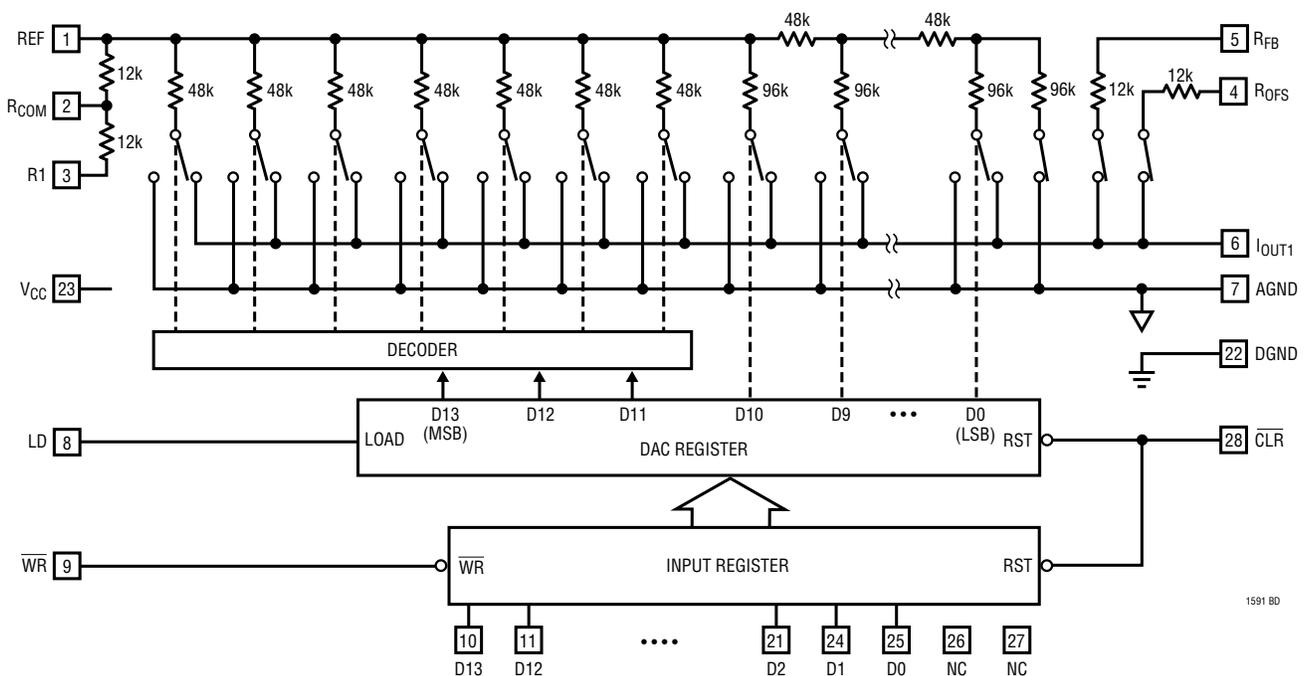
V_{CC} (Pin 23): The Positive Supply Input. $4.5V \leq V_{CC} \leq 5.5V$. Requires a bypass capacitor to ground.

D1, D0 (Pins 24, 25): Digital Input Data Bits.

NC (Pins 26, 27): Do not connect pins.

CLR (Pin 28): Digital Clear Control Function for the DAC. When \overline{CLR} is taken to a logic low, it sets the DAC output and all internal registers to zero code for the LTC1591 and midscale code for the LTC1591-1.

BLOCK DIAGRAM

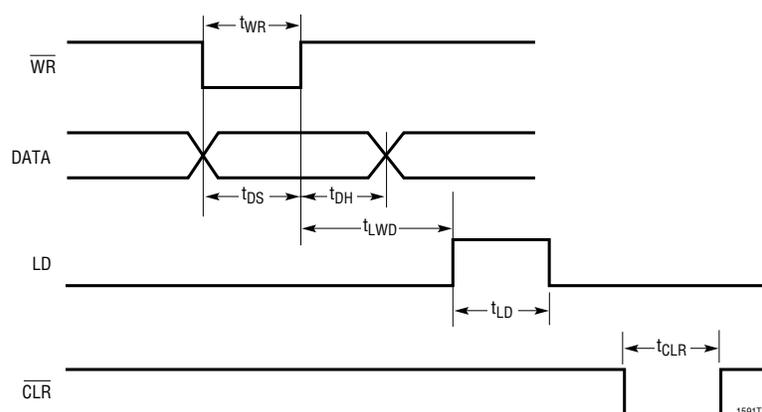


TRUTH TABLE

Table 1

CONTROL INPUTS			REGISTER OPERATION
CLR	WR	LD	
0	X	X	Reset Input and DAC Register to All 0s for LTC1591 and midscale for LTC1591-1 (Asynchronous Operation)
1	0	0	Load Input Register with All 14 Data Bits
1	1	1	Load DAC Register with the Contents of the Input Register
1	0	1	Input and DAC Register Are Transparent
1			CLK = LD and WR Tied Together. The 14 Data Bits Are Loaded into the Input Register on the Falling Edge of the CLK and Then Loaded into the DAC Register on the Rising Edge of the CLK
1	1	0	No Register Operation

TIMING DIAGRAM



APPLICATIONS INFORMATION

Description

The LTC1591 is a 14-bit multiplying, current output DAC with a full parallel 14-bit digital interface. The device operates from a single 5V supply and provides both unipolar 0V to $-10V$ or 0V to 10V and bipolar $\pm 10V$ output ranges from a 10V or $-10V$ reference input. It has three additional precision resistors on chip for bipolar operation. Refer to the block diagram regarding the following description.

The 14-bit DAC consists of a precision R-2R ladder for the 11LSBs. The 3MSBs are decoded into seven segments of resistor value R. Each of these segments and the R-2R ladder carries an equally weighted current of one eighth of full scale. The feedback resistor R_{FB} and 4-quadrant resistor R_{OFS} have a value of R/4. 4-quadrant resistors R1 and R2 have a magnitude of R/4. R1 and R2 together with

an external op amp (see Figure 3) inverts the reference input voltage and applies it to the 14-bit DAC input REF, in 4-quadrant operation. The REF pin presents a constant input impedance of R/8 in unipolar mode and R/12 in bipolar mode. The output impedance of the current output pin I_{OUT1} varies with DAC input code. The I_{OUT1} capacitance due to the NMOS current steering switches also varies with input code from 70pF to 115pF. An added feature of the LTC1591, especially for waveform generation, is a proprietary deglitcher that reduces glitch energy to below 2nV-s over the DAC output voltage range.

Digital Section

The LTC1591 has a 14-bit wide, full parallel data input bus. The device is double-buffered with two 14-bit registers. The double-buffered feature permits the update of several DACs simultaneously. The input register is loaded directly

APPLICATIONS INFORMATION

from a 14-bit microprocessor bus when the \overline{WR} pin is brought to a logic low level. The second register (DAC register) is updated with the data from the input register when the LD pin is brought to a logic high level. Updating the DAC register updates the DAC output with the new data. To make both registers transparent for flowthrough mode, tie \overline{WR} low and LD high. However, this defeats the deglitcher operation and output glitch impulse may increase. The deglitcher is activated on the rising edge of the LD pin. The versatility of the interface also allows the use of the input and DAC registers in a master slave or edge-triggered configuration. This mode of operation occurs when \overline{WR} and LD are tied together. The asynchronous clear pin resets the LTC1591 to zero scale and the LTC1591-1 to midscale. \overline{CLR} resets both the input and DAC registers. The device also has a power-on reset. Table 1 shows the truth table for the device.

Unipolar Mode (2-Quadrant Multiplying, $V_{OUT} = 0V$ to $-V_{REF}$)

The LTC1591 can be used with a single op amp to provide 2-quadrant multiplying operation as shown in Figure 1. With a fixed $-10V$ reference, the circuit shown gives a precision unipolar $0V$ to $10V$ output swing.

Bipolar Mode (4-Quadrant Multiplying, $V_{OUT} = -V_{REF}$ to V_{REF})

The LTC1591 contains on chip all the 4-quadrant resistors necessary for bipolar operation. 4-quadrant multiplying operation can be achieved with a minimum of external

components, a capacitor and a dual op amp, as shown in Figure 3. With a fixed $10V$ reference, the circuit shown gives a precision bipolar $-10V$ to $10V$ output swing.

Op Amp Selection

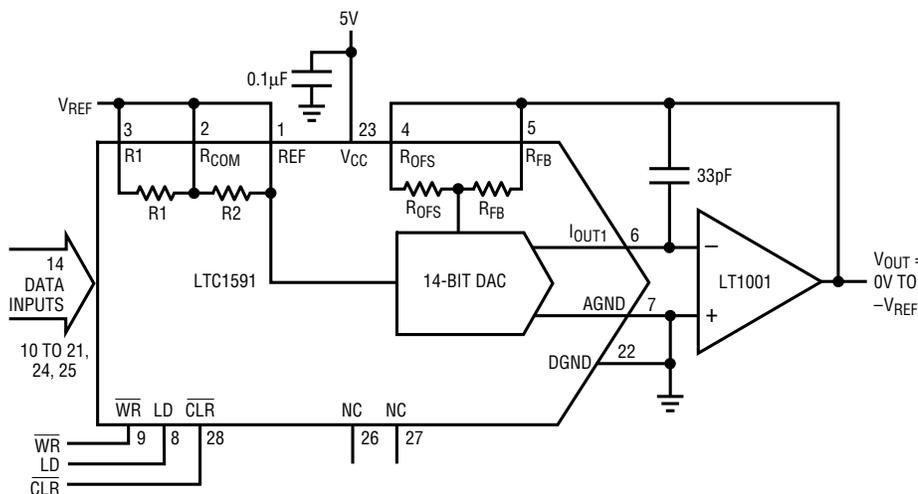
Because of the high accuracy of the 14-bit LTC1591, thought should be given to op amp selection in order to achieve the exceptional performance of which the part is capable. Fortunately, the sensitivity of INL and DNL to op amp offset has been greatly reduced compared to previous generations of multiplying DACs.

Op amp offset will contribute mostly to output offset and gain and will have minimal effect on INL and DNL. For the LTC1591, a $1mV$ op amp offset will cause about $0.28LSB$ INL degradation and $0.08LSB$ DNL degradation with a $10V$ full-scale range. The main effects of op amp offset will be a degradation of zero-scale error equal to the op amp offset, and a degradation of full-scale error equal to twice the op amp offset. For the LTC1591, the same $1mV$ op amp offset will cause a $1.65LSB$ zero-scale error and a $3.25LSB$ full-scale error with a $10V$ full-scale range.

Op amp input bias current (I_{BIAS}) contributes only a zero-scale error equal to $I_{BIAS}(R_{FB}/R_{OFS}) = I_{BIAS}(6k)$.

Grounding

As with any high resolution converter, clean grounding is important. A low impedance analog ground plane and star grounding should be used. AGND must be tied to the star ground with as low a resistance as possible.



Unipolar Binary Code Table

DIGITAL INPUT BINARY NUMBER IN DAC REGISTER		ANALOG OUTPUT V_{OUT}
MSB	LSB	
1111	1111 11	$-V_{REF}$ (16,383/16,384)
1000	0000 00	$-V_{REF}$ (8,192/16,384) = $-V_{REF}/2$
0000	0000 01	$-V_{REF}$ (1/16,384)
0000	0000 00	$0V$

1591 F01

Figure 1. Unipolar Operation (2-Quadrant Multiplication) $V_{OUT} = 0V$ to $-V_{REF}$

APPLICATIONS INFORMATION

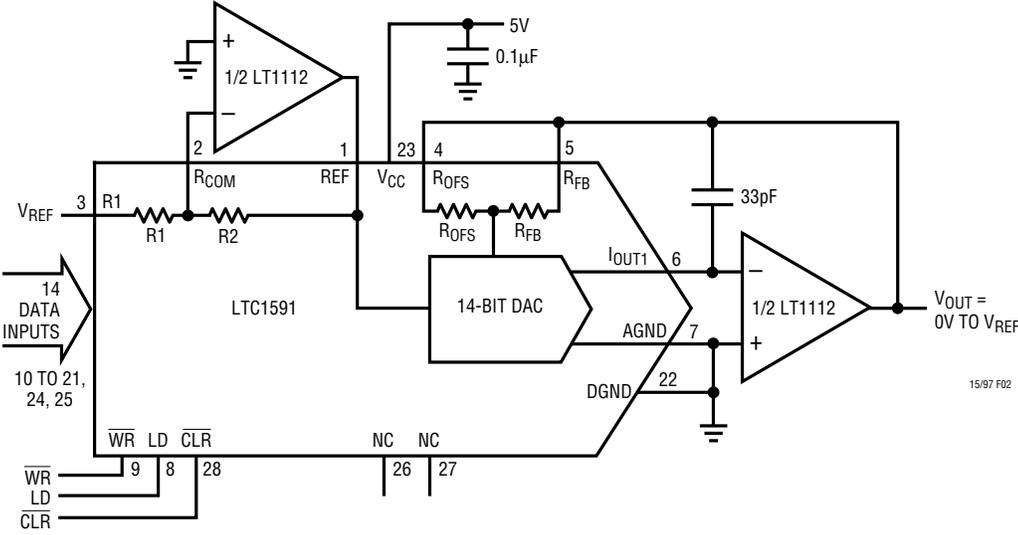
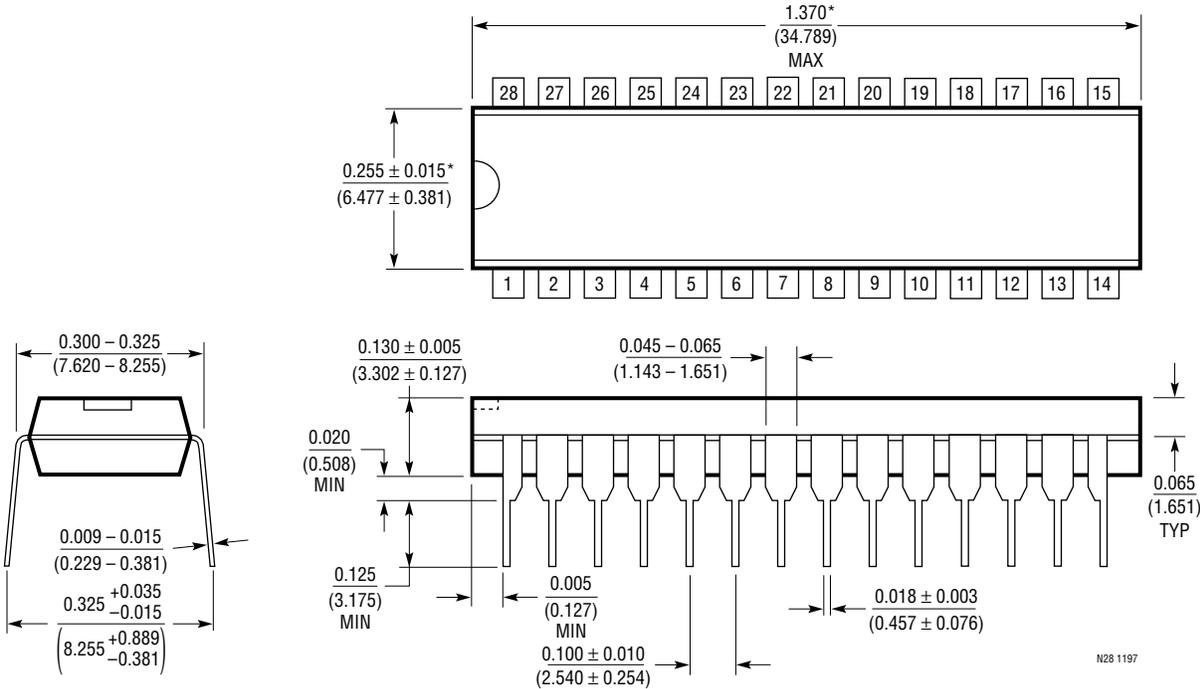


Figure 2. Noninverting Unipolar Operation (2-Quadrant Multiplication) $V_{OUT} = 0V$ to V_{REF}

PACKAGE DESCRIPTION

Dimensions in inches (millimeters) unless otherwise noted.

N Package
28-Lead PDIP (Narrow 0.300)
 (LTC DWG # 05-08-1510)

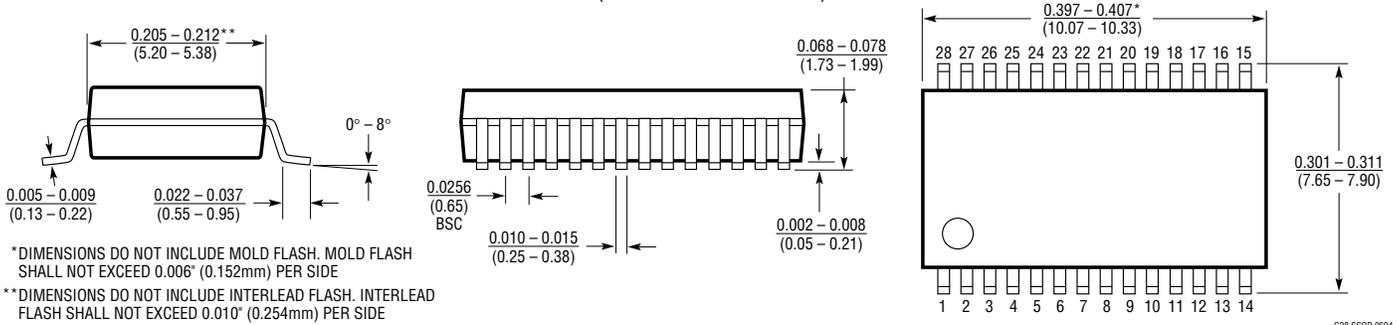


*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
 MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

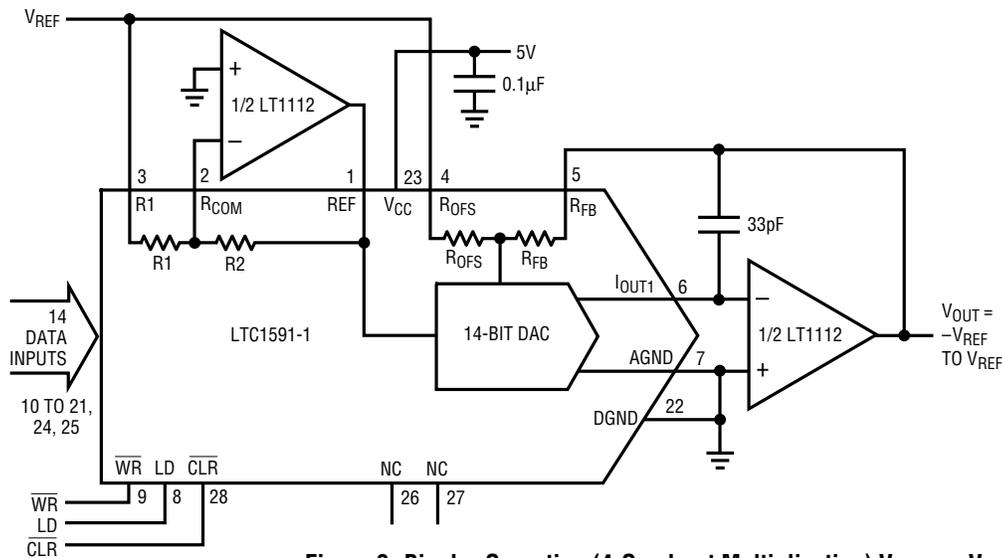
PACKAGE DESCRIPTION

Dimensions in inches (millimeters) unless otherwise noted.

G Package 28-Lead Plastic SSOP (0.209) (LTC DWG # 05-08-1640)



TYPICAL APPLICATION



Bipolar Offset Binary Code Table

DIGITAL INPUT BINARY NUMBER IN DAC REGISTER				ANALOG OUTPUT V_{OUT}
MSB	LSB			
1111	1111	1111	11	V_{REF} (8,191/8,192)
1000	0000	0000	01	V_{REF} (1/8,192)
1000	0000	0000	00	0V
0111	1111	1111	11	$-V_{REF}$ (1/8,192)
0000	0000	0000	00	$-V_{REF}$

Figure 3. Bipolar Operation (4-Quadrant Multiplication) $V_{OUT} = -V_{REF}$ to V_{REF}

RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS	
Op Amps	LT1001	Precision Operational Amplifier	Low Offset, Low Drift
	LT1112	Dual Low Power, Precision Picoamp Input Op Amp	Low Offset, Low Drift
	LT1468	90MHz, 22V/µs, 16-Bit Accurate Op Amp	Precise, 1µs Settling to 0.0015%
DACs	LTC1595/LTC1596	Serial 16-Bit Current Output DACs	Low Glitch, ±1LSB Maximum INL, DNL
	LTC1597/LTC1597-1	Parallel 16-Bit Current Output DACs	On-Chip 4-Quadrant Resistors
	LTC1650	Serial 16-Bit Voltage Output DAC	Low Noise and Glitch Rail-to-Rail V_{OUT}
	LTC1658	Serial 14-Bit Voltage Output DAC	Low Power, 8-Lead MSOP Rail-to-Rail V_{OUT}
ADCs	LTC1418	14-Bit, 200ksps, 5V Sampling ADC	15mW Dissipation, Serial and Parallel Outputs
	LTC1604	16-Bit, 333ksps Sampling ADC	±2.5V Input, SINAD = 90dB, THD = 100dB
	LTC1605	Single 5V, 16-Bit 100ksps ADC	Low Power, ±10V Inputs