

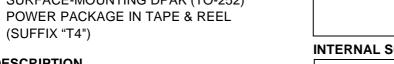
## STD55NH02L

# N-CHANNEL 24V - $0.008 \Omega$ - 55A DPAK/IPAK ULTRA LOW GATE CHARGE STripFET™ POWER MOSFET

#### **TARGET DATA**

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	I <sub>D</sub>	
STD55NH02L	24 V	< 0.010 Ω	55 A	

- TYPICAL  $R_{DS}(on) = 0.008 \Omega$  @ 10 V
- TYPICAL R<sub>DS</sub>(on) = 0.011  $\Omega$  @ 4.5 V
- R<sub>DS(ON)</sub> \* Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- THROUGH-HOLE IPAK (TO-251) POWER PACKAGE IN TUBE (SUFFIX "-1")
- SURFACE-MOUNTING DPAK (TO-252) POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")



#### **DESCRIPTION**

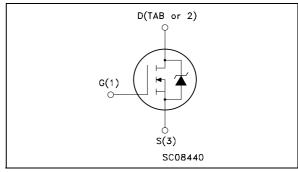
The STD55NH02L is based on the latest generation of ST's proprietary STripFET™ technology. An innovative layout enables the device to also exhibit extremely low gate charge for the most demanding requirements as high-side switch in high-frequency DC-DC converters. It's therefore ideal for high-density converters in Telecom and Computer applications.

#### **APPLICATIONS**

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTES

# **IPAK** TO-251 TO-252 (Suffix "-1") (Suffix "T4")

#### INTERNAL SCHEMATIC DIAGRAM



#### **Ordering Information**

<u> </u>			
SALES TYPE	MARKING	PACKAGE	PACKAGING
STD55NH02LT4	D55NH02L	TO-252	TAPE & REEL
STD55NH02L-1	D55NH02L	TO-251	TUBE

#### **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit
V <sub>spike</sub> (1)	Drain-source Voltage Rating	30	V
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	24	V
$V_{DGR}$	Drain-gate Voltage ( $R_{GS} = 20 \text{ k}\Omega$ )	24	V
$V_{GS}$	Gate- source Voltage	± 18	V
I <sub>D</sub>	Drain Current (continuous) at T <sub>C</sub> = 25°C	55	Α
I <sub>D</sub>	Drain Current (continuous) at T <sub>C</sub> = 100°C	39	Α
I <sub>DM</sub> (2)	Drain Current (pulsed)	220	Α
P <sub>tot</sub>	Total Dissipation at T <sub>C</sub> = 25°C	60	W
	Derating Factor	0.4	W/°C
E <sub>AS</sub> (3)	Single Pulse Avalanche Energy	TBD	mJ
T <sub>stg</sub>	Storage Temperature	-55 to 175	°C
Tj	Operating Junction Temperature	-55 to 175	

#### THERMAL DATA

# **ELECTRICAL CHARACTERISTICS** ( $T_{CASE} = 25~^{\circ}C$ UNLESS OTHERWISE SPECIFIED) OFF

Symbol	Parameter Test Conditions		Min.	Тур.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	$I_D = 250 \mu\text{A},  V_{GS} = 0$	24			V
I <sub>DSS</sub>	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	$V_{DS} = Max Rating$ $V_{DS} = Max Rating T_C = 125^{\circ}C$			1 10	μA μA
I <sub>GSS</sub>	Gate-body Leakage Current (V <sub>DS</sub> = 0)	V <sub>GS</sub> = ± 18V			±100	nA

#### ON (4)

Symbol	Parameter	Test C	Test Conditions		Тур.	Max.	Unit
V <sub>GS(th)</sub>	Gate Threshold Voltage	$V_{DS} = V_{GS}$	I <sub>D</sub> = 250 μA	1			V
R <sub>DS(on)</sub>	Static Drain-source On Resistance	V <sub>GS</sub> = 10 V V <sub>GS</sub> = 4.5 V	I <sub>D</sub> = 27.5 A I <sub>D</sub> = 27.5 A		0.008 0.011	0.010 0.014	Ω Ω

#### **DYNAMIC**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
gfs (4)	Forward Transconductance	$V_{DS} = 20 \text{ V}$ $I_{D} = 27.5 \text{ A}$		TBD		S
C <sub>iss</sub> C <sub>oss</sub> C <sub>rss</sub>	Input Capacitance Output Capacitance Reverse Transfer Capacitance	V <sub>DS</sub> = 10V f = 1 MHz V <sub>GS</sub> = 0		860 450 56		pF pF pF
R <sub>G</sub>	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.5		Ω

#### **ELECTRICAL CHARACTERISTICS** (continued)

#### **SWITCHING ON**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(on)</sub> t <sub>r</sub>	Turn-on Delay Time Rise Time	$\begin{split} V_{DD} &= 10 \text{ V} & I_D = 27.5 \text{ A} \\ R_G &= 4.7 \Omega & V_{GS} = 10 \text{ V} \\ \text{(Resistive Load, Figure 3)} \end{split}$		TBD TBD		ns ns
Q <sub>g</sub> Q <sub>gs</sub> Q <sub>gd</sub>	Total Gate Charge Gate-Source Charge Gate-Drain Charge	$0.44V \le V_{DD} \le 10V$ , $I_{D} = 55 \text{ A}$ $V_{GS} = 4.5 \text{ V}$		9 TBD TBD	12	nC nC nC
Q <sub>oss</sub> (5)	Output Charge	V <sub>DS</sub> = 10 V V <sub>GS</sub> = 0 V		TBD		nC

#### **SWITCHING OFF**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(off)</sub>	Turn-off Delay Time Fall Time	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$		TBD TBD		ns ns

#### **SOURCE DRAIN DIODE**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I <sub>SD</sub>	Source-drain Current Source-drain Current (pulsed)				55 220	A A
V <sub>SD</sub> (4)	Forward On Voltage	I <sub>SD</sub> = 27.5 A V <sub>GS</sub> = 0			1.3	V
t <sub>rr</sub> Q <sub>rr</sub> I <sub>RRM</sub>	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 55 \text{ A}$ di/dt = 100A/ $\mu$ s $V_{DD} = 20 \text{ V}$ $T_j = 150^{\circ}\text{C}$ (see test circuit, Figure 5)		TBD TBD TBD		ns nC A

<sup>(1)</sup> Garanted when external Rg=4.7  $\Omega$  and  $t_f < t_{fmax}$ . (2) Pulse width limited by safe operating area (3) Starting  $T_j = 25$  °C,  $I_D = 25$ A,  $V_{DD} = 15$ V

<sup>(4)</sup> Pulsed: Pulse duration = 300  $\mu$ s, duty cycle 1.5 %. (5)  $Q_{oss} = C_{oss}^* \Delta V_{in}$ ,  $C_{oss} = C_{gd} + C_{ds}$ . See Appendix A

Fig. 1: Unclamped Inductive Load Test Circuit

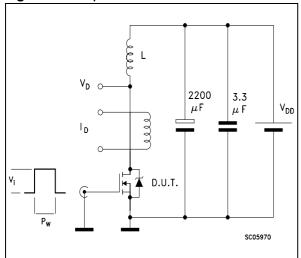
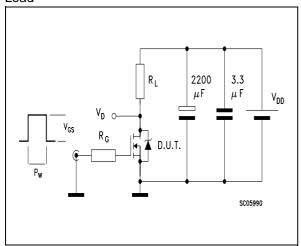


Fig. 3: Switching Times Test Circuits For Resistive Load



**Fig. 5:** Test Circuit For Inductive Load Switching And Diode Recovery Times

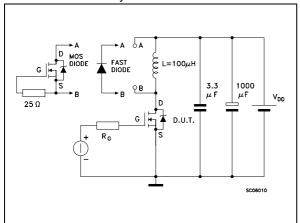


Fig. 2: Unclamped Inductive Waveform

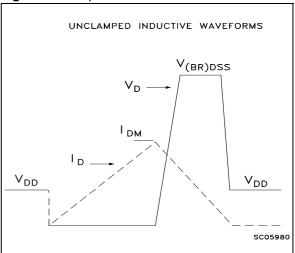
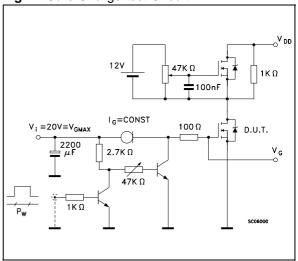
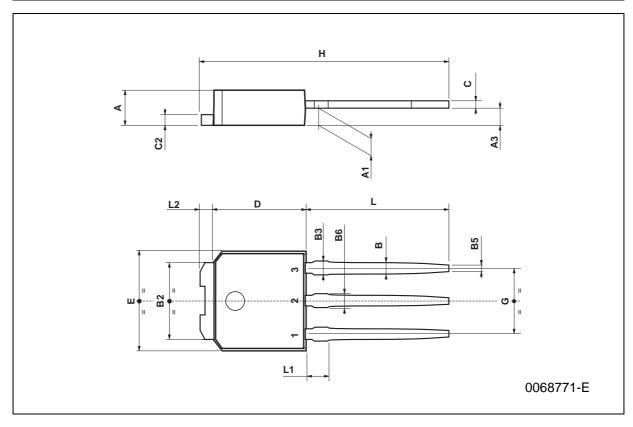


Fig. 4: Gate Charge test Circuit



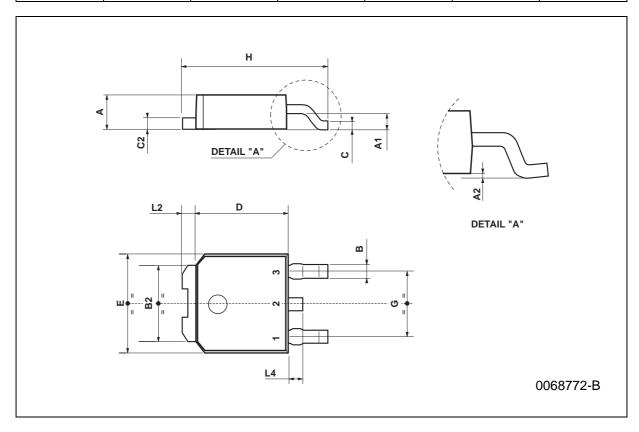
# TO-251 (IPAK) MECHANICAL DATA

DIM.		mm		inch		
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A3	0.7		1.3	0.027		0.051
В	0.64		0.9	0.025		0.031
B2	5.2		5.4	0.204		0.212
В3			0.85			0.033
B5		0.3			0.012	
B6			0.95			0.037
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
E	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	15.9		16.3	0.626		0.641
L	9		9.4	0.354		0.370
L1	0.8		1.2	0.031		0.047
L2		0.8	1		0.031	0.039



# **TO-252 (DPAK) MECHANICAL DATA**

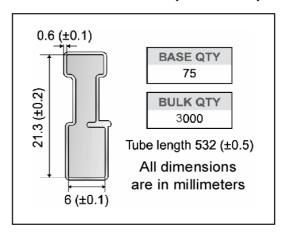
DIM.		mm		inch		
Diwi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A2	0.03		0.23	0.001		0.009
В	0.64		0.9	0.025		0.035
B2	5.2		5.4	0.204		0.212
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
Е	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	9.35		10.1	0.368		0.397
L2		0.8			0.031	
L4	0.6		1	0.023		0.039



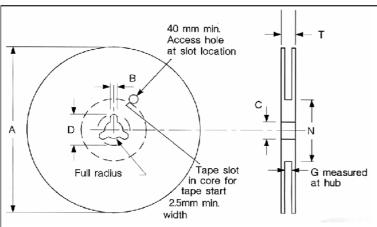
#### **DPAK FOOTPRINT**

# 6.7 1.8 3.0 1.6 2.3 1.6 All dimensions are in millimeters

### **TUBE SHIPMENT (no suffix)\***



## TAPE AND REEL SHIPMENT (suffix "T4")\*

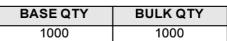


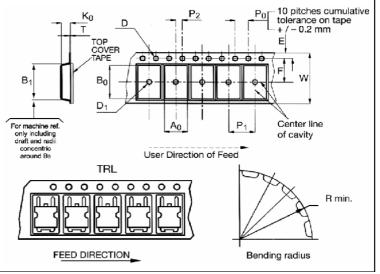
#### REEL MECHANICAL DATA

DIM.	mm		inch	
	MIN.	MAX.	MIN.	MAX.
Α		330		12.992
В	1.5		0.059	
С	12.8	13.2	0.504	0.520
D	20.2		0.795	
G	16.4	18.4	0.645	0.724
N	50		1.968	
T		22.4		0.881

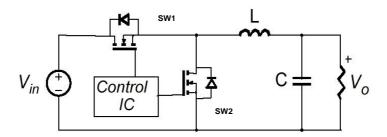
#### TAPE MECHANICAL DATA

DIM.	mm		inch	
	MIN.	MAX.	MIN.	MAX.
A0	6.8	7	0.267	0.275
В0	10.4	10.6	0.409	0.417
B1		12.1		0.476
D	1.5	1.6	0.059	0.063
D1	1.5		0.059	
E	1.65	1.85	0.065	0.073
F	7.4	7.6	0.291	0.299
K0	2.55	2.75	0.100	0.108
P0	3.9	4.1	0.153	0.161
P1	7.9	8.1	0.311	0.319
P2	1.9	2.1	0.075	0.082
R	40		1.574	
W	15.7	16.3	0.618	0.641





# **APPENDIX A Buck Converter: Power Losses Estimation**



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is converted to allow for a safer working junction temperature.

The low side (SW2) device requires:

- $\bullet \qquad \text{Very low } R_{DS(on)} \text{ to reduce conduction losses} \\$
- $\bullet \qquad Small \ Q_{gls} \ to \ reduce \ the \ gate \ charge \ losses$
- Small Coss to reduce losses due to output capacitance
- Small Q<sub>rr</sub> to reduce losses on SW<sub>1</sub> during its turn-on
- The  $C_{gd}/C_{gs}$  ratio lower than  $V_{th}/V_{gg}$  ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

- $\bullet$  Small  $R_g$  and  $L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate
- Small Q<sub>g</sub> to have a faster commutation and to reduce gate charge losses
- $\bullet \qquad \text{Low } R_{DS(on)} \text{ to reduce the conduction losses}.$

		High Side Switch (SW1)	Low Side Switch (SW2)
Pconducti	on	$R_{DS(on)SW1} * I_L^2 * d$	$R_{DS(on)SW2} * I_L^2 * (1-d)$
Pswitchin	g	$V_{\text{in}} * (Q_{\text{gsth(SW1)}} + Q_{\text{gd(SW1)}}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P <sub>diode</sub>	Recovery	Not Applicable	<sup>1</sup> V <sub>in</sub> *Q <sub>rr(SW2)</sub> * f
	Conduction	Not Applicable	$V_{f(SW2)} * I_L * t_{deadtime} * f$
$P_{\text{gate}(Q_G)}$	)	$Q_{g(SW1)}*V_{gg}*f$	$Q_{gls(SW2)}*V_{gg}*f$
P <sub>Qoss</sub>		$\frac{V_{in} *Q_{oss(SW1)} *f}{2}$	$\frac{V_{in} *Q_{oss(SW2)} *f}{2}$

Parameter	Meaning	
d	Duty-cycle	
Qgsth	Post threshold gate charge	
$Q_{ m gls}$	Third quadrant gate charge	
Pconduction	On state losses	
Pswitching	On-off transition losses	
Pdiode	Conduction and reverse recovery diode losses	
Pgate	Gate drive losses	
P <sub>Qoss</sub>	Output capacitance losses	

<sup>&</sup>lt;sup>1</sup> Dissipated by SW1 during turn-on

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